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**2.5MHz, Precision Operational Amplifier**

The Intersil HA-5135 is a precision operational amplifier manufactured using a combination of key technological advancements to provide outstanding input characteristics.

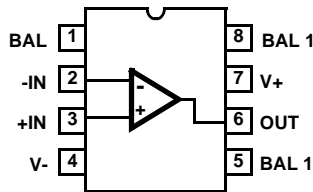
A Super Beta input stage is combined with laser trimming, dielectric isolation and matching techniques to produce 75µV (Maximum) input offset voltage and 0.4µV/°C input offset voltage average drift. Other features enhanced by this process include 9nV/√Hz (Typ) Input Noise Voltage, 1nA Input Bias Current and 140dB Open Loop Gain.

These features coupled with 120dB CMRR and PSRR make the HA-5135 an ideal device for precision DC instrumentation amplifiers. Excellent input characteristics in conjunction with 2.5MHz bandwidth and 0.8V/µs slew rate, make this amplifier extremely useful for precision integrator and biomedical amplifier designs. This amplifier is also well suited for precision data acquisition and for accurate threshold detector applications.

HA-5135 offers added features over the industry standard OP-07 in regards to bandwidth and slew rate specifications. For the military grade product, refer to the HA-5135/883 data sheet.

**Pinout**

**HA-5135 (CERDIP)  
TOP VIEW**



NOTE: Both BAL 1 pins are connected together internally.

**Features**

- Low Offset Voltage . . . . . 75µV (Max)
- Low Offset Voltage Drift . . . . . 0.4µV/°C
- Low Noise . . . . . 9nV/√Hz
- Open Loop Gain . . . . . 140dB
- Unity Gain Bandwidth . . . . . 2.5MHz
- All Bipolar Construction

**Applications**

- High Gain Instrumentation
- Precision Data Acquisition
- Precision Integrators
- Biomedical Amplifiers
- Precision Threshold Detectors

**Part Number Information**

PART NUMBER	TEMP. RANGE (°C)	PACKAGE	PKG. NO.
HA7-5135-5	0 to 75	8 Ld CERDIP	F8.3A

**Absolute Maximum Ratings**

Voltage Between V+ and V- Terminals ..... 40V  
 Differential Input Voltage ..... 7V  
 Output Short Circuit Duration ..... Indefinite

**Thermal Information**

Thermal Resistance (Typical, Note 2)  $\theta_{JA}$  (°C/W)  $\theta_{JC}$  (°C/W)  
 CERDIP Package ..... 115 28  
 Maximum Junction Temperature (Note 1) ..... 175°C  
 Maximum Storage Temperature Range ..... -65°C to 150°C  
 Maximum Lead Temperature (Soldering 10s) ..... 300°C

**Operating Conditions**

Temperature Ranges  
 HA-5135-5 ..... 0°C to 75°C

*CAUTION: Stresses above those listed in "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress only rating and operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied.*

**NOTES:**

1. Maximum power dissipation, including output load, must be designed to maintain the maximum junction temperature below 175°C.
2.  $\theta_{JA}$  is measured with the component mounted on an evaluation PC board in free air.

**Electrical Specifications**  $V_{SUPPLY} = \pm 15V$

PARAMETER	TEST CONDITIONS	TEMP. (°C)	HA-5135-5			UNITS
			MIN	TYP	MAX	
<b>INPUT CHARACTERISTICS</b>						
Offset Voltage		25	-	10	75	$\mu V$
		Full	-	50	130	$\mu V$
Average Offset Voltage Drift		Full	-	0.4	1.3	$\mu V/^\circ C$
Bias Current		25	-	$\pm 1$	$\pm 4$	nA
		Full	-	-	$\pm 6$	nA
Bias Current Average Drift		Full	-	0.02	0.04	nA/°C
Offset Current		25	-	-	4	nA
		Full	-	-	5.5	nA
Offset Current Average Drift		Full	-	0.02	0.04	nA/°C
Common Mode Range		Full	$\pm 12$	-	-	V
Differential Input Resistance		25	20	30	-	M $\Omega$
Input Noise Voltage (Note 3)	0.1Hz to 10Hz	25	-	-	0.6	$\mu V_{P-P}$
Input Noise Voltage Density (Note 3)	f = 10Hz	25	-	13.0	18.0	nV/ $\sqrt{Hz}$
	f = 100Hz		-	10.0	13.0	nV/ $\sqrt{Hz}$
	f = 1000Hz		-	9.0	11.0	nV/ $\sqrt{Hz}$
Input Noise Current (Note 3)	0.1Hz to 10Hz	25	-	15	30	pA $_{P-P}$
Input Noise Current Density (Note 3)	f = 10Hz	25	-	0.4	0.8	pA/ $\sqrt{Hz}$
	f = 100Hz		-	0.17	0.23	pA/ $\sqrt{Hz}$
	f = 1000Hz		-	0.14	0.17	pA/ $\sqrt{Hz}$
<b>TRANSFER CHARACTERISTICS</b>						
Large Signal Voltage Gain	$V_{OUT} = \pm 10V, R_L = 2k\Omega$	25	120	140	-	dB
		Full	120	-	-	dB
Common Mode Rejection Ratio	$V_{CM} = \pm 10V$	Full	106	120	-	dB
Closed Loop Bandwidth	$A_{VCL} = +1$	25	0.6	2.5	-	MHz
<b>OUTPUT CHARACTERISTICS</b>						
Output Voltage Swing	$R_L = 600\Omega$	25	$\pm 10$	$\pm 12$	-	V
		Full	$\pm 10$	-	-	V

**Electrical Specifications**  $V_{SUPPLY} = \pm 15V$  (Continued)

PARAMETER	TEST CONDITIONS	TEMP. (°C)	HA-5135-5			UNITS
			MIN	TYP	MAX	
Full Power Bandwidth (Note 4)	$R_L = 2k\Omega$	25	8	10	-	kHz
Output Current	$V_{OUT} = 10V$	25	$\pm 15$	$\pm 20$	-	mA
Output Resistance	Note 5	25	-	45	-	$\Omega$
<b>TRANSIENT RESPONSE</b> (Note 6)						
Rise Time		25	-	340	-	ns
Slew Rate		25	0.5	0.8	-	V/ $\mu$ s
Settling Time (Note 7)		25	-	11	-	$\mu$ s
<b>POWER SUPPLY CHARACTERISTICS</b>						
Supply Current		Full	-	1.0	1.7	mA
Power Supply Rejection Ratio	$V_S = \pm 5V$ to $\pm 20V$	Full	94	130	-	dB

NOTES:

3. Not tested. 90% of units meet or exceed these specifications.
4. Full power bandwidth guaranteed based on slew rate measurement using:  $FPBW = \frac{\text{Slew Rate}}{2\pi V_{PEAK}}$ .
5. Output resistance measured under open loop conditions ( $f = 100\text{Hz}$ ).
6. Refer to test circuits section of the data sheet.
7. Settling time is measured to 0.1% of final value for a 10V output step and  $A_V = -1$ .

**Test Circuits and Waveforms**

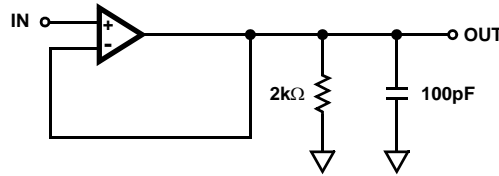
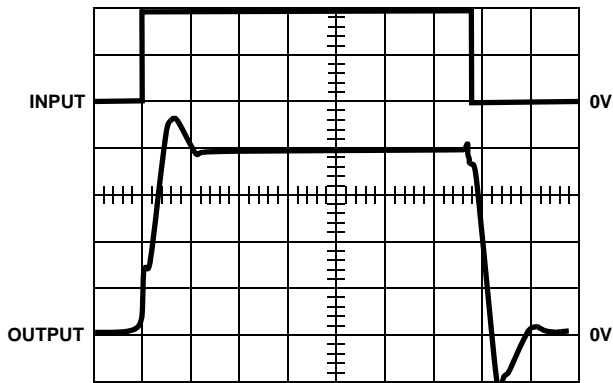
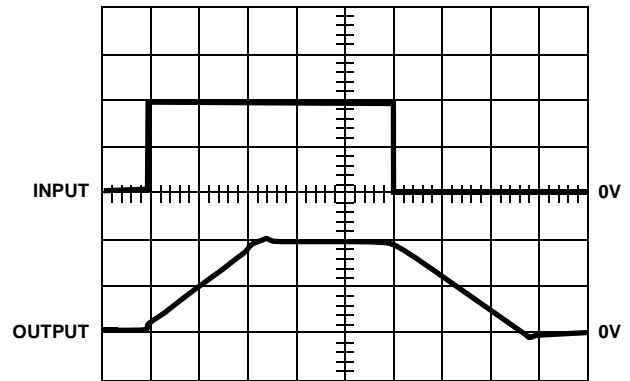


FIGURE 1. SLEW RATE AND TRANSIENT RESPONSE TEST CIRCUIT



Vertical Scale: Input = 50mV/Div. Output = 100mV/Div.  
Horizontal Scale: 1 $\mu$ s/Div.

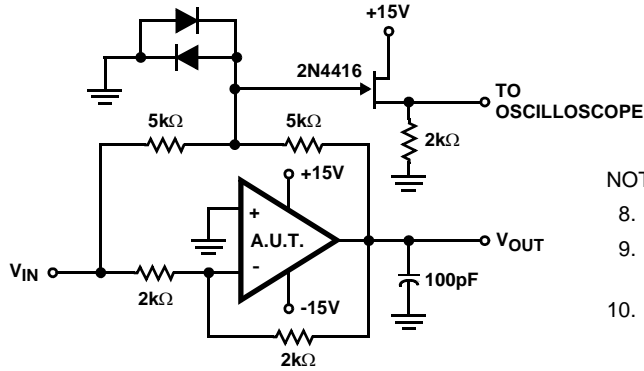
SMALL SIGNAL RESPONSE



Vertical Scale: 5V/Div.  
Horizontal Scale: 5 $\mu$ s/Div.

LARGE SIGNAL RESPONSE

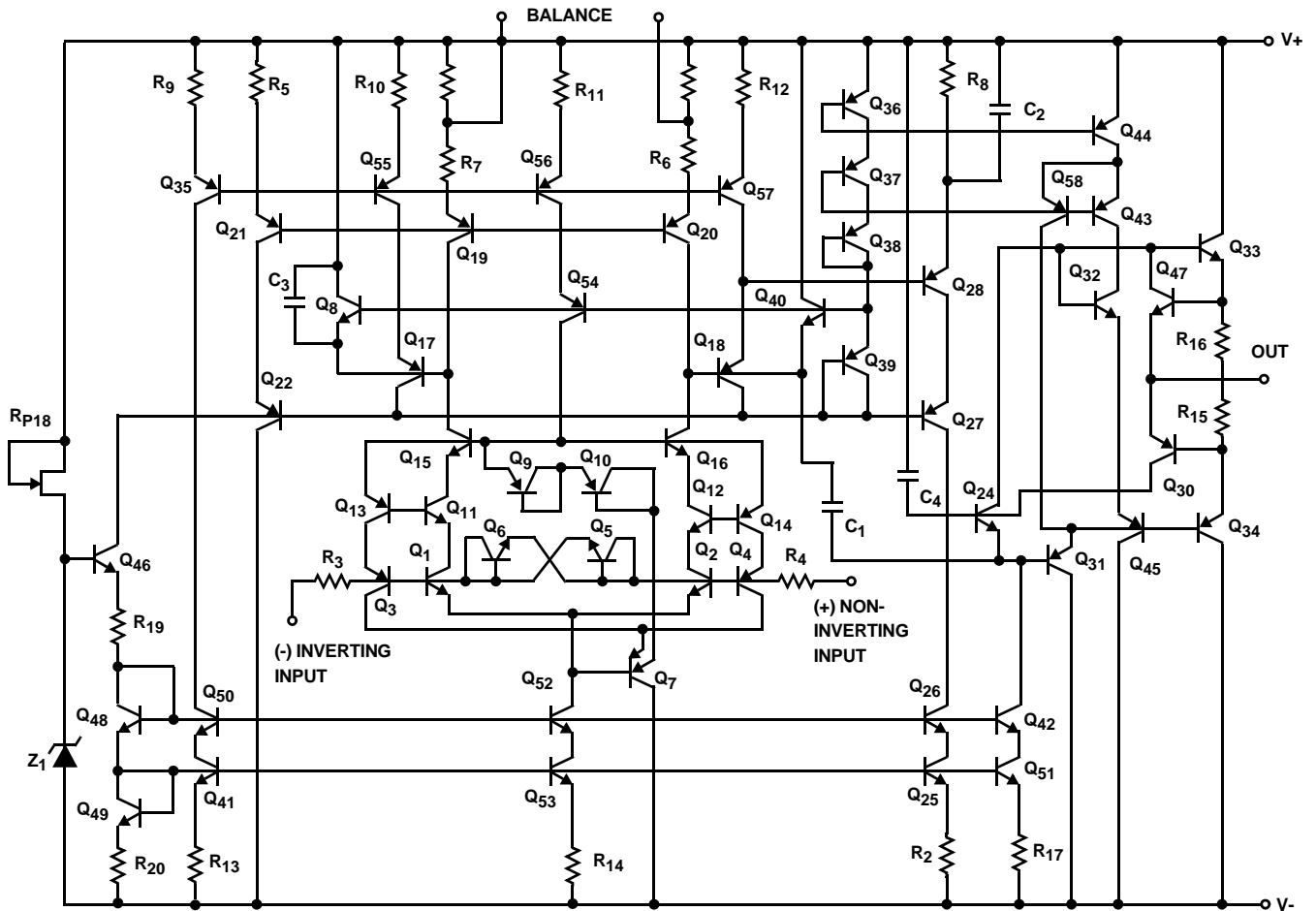
Test Circuits and Waveforms (Continued)



- NOTES:
8.  $A_V = -1$ .
  9. Feedback and summing resistors should be 0.1% matched.
  10. Clipping diodes are optional. HP5082-2810 recommended.

FIGURE 2. SETTLING TIME CIRCUIT

Schematic Diagram



**Application Information**

**Power Supply Decoupling**

Although not absolutely necessary, it is recommended that all power supply lines be decoupled with 0.01μF ceramic capacitors to ground. Decoupling capacitors should be located as near to the amplifier terminals as possible.

**Considerations For Prototyping:**

The following list of recommendations are suggested for prototyping.

1. Resolving low level signals requires minimizing leakage currents caused by external circuitry. Use of quality insulating materials, thorough cleaning of insulating surfaces and implementation of moisture barriers when required is suggested.
2. Error voltages generated by thermocouples formed between dissimilar metals in the presence of temperature gradients should be minimized. Isolation of low level circuitry from heat generating components is recommended.
3. Shielded cable input leads, guard rings and shield drivers are recommended for the most critical applications.

**Large Capacitive Loads**

When driving large capacitive loads (>500pF), a small value resistor (≈50Ω) should be connected in series with the output and inside the feedback loop.

**Offset Voltage Adjustment (See Figure 3)**

A 20kΩ balance potentiometer is recommended if offset nulling is required. However, other potentiometer values such as 10kΩ, 50kΩ and 100kΩ may be used. The minimum adjustment range for given values is ±2mV.  $V_{OS}$  TC of the amplifier is optimized at minimal  $V_{OS}$ . Tested Offset Adjustment is  $|V_{OS} + 1mV|$  minimum referred to output.

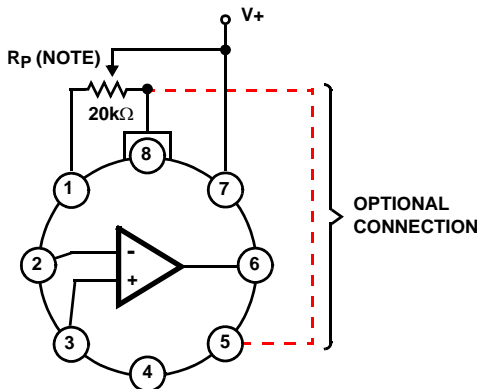


FIGURE 3. OFFSET NULLING CONNECTIONS

**Saturation Recovery**

Input and output saturation recovery time is negligible in most applications. However, care should be exercised to avoid exceeding the absolute maximum ratings of the device.

**Differential Input Voltages**

Inputs are shunted with back-to-back diodes for overvoltage protection. In applications where differential input voltages in excess of 1V are applied between the inputs, the use of limiting resistors at the inputs is recommended.

**Typical Applications**

The excellent input and gain characteristics of HA-5135 are well suited for precision integrator applications. Accurate integration over seven decades of frequency using HA-5135, virtually nullifies the need for more expensive chopper-type amplifiers.

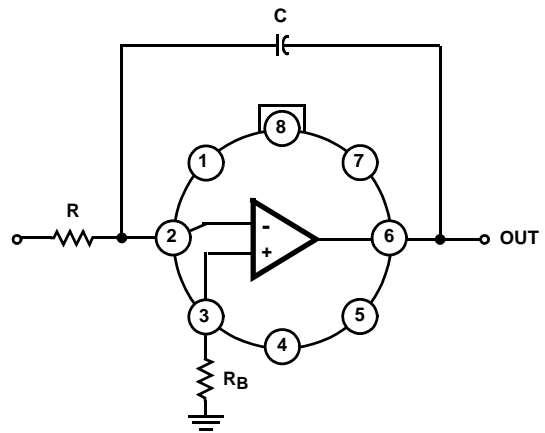


FIGURE 4. PRECISION INTEGRATOR

Low  $V_{OS}$  coupled with high open loop Gain, high CMRR and high PSRR make HA-5135 ideally suited for precision detector applications, such as the zero crossing detector shown in Figure 5.

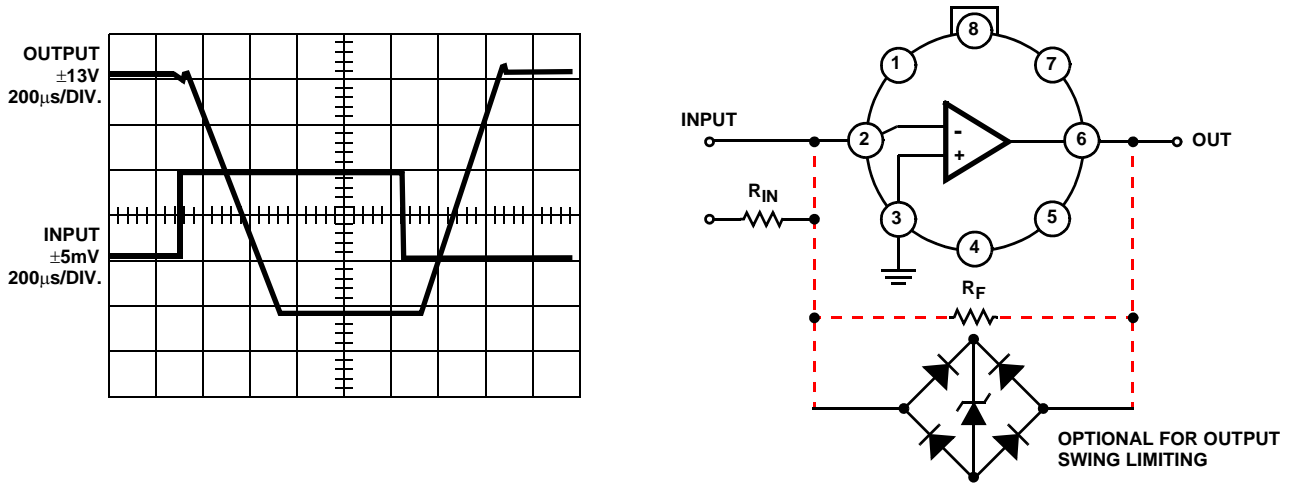


FIGURE 5. ZERO CROSSING DETECTOR

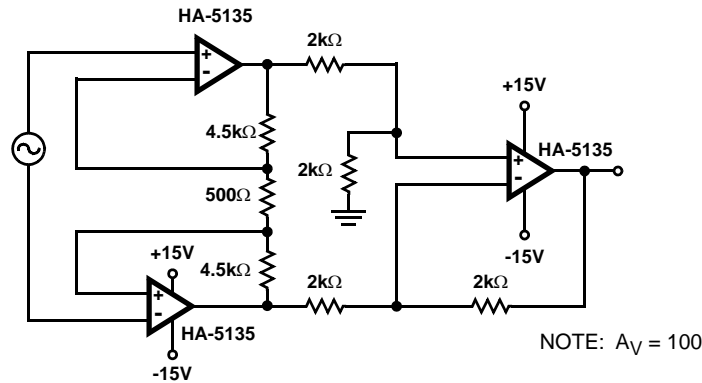


FIGURE 6. PRECISION INSTRUMENTATION AMPLIFIER

Typical Performance Curves

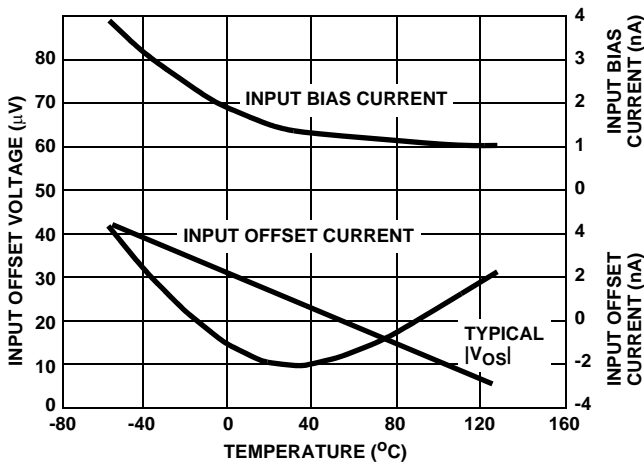


FIGURE 7. INPUT OFFSET VOLTAGE, INPUT BIAS AND OFFSET CURRENT vs TEMPERATURE

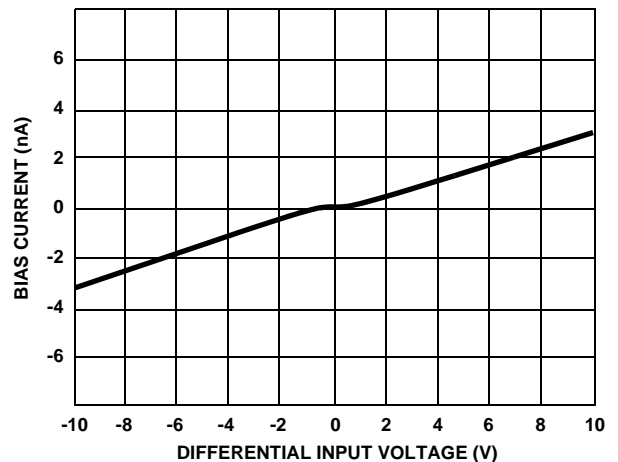


FIGURE 8. INPUT BIAS CURRENT vs DIFFERENTIAL INPUT VOLTAGE

Typical Performance Curves (Continued)

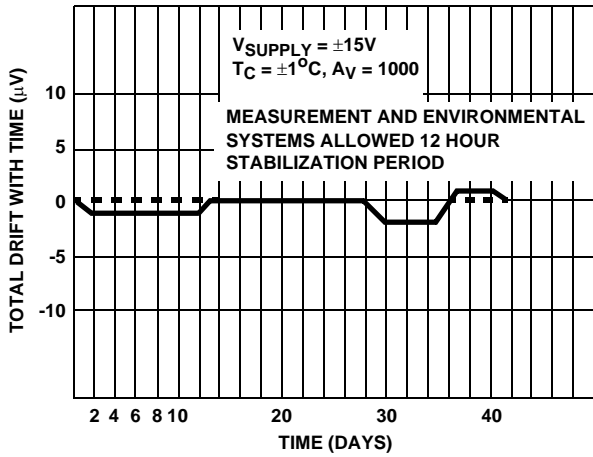


FIGURE 9. HA-5135 OFFSET VOLTAGE STABILITY vs TIME

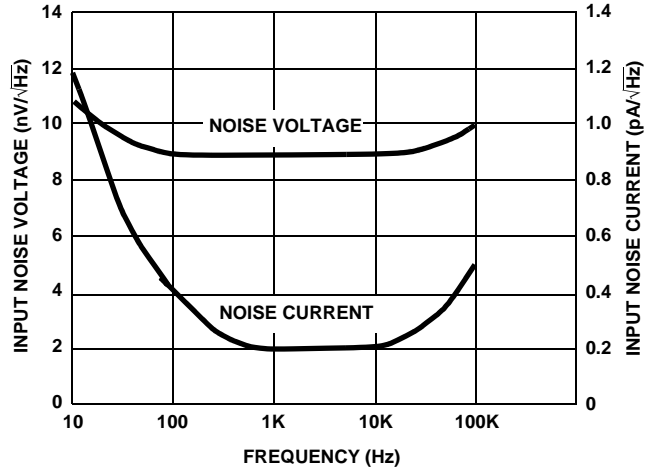


FIGURE 10. INPUT NOISE vs FREQUENCY

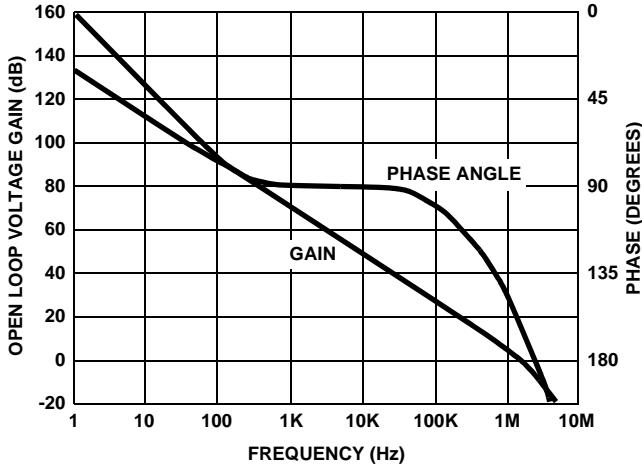


FIGURE 11. OPEN LOOP FREQUENCY RESPONSE

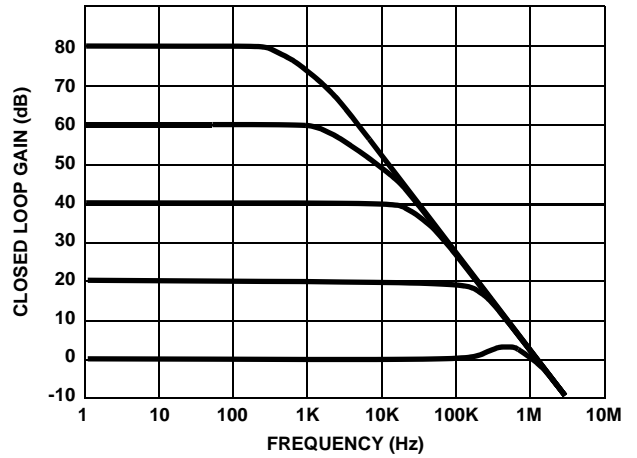


FIGURE 12. CLOSED LOOP FREQUENCY RESPONSE

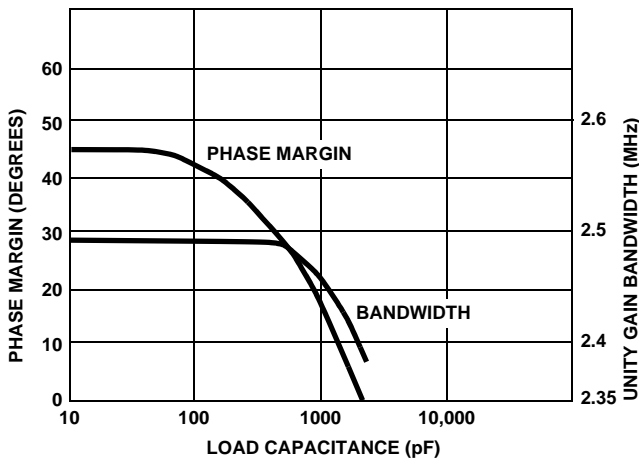


FIGURE 13. SMALL SIGNAL BANDWIDTH AND PHASE MARGIN vs LOAD CAPACITANCE

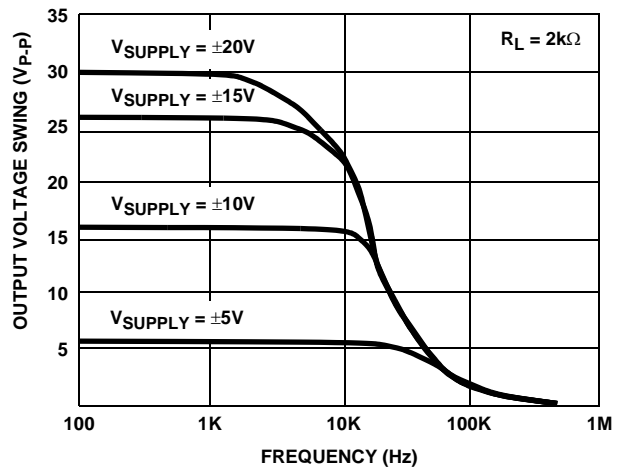


FIGURE 14. OUTPUT VOLTAGE SWING vs FREQUENCY

Typical Performance Curves (Continued)

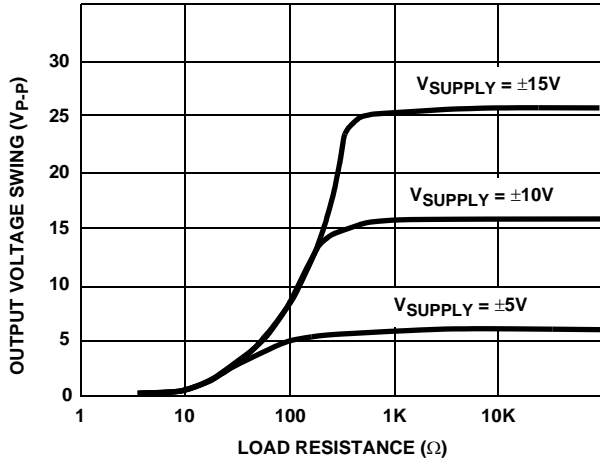


FIGURE 15. MAXIMUM OUTPUT VOLTAGE SWING vs LOAD RESISTANCE

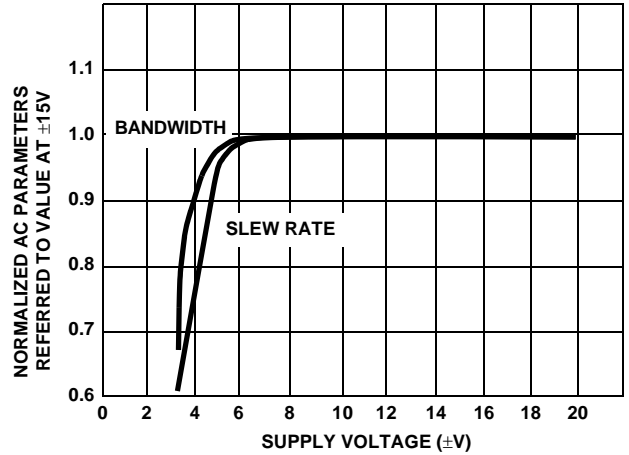


FIGURE 16. NORMALIZED AC PARAMETERS vs SUPPLY VOLTAGE

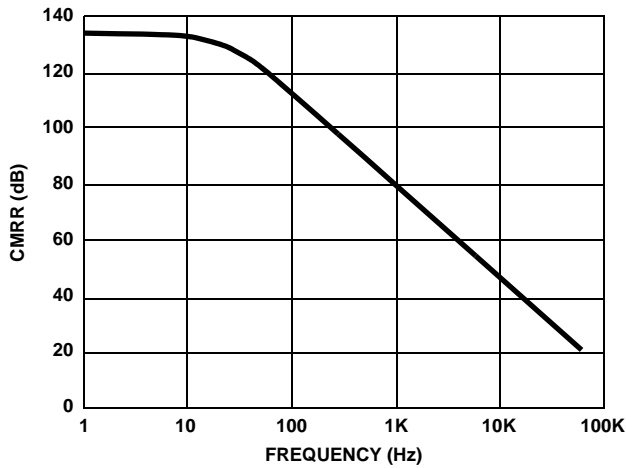


FIGURE 17. CMRR vs FREQUENCY

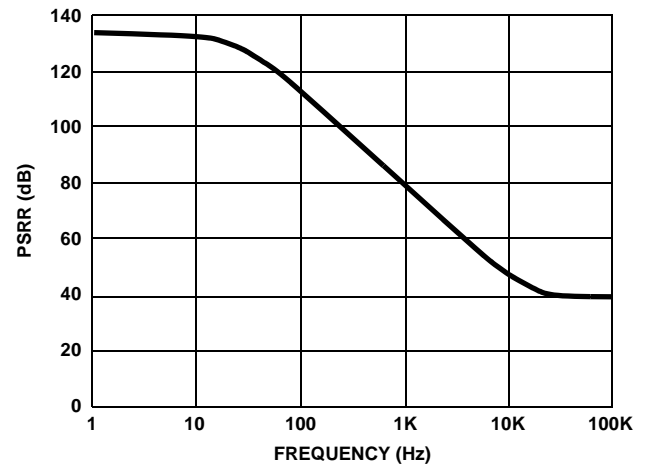


FIGURE 18. PSRR vs FREQUENCY

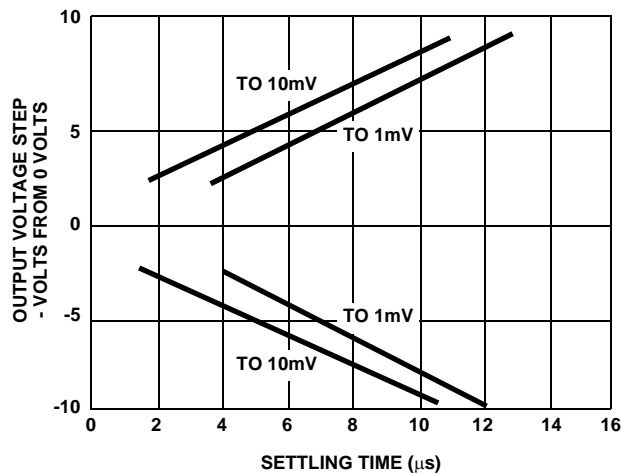


FIGURE 19. SETTLING TIME FOR VARIOUS OUTPUT STEP VOLTAGES

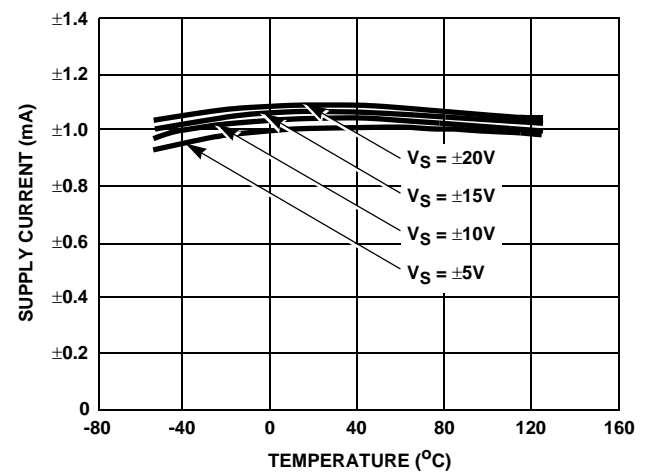


FIGURE 20. POWER SUPPLY CURRENT vs TEMPERATURE



**Die Characteristics**

**DIE DIMENSIONS:**

72 mils x 103 mils x 19 mils  
(1840µm x 2620µm x 483µm)

**METALLIZATION:**

Type: Al, 1% Cu  
Thickness: 16kÅ ±2kÅ

**SUBSTRATE POTENTIAL (POWERED UP):**

V-

**PASSIVATION:**

Type: Nitride (Si<sub>3</sub>N<sub>4</sub>) over Silox (SiO<sub>2</sub>, 5% Phos.)  
Silox Thickness: 12kÅ ±2kÅ  
Nitride Thickness: 3.5kÅ ±1.5kÅ

**TRANSISTOR COUNT:**

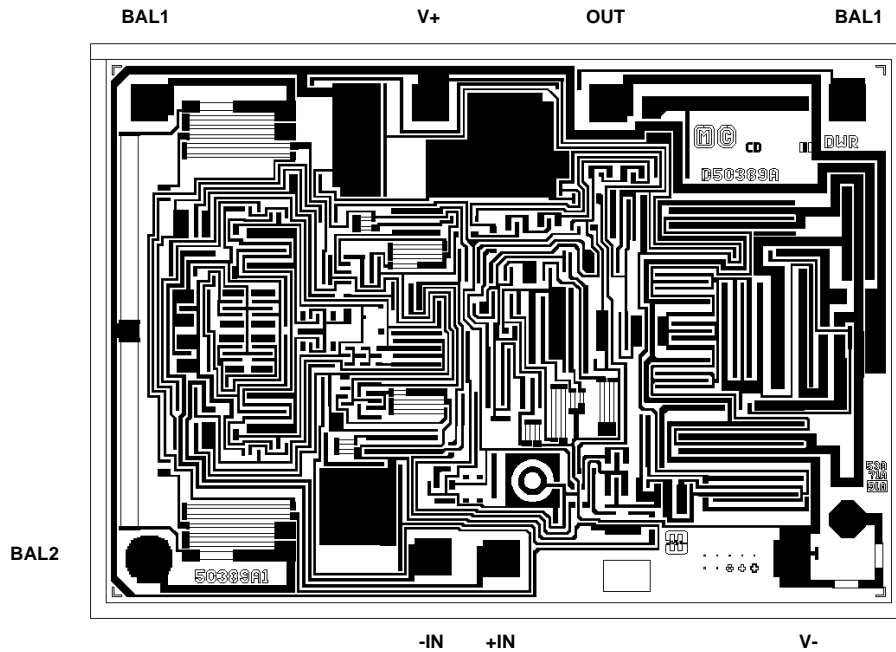
71

**PROCESS:**

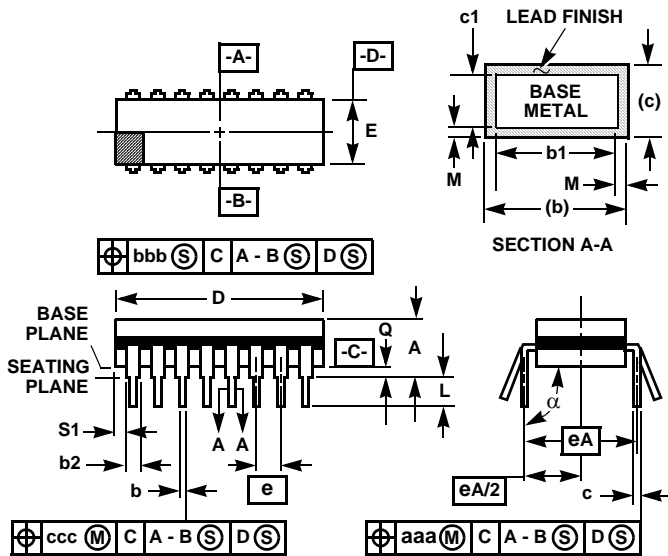
Bipolar Dielectric Isolation

**Metallization Mask Layout**

HA-51350



**Ceramic Dual-In-Line Frit Seal Packages (CERDIP)**



**F8.3A MIL-STD-1835 GDIP1-T8 (D-4, CONFIGURATION A)  
8 LEAD CERAMIC DUAL-IN-LINE FRIT SEAL PACKAGE**

SYMBOL	INCHES		MILLIMETERS		NOTES
	MIN	MAX	MIN	MAX	
A	-	0.200	-	5.08	-
b	0.014	0.026	0.36	0.66	2
b1	0.014	0.023	0.36	0.58	3
b2	0.045	0.065	1.14	1.65	-
b3	0.023	0.045	0.58	1.14	4
c	0.008	0.018	0.20	0.46	2
c1	0.008	0.015	0.20	0.38	3
D	-	0.405	-	10.29	5
E	0.220	0.310	5.59	7.87	5
e	0.100 BSC		2.54 BSC		-
eA	0.300 BSC		7.62 BSC		-
eA/2	0.150 BSC		3.81 BSC		-
L	0.125	0.200	3.18	5.08	-
Q	0.015	0.060	0.38	1.52	6
S1	0.005	-	0.13	-	7
$\alpha$	90°	105°	90°	105°	-
aaa	-	0.015	-	0.38	-
bbb	-	0.030	-	0.76	-
ccc	-	0.010	-	0.25	-
M	-	0.0015	-	0.038	2, 3
N	8		8		8

**NOTES:**

1. Index area: A notch or a pin one identification mark shall be located adjacent to pin one and shall be located within the shaded area shown. The manufacturer's identification shall not be used as a pin one identification mark.
2. The maximum limits of lead dimensions b and c or M shall be measured at the centroid of the finished lead surfaces, when solder dip or tin plate lead finish is applied.
3. Dimensions b1 and c1 apply to lead base metal only. Dimension M applies to lead plating and finish thickness.
4. Corner leads (1, N, N/2, and N/2+1) may be configured with a partial lead paddle. For this configuration dimension b3 replaces dimension b2.
5. This dimension allows for off-center lid, meniscus, and glass overrun.
6. Dimension Q shall be measured from the seating plane to the base plane.
7. Measure dimension S1 at all four corners.
8. N is the maximum number of terminal positions.
9. Dimensioning and tolerancing per ANSI Y14.5M - 1982.
10. Controlling dimension: INCH

Rev. 0 4/94

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