

# **SPT5420**

# 13-BIT, OCTAL D/A CONVERTER

# **TECHNICAL DATA**

JUNE 26, 2001

### **FEATURES**

- 13-bit resolution
- Pin compatible with AD7839
- · Eight DACs in one package
- · Buffered voltage outputs
- Wide output voltage swing  $V_{DD}$ -2.5 V to  $V_{SS}$ +2.5 V
- 15 µs settling time to ±0.5 LSB
- · Double-buffered digital inputs
- · Microprocessor and TTL/CMOS compatible

# GENERAL DESCRIPTION

The SPT5420 contains eight 13-bit digital-to-analog CMOS converters designed primarily for automatic test equipment applications. It uses novel circuit topology to convert the 13-bit digital inputs into output voltages which are proportionate to the applied reference voltages. Each

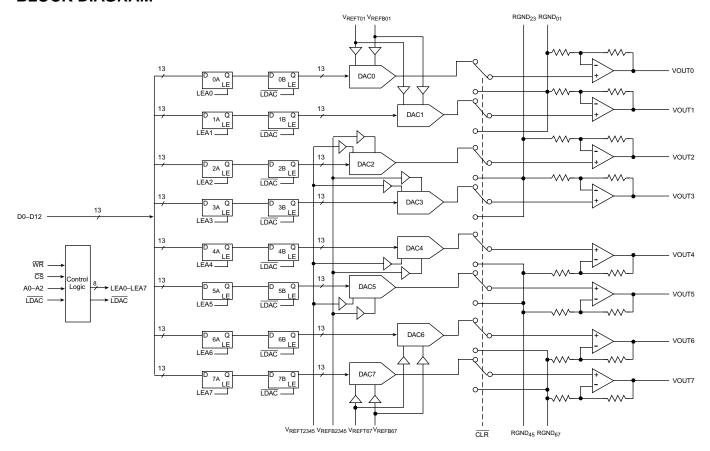
### **APPLICATIONS**

- Automatic test equipment
- Instrumentation
- Process control

DAC's full-scale output voltage and output voltage offset are adjustable with analog inputs (RGND, V<sub>REFB</sub>, V<sub>REFT</sub>).

The SPT5420 operates over an industrial temperature range of –40 °C to +85 °C and is available in a 10 x 10 mm, 44-lead metric quad flat pack (MQFP) plastic package.

### **BLOCK DIAGRAM**



# ABSOLUTE MAXIMUM RATINGS (Beyond which damage may occur)1 25 °C

#### 

# **ELECTRICAL SPECIFICATIONS**

 $T_{A} = T_{MIN} \text{ to } T_{MAX}, V_{CC} = +5.0 \text{ V}, V_{DD} = +11.5 \text{ V}, V_{SS} = -8.0 \text{ V}, V_{REFT} = 3.5 \text{ V}, V_{REFB} = -1.5 \text{ V}, R_{L} = +10 \text{ k}\Omega, C_{L} = 50 \text{ pF, unless otherwise specified.}$ 

PARAMETERS	TEST CONDITIONS	TEST LEVEL	MIN	SPT5420 TYP	MAX	UNITS
Accuracy Resolution Integral Linearity Error (ILE) Differential Linearity Error (DLE) Zero-Scale Error Full Scale Error Gain Error		VI VI VI VI VI	13 -2.0 -1.0 -25 -25 -25	±0.5 ±0.3	+2.0 +1.0 +25 +25 +25	Bits LSB LSB mV mV
Reference Inputs Input Current VREFT <sup>1</sup> VREFB <sup>2</sup>		IV VI VI	0 -5.0	+3.5 -1.5	±100 +5.0 0	nA V V
RGND Inputs DC Input Impedance Input Range		V IV	-2.0	60	2.0	kΩ V
Output Characteristics Output Swing <sup>3,4</sup> Short Circuit Current Resistive Load DC Output Impedance		VI IV VI IV	5	+7/-3	15 1.0	V mA kΩ
Digital Inputs Logic 1 Voltage Logic 0 Voltage Maximum Input Current Input Capacitance		VI VI VI V	2.4 -10	10	0.8 10	V V μΑ/pin pF

### Notes:

4. 
$$V_{OUT} = 2 X (V_{REFB} + [V_{REFT} - V_{REFB}] X \frac{INPUT CODE}{8192}) - V_{RGND}$$

<sup>1.</sup>  $V_{REFT}$  < 8 V + ( $V_{SS}$  x 0.5); e.g., if  $V_{SS}$  = -8 V, then  $V_{REFT}$  < 4 V

<sup>2.</sup>  $V_{REFB} > (V_{DD} \times 0.5) - 9.5 \text{ V}$ ; e.g., if  $V_{DD} = 11 \text{ V}$ , then  $V_{REFB} > -4 \text{ V}$ 

<sup>3.</sup>  $V_{SS} + 2.5 \text{ V} \le V_{OUT} \le V_{SS} + 16.0 \text{ V}$  for  $18.5 \text{ V} \le V_{DD} - V_{SS} \le 20.0 \text{ V}$   $V_{SS} + 2.5 \text{ V} \le V_{OUT} \le V_{DD} - 2.5 \text{ V}$  for  $V_{DD} - V_{SS} \le 18.5 \text{ V}$ 

## **ELECTRICAL SPECIFICATIONS**

 $T_A = T_{MIN}$  to  $T_{MAX}$ ,  $V_{CC} = +5.0$  V,  $V_{DD} = +11.5$  V,  $V_{SS} = -8.0$  V,  $V_{REFT} = 3.5$  V,  $V_{REFB} = -1.5$  V,  $V_{R} = +10$  k $\Omega$ ,  $C_L = 50$  pF, unless otherwise specified.

PARAMETERS	TEST CONDITIONS	TEST LEVEL	MIN	SPT5420 TYP	MAX	UNITS
Power Requirements  V <sub>CC</sub> Supply Voltage (Digital)  V <sub>DD</sub> Supply Voltage (Analog)1,2  V <sub>SS</sub> Supply Voltage (Analog)1,2  I <sub>CC</sub> Supply Current  I <sub>DD</sub> Supply Current  I <sub>SS</sub> Supply Current  Power Supply Rejection Ratio	Outputs Unloaded Outputs Unloaded ΔV <sub>DD</sub> / ΔFull Scale ΔV <sub>SS</sub> / ΔFull Scale	IV VI VI VI VI IV IV	4.75 5 –12.5	5 11.5 -8 5 5 80 80	5.25 12.5 -5 0.5 10	V V V mA mA dB dB
Dynamic Performance Output Settling Time3 (Full Scale Change to ±0.5 LSB) Slew Rate Glitch Impulse Channel to Channel Isolation DAC to DAC Crosstalk Digital Crosstalk Digital Feedthrough	C <sub>L</sub> ≤ 220 pF	IV V V V V		2.0 35 100 40 1	15	μs V/μs nV-s dB nV-s nV-s
Timing Characteristics (See page 4)		IV				

<sup>1.</sup> Supplies should provide 2.5 V headroom above and below max output swing.

### **DEFINITION OF SELECTED TERMINOLOGY**

### **Channel-to-Channel Isolation**

Channel-to-Channel isolation refers to the proportion of input signal from one DAC's reference input that appears at the output of the other DAC. It is expressed in dBs.

### **DAC-to-DAC Crosstalk**

DAC-to-DAC crosstalk is defined as the glitch impulse that appears at one DAC's output due to both the digital change and subsequent analog output change at any other DAC. It is specified in nV-s.

### **Digital Crosstalk**

The glitch impulse transferred to one DAC's output due to a change in digital input code of any other DAC. It is specified in nV-s.

### **Digital Feedthrough**

Digital feedthrough is the noise at a DAC's output caused by changes to D0-D12 while WR is high.

TEST LEVEL CODES	LEVEL	TEST PROCEDURE
All electrical characteristics are subject to the following conditions:  All parameters having min/max specifications are guaranteed. The Test Level column indicates the specific device testing actually performed during production and Quality Assurance inspection. Any blank section in the data	1 11 111	100% production tested at the specified temperature. 100% production tested at $T_A$ = +25 °C, and sample tested at the specified temperatures. QA sample tested only at the specified temperatures.
	IV	Parameter is guaranteed (but not tested) by design and characterization data.
column indicates that the specification is not	V	Parameter is a typical value for information purposes only.
tested at the specified condition.	VI	100% production tested at $T_A$ = +25 °C. Parameter is guaranteed over specified temperature range.

<sup>2.</sup>  $V_{DD} - V_{SS} \le 20 \text{ V}$ 

<sup>3.</sup> Output can drive 10,000 pF without oscillation, but with settling time degradation.

# **TIMING CHARACTERISTICS**

Figure 1a – Timing Diagram: Latched Mode (LDAC Strobed)

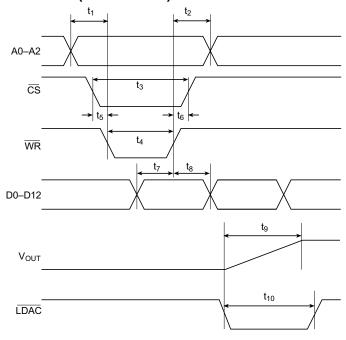
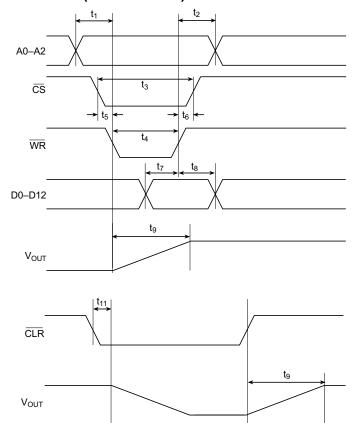


Figure 1b – Timing Diagram: Transparent Mode (LDAC Held Low)



PARAMETER	SYMBOL	MIN	TYP	MAX	UNIT
Address Valid to WR Setup	t <sub>1</sub>	20			ns
Address Valid to WR Hold	$t_2$	0			ns
CS Pulse Width Low	$t_3$	50			ns
WR Pulse Width Low	t <sub>4</sub>	50			ns
CS to WR Setup	t <sub>5</sub>	0			ns
WR to CS Hold	t <sub>6</sub>	0			ns
Data Setup	t <sub>7</sub>	25			ns
Data Hold	t <sub>8</sub>	0			ns
Settling Time <sup>1</sup>	t <sub>9</sub>			15	us
LDAC Pulse Width Low	t <sub>10</sub>	50			ns
CLR Pulse Activation	t <sub>11</sub>			300	ns
<del>_</del>					

NOTES:

All digital input rise and fall times are measured from 10% to 90% of +5 V.  $t_{\text{r}}$  =  $t_{\text{f}}$  = 5 ns.

1.  $R_L = 10 \text{ k}\Omega$ 

 $C_L \le 220 \text{ pF}$ 

# VOLTAGE REFERENCES AND ANALOG GROUND INPUTS

Three  $V_{REFTXX}$  and three  $V_{REFBXX}$  inputs set the output range of the three corresponding groups of DACs (0 and 1; 2 through 5; 6 and 7). Four RGND<sub>XX</sub> inputs set the output offset voltage of the four corresponding groups of DACs (0 and 1; 2 and 3; 4 and 5; 6 and 7). The formula for output swing and offset is presented in the "Analog Outputs" section below.

### DAC ADDRESSING AND LATCHING

Each DAC has an input latch which receives data from the data bus, and a DAC latch which receives data from the input latch. The analog output of each DAC corresponds to the data in its DAC latch. One of the eight input latches is addressed by the address lines A(2:0) according to Table I. While CS and WR are low, the addressed input latch is transparent and the seven other input latches are latched. Bringing CS or WR high latches data into the addressed input latch. While LDAC is low, all eight DAC latches are transparent. Bringing LDAC high latches data into the DAC latches. While CS, WR and LDAC are low, both latches are transparent and input data is transferred directly to the selected DAC. While CLR is low, all DAC outputs are set to their corresponding RGND<sub>XX</sub>. Bringing CLR high returns each DAC's output to the voltage corresponding to the data in each DAC latch.

Table II summarizes this information, and figures 1a and 1b should be referenced for timing limitations.

#### POWER SUPPLY SEQUENCING

The sequence in which  $V_{DD}$ ,  $V_{SS}$  and  $V_{CC}$  come up is not critical. The reference inputs,  $V_{REFTXX}$  and  $V_{REFBXX}$ , must come on only after  $V_{DD}$  and  $V_{SS}$  have been established. However, they may be turned on prior to  $V_{CC}$ . The digital inputs must be driven only after  $V_{DD}$ ,  $V_{SS}$  and  $V_{CC}$  have been established. Reverse the power-on sequence for power-down.

# ANALOG OUTPUTS VS DIGITAL INPUT CODE

The output voltage range is equal to twice the difference between  $V_{REFTXX}$  and  $V_{REFBXX}$ . The output voltage is given by:

$$V_{OUT} = 2 \text{ X } (V_{REFB} + [V_{REFT} - V_{REFB}] \text{ X } \frac{INPUT CODE}{8192}) - V_{RGND}$$

$$CODE = 0 - 8191$$

Table I - DAC Addressing

A2	<b>A</b> 1	Α0	Addressed Input Latch DAC#	
0	0	0	0	
0	0	1	1	
0	1	0	2	
0	1	1	3	
1	0	0	4	
1	0	1	5	
1	1	0	6	
1	1	1	7	

Table II - Control Logic Table

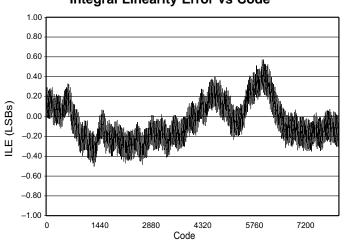
$\overline{WR}$	CS	LDAC	CLR	Input Latch	DAC Latch
0	0	Х	1	transparent1	Х
1	Х	Х	1	latched	Х
Х	1	Х	1	latched	Х
Х	Х	0	1	Х	transparent
Х	Х	1	1	Х	latched
Х	Х	Х	0	DAC outputs a	t RGND <sub>XX</sub>

#### Note

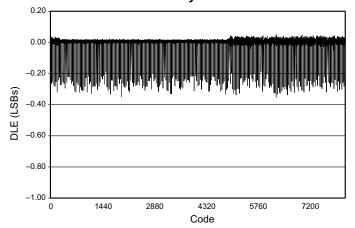
<sup>1.</sup> Only the input latch addressed by A(2:0) is transparent. The other input latches are latched.

## TYPICAL PERFORMANCE CHARACTERISTICS

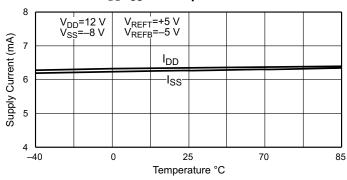
# Integral Linearity Error vs Code



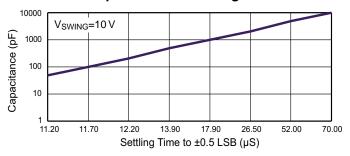
## **Differential Linearity Error vs Code**



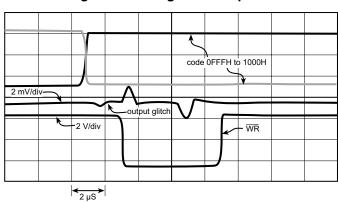
# I<sub>DD</sub>/I<sub>SS</sub> vs Temperature



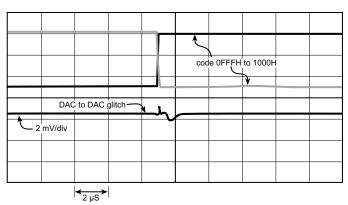
# **Load Capacitance vs Settling Time**



# Digital-to-Analog Glitch Impulse

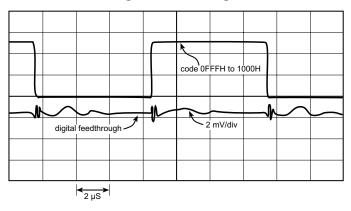


## **DAC to DAC Crosstalk**

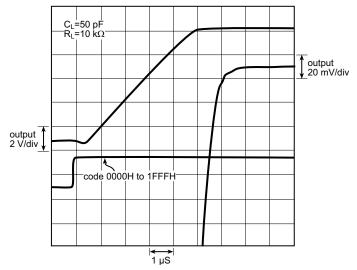


# **TYPICAL PERFORMANCE CHARACTERISTICS**

# **Digital Feedthrough**

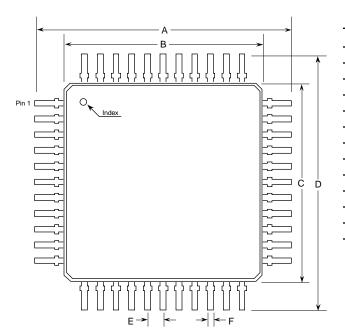


# Slew and Settling Time

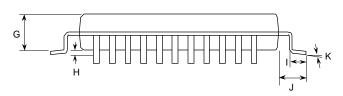


# **PACKAGE OUTLINE**

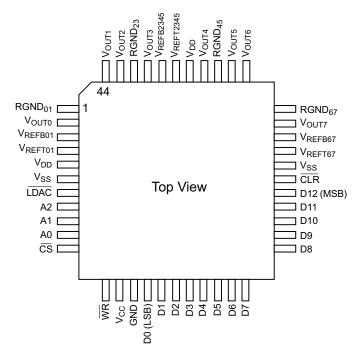
## 44-Lead MQFP



	INC	HES	MILLI	METERS
SYMBOL	MIN	MAX	MIN	MAX
Α	0.5098	0.5295	12.95	13.45
В	0.3917	0.3957	9.95	10.05
С	0.3917	0.3957	9.95	10.05
D	0.5098	0.5295	12.95	13.45
E	0.0311	0.0319	0.79	0.81
F	0.0118	0.0177	0.30	0.45
G	0.0768	0.0827	1.95	2.10
Н	0.0039	0.0098	0.10	0.25
ı	0.0287	0.0406	0.73	1.03
J	0.0630 REF		1.60	REF
K	0°	7°	0°	7°



### **PIN ASSIGNMENTS**



### PIN FUNCTIONS

Name	Function
DIGITAL	. CONTROL PINS
CS	Chip Select (Active Low)
WR	Level Triggered Write Input (Active Low). Used in conjunction with $\overline{\text{CS}}$ to write data to the SPT5420 input data latches. Data is latched into selected input data latch on the rising edge of $\overline{\text{WR}}$ .

CLR	(Active Low) Analog Clear. Sets the output voltages to RGND. (Each RGND is common to a DAC pair.) CLR does not reset the digital latches. When CLR is brought back high, the DAC outputs revert back to their original outputs as determined by the data in their DAC latches.
LDAC	When this logic input is taken low, the contents of the input latches are transferred to their respective DAC latches. (Active Low) Data is latched on rising edge.
A0 – A2	Addresses DAC0 to DAC7 for loading the eight input latches.
D0 – D12	Digital Inputs (D0 = LSB)
ANALOG	PINS
V <sub>REFT01</sub>	Top Reference Voltage for DACs 0 and 1
V <sub>REFT2345</sub>	Top Reference Voltage for DACs 2, 3, 4 and 5
V <sub>REFT67</sub>	Top Reference Voltage for DACs 6 and 7
V <sub>REFB01</sub>	Bottom Reference Voltage for DACs 0 and 1
$V_{REFB2345}$	Bottom Reference Voltage for DACs 2, 3, 4 and 5
V <sub>REFB67</sub>	Bottom Reference Voltage for DACs 6 and 7
RGND <sub>01</sub>	Reference Ground for Output Amplifiers 0 and 1
RGND <sub>23</sub>	Reference Ground for Output Amplifiers 2 and 3
RGND <sub>45</sub>	Reference Ground for Output Amplifiers 4 and 5
RGND <sub>67</sub>	Reference Ground for Output Amplifiers 6 and 7
V <sub>OUT0-7</sub>	Output Voltage Pins for DAC0 - DAC7
POWER S	UPPLY PINS
$V_{CC}$	Digital +5 V Supply
$V_{DD}$	Analog +11.5 V Supply (Nominal)
$V_{SS}$	Analog –8 V Supply (Nominal)
GND	Ground

## ORDERING INFORMATION

PART NUMBER	TEMPERATURE RANGE	PACKAGE
SPT5420SIM	−40 to +85 °C	44L MQFP

### **DISCLAIMER**

FAIRCHILD SEMICONDUCTOR RESERVES THE RIGHT TO MAKE CHANGES WITHOUT FURTHER NOTICE TO ANY PRODUCTS HEREIN TO IMPROVE RELIABILITY, FUNCTION OR DESIGN. FAIRCHILD DOES NOT ASSUME ANY LIABILITY ARISING OUT OF THE APPLICATION OR USE OF ANY PRODUCT OR CIRCUIT DESCRIBED HEREIN; NEITHER DOES IT CONVEY ANY LICENSE UNDER ITS PATENT RIGHTS, NOR THE RIGHTS OF OTHERS.

### LIFE SUPPORT POLICY

FAIRCHILD'S PRODUCTS ARE NOT AUTHORIZED FOR USE AS CRITICAL COMPONENTS IN LIFE SUPPORT DEVICES OR SYSTEMS WITHOUT THE EXPRESS WRITTEN APPROVAL OF THE PRESIDENT OF FAIRCHILD SEMICONDUCTOR CORPORATION. As used herein:

- 1. Life support devices or systems are devices or systems which, (a) are intended for surgical implant into the body, or (b) support or sustain life, and whose failure to perform, when properly used in accordance with instructions for use provided in the labeling, can be reasonably expected to result in a significant injury to the user.
- A critical component is any component of a life support device or system whose failure to perform can be reasonably expected to cause the failure of the life support device or system, or to affect its safety or effectiveness.

www.fairchildsemi.com

© Copyright 2002 Fairchild Semiconductor Corporation