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Equator Hardware Reference

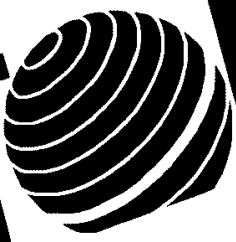
# MAP-CA DSP Datasheet

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**Equator Technologies, Inc.**

June 20, 2001

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# Equator Hardware Reference MAP-CA DSP Datasheet

June 20, 2001

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## Introduction to the MAP-CA Digital Signal Processor Chip

The MAP-CA digital signal processor offers a highly integrated single chip solution for broadband products such as set-top boxes, digital TVs, video conferencing systems, medical imaging products, digital video editing equipment, and office automation products. The MAP-CA DSP is a member of the MAP Series VLIW processors. A parallelizing C compiler, linker, source level debugger, simulators, and libraries are available. Reference software modules, including MPEG-2 encode and decode, JPEG encode and decode, video post-filtering, audio, telephony, and video teleconferencing codecs are also available to accelerate customer product development. Because core media applications can be delivered in software on the MAP-CA DSP platform, it is easy to add, remove or enhance the functions of final products. The MAP-CA DSP provides the proven and effective solution for rapidly evolving broadband applications.

### VLIW Core

- Highly pipelined Very Long Instruction Word processor that issues four operations per clock cycle
  - four 32-bit integer ALUs, two 64-bit shuffle/partitioned add units, and two 128-bit multimedia units
  - 128 32-bit general purpose registers, which can be treated as sixty-four 64-bit general purpose registers
  - thirty-two 1-bit predicate registers
  - eight special 128-bit registers
- 11+ GOPS sustained 16-bit SIMD operations @300 MHz
- 24+ GOPS sustained 8-bit SIMD operations @300 MHz
- 30+ GOPS @ 300 MHz for sum of absolute

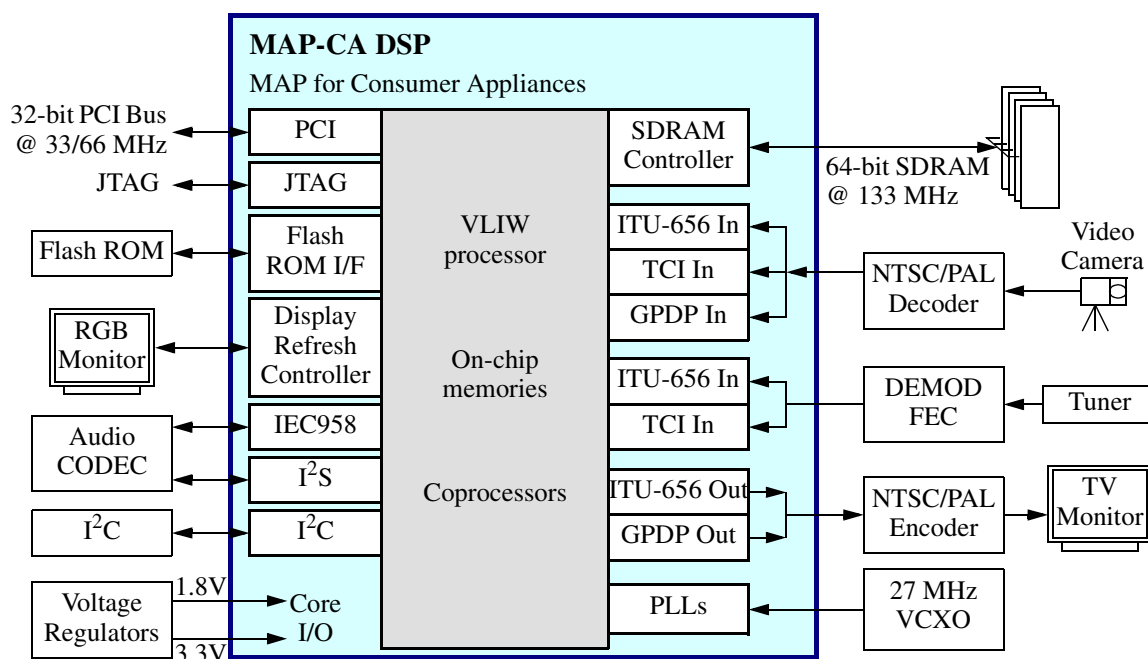
differences block matching

- 1800 MIPS @ 300 MHz in 32-bit integer arithmetic
- Bi-endian support

### Memory Hierarchy

- 32 KB two way set associative, LRU replacement policy, compressed format instruction cache
- 32 KB four way set associative, four bank interleaved, true LRU, write-back data cache
- Separate MMUs for instruction, data and DMA with fully associative sixteen entry TLB for each MMU
- Glueless high speed 133 MHz SDRAM/SGRAM interface, supporting up to 128 MB

### System Diagram



## Coprocessors

- Programmable VLx (16-bit RISC processor) with acceleration for variable length decoding and encoding with 4 KB data memory and 4 KB instruction memory
- 4 (vertical) × 5 (horizontal) / 3 × 5 / 2 × 5 tap Video Filter with 6KB line buffer memory
- Programmable DataStreamer (64 channel DMA engine) with 8 KB buffer memory
- DES support

## IO Interfaces

- 33 MHz/66 MHz 32-bit PCI bus
- IEC958 audio interfaces
- I<sup>2</sup>S audio interfaces
- Video input:
  - two DVB compliant transport channel interfaces
  - or
  - one DVB compliant transport channel interface and one ITU-R BT.601/656 input
  - or
  - two ITU-R BT.601/656 inputs
  - or
  - one of any of the above and one general data port
- One ITU-R BT.601/656 video output or one general data port
- Display Refresh Controller (DRC) with on-chip color space conversion, palette table lookup, alpha-blending, and hardware cursor
- 110 MHz RAMDAC with sync on green for analog RGB monitor
- I<sup>2</sup>C master/slave interface
- Flash ROM (EEPROM) interface

## Data Sheet Overview

This data sheet provides the following information:

- An overview of the MAP-CA DSP architecture
- A description of the software development platform
- A description of the hardware development platform
- Packaging information
- Electrical specifications

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# Chapter 1 Architecture Overview

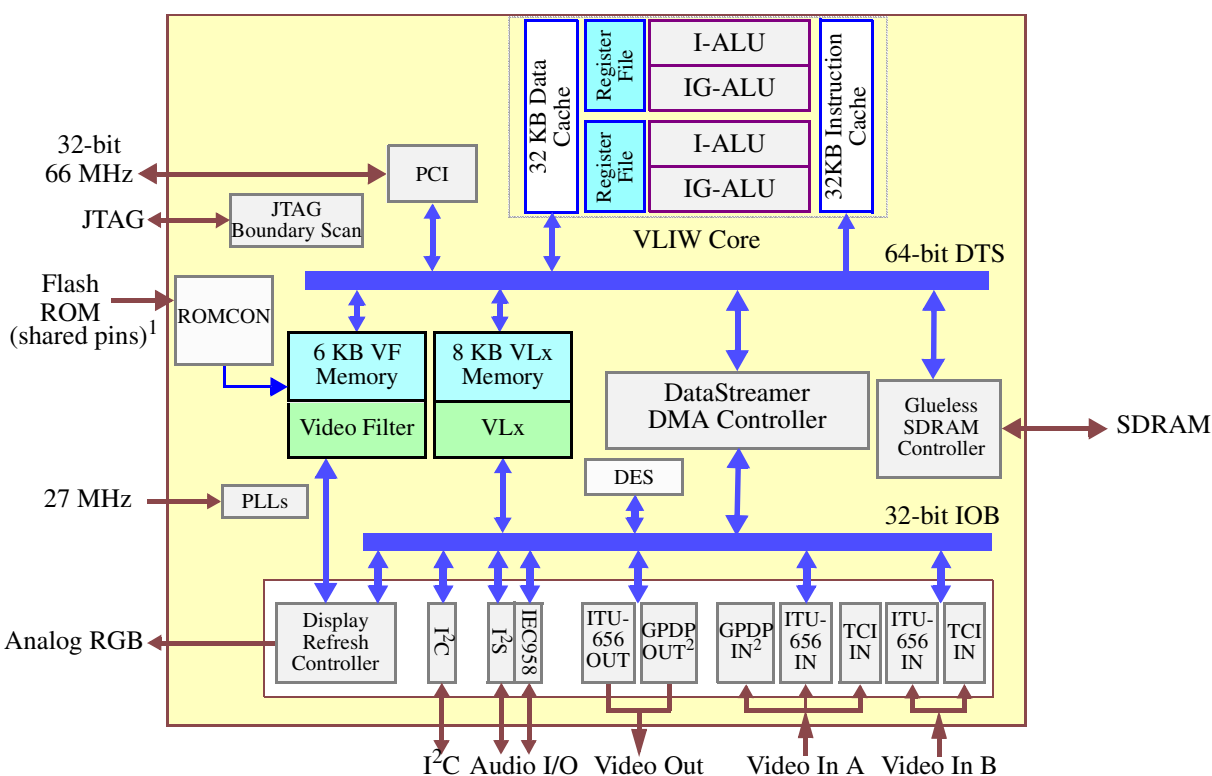


Figure 1-1: MAP-CA DSP Block Diagram

1. Some pins have multiple uses. See Chapter 3, Chapter 4, and Chapter 6 for more information
2. ITU-656 Out is unavailable for use when GDPD In is enabled due to sharing of interface signals.

The MAP-CA digital signal processor is a high performance processor providing broadband applications with solutions addressing the convergence of communications, consumer appliances and computing. The MAP-CA DSP combines general purpose RISC-like processing with high performance signal and image processing. The MAP-CA DSP supports programmable video, image, and signal processing software implementations of compression and decompression algorithms. The MAP-CA DSP matches the cost and performance features of dedicated fixed function chips, with the added flexibility to rapidly respond to evolving standards. Figure 1-1 shows a block diagram of the MAP-CA DSP. The MAP-CA DSP consists of a VLIW core,

programmable coprocessors, on-chip memories, and I/O interfaces.

The VLIW core executes four operations in parallel and supports partitioned SIMD operations for 8, 16, 32, and 64-bit data types. Coprocessors on the MAP-CA digital signal processor help accelerate serial operations like variable length encoding/decoding and video filtering.

Several audio/video interfaces are supported, including ITU-R BT.601/656 input and output; MPEG-2 transport channel interface (TCI); IEC958 and I<sup>2</sup>S digital audio interfaces. Two video inputs can be used at the same time. The Display Refresh Controller (DRC) supports RGB computer screen refresh and also has hardware support for overlay-

ing a hardware cursor and graphics/text or a secondary video channel on the primary video channel. An I<sup>2</sup>C bus interface is also provided.

These I/O functions execute in parallel with the CPU and eliminate the need for several external ASICs with their associated cost and bandwidth issues.

A glueless SDRAM controller supports access up to a 133 MHz SDRAM. The MAP-CA digital signal processor supports a 128 MB memory size. A 32-bit 33/66 MHz PCI bus interface is also supported. The MAP-CA DSP boots from either the PCI bus or the Flash ROM interface.

There are three on-chip PLLs (core/SDRAM, pixel, audio) that generate all the internal clocks from a single 27 MHz external clock input (`pc1k`). The `tc1_vdac` pin can be used to output a controlling signal that can be used to drive a one bit sigma-delta modulator for an external VCXO which in turns modulates the frequency on `pc1k`.

## 1.1 The VLIW Core

Real-time handling of multimedia data stresses processor performance. There are three basic ways to increase a processor's performance: decrease the cycle time, decrease the number of cycles required to execute an instruction, and execute more instructions per cycle. The first two are becoming increasingly difficult to improve beyond process scheduling, while the last is now receiving more attention. Executing more instructions per cycle exploits the natural parallelism available in most software. Very Long Instruction Word (VLIW) processors use this parallelism by packing multiple operations into a single instruction word, which is then executed as a unit.

VLIW architectures differ from superscalar architectures in that the grouping and scheduling of instructions for execution is done at compile time, rather than execution time. The iMMediaC compiler searches for eligible operations, checks for dependencies and resource conflicts, and packages these eligible operations into VLIWs. The compiler can explore beyond the limited search window seen in superscalar architectures and cross natural boundaries, such as branches, to search for opportunities for parallelism. The iMMediaC compiler uses

a technique known as "trace scheduling" to search a whole routine for eligible operations.

By moving the difficult task of finding parallelism into software, VLIW techniques dramatically simplify the CPU design by reducing gate count and freeing valuable die area for other performance enhancements or lower costs. While VLIW is primarily designed to exploit parallelism, its simplification of the processor architecture allows for reduced cycle times as well.

### 1.1.1 Execution Units

The MAP-CA DSP operations are primarily 3-operand RISC operations. As in a typical RISC architecture, load and store operations are the only means of referencing memory. The MAP-CA DSP has four functional units: two I-ALUs, and two IG-ALUs. Each I-ALU contains a load-store unit, an integer ALU, and a branch unit. Each IG-ALU contains an integer/graphics unit and a multimedia operation unit. The I-ALU and IG-ALU support different operations, but many integer and logical operations are implemented in both units. This overlap allows the compiler to schedule more operations in parallel and make more efficient use of all the functional units.

There are 128 32-bit registers usable separately or in pairs as 64-bit registers, 32 1-bit predicate registers, and eight special 128-bit registers. The 128-bit (PLC/PLV) registers are used for FIR filter, SAD, FFT, ADD, DCT, and other specialized partitioned integer operations. The large register files help minimize unnecessary instruction dependencies caused by logically distinct register reuses.

Each MAP-CA digital signal processor instruction contains four operations. The Media Intrinsic operations include partitioned operations over these data types. Load and store operations can perform one, two, four, and eight byte accesses, with support for both little-endian and big-endian byte orderings. Dynamic address translation and virtual memory protection are fully supported. The 1-bit logical values are also used to support predicated execution, which substantially enhances available parallelism by allowing partial speculation and eliminating branching.



### 1.1.1.1 I-ALU

The I-ALU performs the following operations:

- 32-bit integer arithmetic operations including compare
- Logical and bitwise logical operations whose results can be sent to general or predicate registers
- Address calculations for indexed addressing
- Memory reference
- Branching
- System control operations

### 1.1.1.2 IG-ALU

The IG-ALU performs the following operations:

- 32-bit integer arithmetic operations (same as the I-ALU)
- Logical and bitwise logical operations (same as the I-ALU)
- 64-bit integer arithmetic operations
- Shift/extract/merge operations
- 64-bit SIMD operations (with 8-bit, 16-bit, and 32-bit partitions) including selection, comparison, selection of maximums and minimums, addition, multiply-add, complex multiplication, inner product, and sum of absolute differences
- 128-bit partitioned (with 8-bit, 16-bit, and 32-bit partitions) SIMD operations including inner-product with new partition shift-in for efficient FIR operation and sum of absolute differences with new partition shift-in for efficient block matching operation

### 1.1.1.3 Simple Interlocks

Certain operations require more than one cycle to complete. No hardware interlocks are needed to prevent issue of an operation that attempts to read a result not yet completed. The iMMediaC compiler is responsible for correct scheduling, not hardware. Register scoreboarding is supported for outstanding loads.

### 1.1.1.4 Extensive Predication

Nearly all operations can have their effect controlled by the value of a selected (1-bit) predicate register. A predicate register is tested to determine whether or not the operation should be performed. This allows the compiler to aggressively convert

control flow into data flow, enabling a substantially higher degree of instruction-level parallelism. This also greatly helps to reduce any penalties for branching, without the cost and complexity of hardware branch prediction.

## 1.1.2 Register Resources

There are several types of registers on the MAP-CA digital signal processor. These include system registers, breakpoint registers, general purpose registers, predicate registers, and special purpose 128-bit registers.

### 1.1.2.1 Global Registers

Global registers on the MAP-CA DSP consist of system registers and implementation-dependent I/O registers (PIO registers). Dedicated operations manipulate the system registers; conventional load and store operations manipulate the I/O registers.

### 1.1.2.2 Breakpoint Registers

MAP-CA DSP has two sets of breakpoint registers: instruction-breakpoint and data-breakpoint registers. These registers provide hardware breakpoint capability for various debugging tools. Instruction-breakpoint registers cause an exception when an operation in the specified address is about to be executed. Similarly, the data-breakpoint registers cause an exception when the data at the specified address is about to be accessed. In both cases, a mask can be used to specify a range of addresses.

By registering an exception handling routine associated with either of these exceptions, a software developer can control what happens when a hardware breakpoint occurs. For example, the exception handling routine may be used to signal an external application such as a source-level debugger that a breakpoint has occurred.

### 1.1.2.3 General Registers

There are 128 32-bit registers that can be treated as 64-bit general registers using even-odd pairs of the 32-bit registers.

### 1.1.2.4 Predicate Registers

There are 32 1-bit predicate registers. Predicate registers are used in predicated operations, logical operations, and branches. They provide a destination for operations with a judged condition.

### 1.1.2.5 PLC/PLV 128-bit registers

The IG-ALU has eight special 128-bit registers – two pairs of Partitioned Local Constant (PLC) registers and two pairs of Partitioned Local Variable (PLV) registers. These registers are used for powerful SIMD digital signal processor partitioned operations. The registers can be configured as sixteen 8-bit operation partitions, eight 16-bit operation partitions, or four 32-bit operation partitions. For numerous digital signal processing and compression algorithms, this allows MAP-CA DSP to match the cost/performance of fixed-function chips without the loss of re-programmability.

## 1.2 Interrupts and Exceptions

The MAP-CA DSP has a flexible interrupt structure. Interrupts and exceptions internal to the core are reflected directly in system registers. All other interrupts from on-chip devices and PCI interrupts from external devices are gathered by an on-chip interrupt controller. The interrupt controller also provides a number of software interrupts.

Routing, masking, and prioritization of interrupts is completely software programmable. Each of the interrupts handled by the interrupt controller can be individually masked, or routed to one of four core interrupts or to one of two PCI interrupt signals.

### 1.2.1 Core Interrupts and Exceptions

Table 1-1 and Table 1-2 list the events that can trigger interrupts or exceptions within the core. When an event occurs, a bit is set in an “Event Seen” system register. If the event is not masked (or not maskable), the address for a handler will be fetched from one of nine Event Vector system registers, depending on the event.

Table 1-1: System Events

Name	Event	Maskable
IO0..IO3	I/O interrupts (from interrupt controller)	Yes
SINT0..SINT1	Software interrupts	Yes
FCNT	Free running counter overflow	Yes
INTV0..INTV1	Interval timers	Yes
ILPC	Illegal program counter	No
IBPT	Instruction address break	Yes
BPOP	Breakpoint operation	No
SYS	System call (trap instruction)	No
ITLBAA	ITLB application access	No
ITLBR	ITLB reference	No
ITLBM	ITLB miss	No

Table 1-2: Operation Events

Name	Event	Maskable
ILLO	Illegal operation	No
PLV	Privilege violation	No
DBPT	Data address break	Yes
DALN	Data alignment error	No
DTLBKW	DTLB kernel write	No
DTLBAW	DTLB application write	No
DTLBAA	DTLB application access	No
DTLBR	DTLB reference	No
DTLBM	DTLB miss	No

### 1.2.2 Interrupt Controller

The MAP-CA digital signal processor interrupt controller supports multiple maskable interrupts from outside of the core. Non-core interrupt sources include on-chip devices such as TCI, the DRC, the

DataStreamer DMA Controller, and PCI interrupts from external devices or hosts. Software-generated shoulder-tap interrupts are provided for multiprocessing or inter-process communication support.

MAP-CA DSP interrupts can be examined and controlled via PIO registers in the ROMCON control block. Routing and masking of interrupts is programmable. Each interrupt can be individually masked or routed to one of four core interrupts or to one of two PCI interrupt signals. Software interrupts can be similarly masked and routed. They may be asserted or de-asserted under software control.

Table 1-3 shows the supported interrupts.

Table 1-3: Supported Interrupts

Name	Interrupt
IrqAlwaysOne	debug interrupt, always asserted
IrqIIC	I <sup>2</sup> C
IrqTCIO	primary TCI
IrqDRC	display refresh controller
IrqNTSCIn0	primary ITU-R BT.601/656 in
IrqNTSCIn1	secondary ITU-R BT.601/656 in
IrqTCII	secondary TCI
IrqPCIAA	PCI interrupt pin A
IrqPCIAB	PCI interrupt pin B
IrqNTSCOut	ITU-R BT.601/656 output
IrqIEC958	IEC958 audio
IrqIIS	I <sup>2</sup> S audio
IrqPCIAPME	PCIA power management event
IrqDS0	DataStreamer interrupt 0
IrqDS1	DataStreamer interrupt 1
IrqDSTLB	DataStreamer TLB miss
IrqDSBufOvrFlow	DataStreamer I/O input overflow
IrqSoftWare	Software-controlled interrupts

## 1.3 Timers

The MAP-CA digital signal processor has two independent programmable interval timers plus a free-running counter. Each interval timer has a 32-bit counter register and period register. The counter is incremented once per cycle. When the counter reaches the period value, the counter is reloaded, a bit is set in the system Event Seen Register (ESR), and a maskable interrupt is asserted. The free-running counter counts up once per cycle

as well. When it overflows to zero, a bit is set in ESR and a maskable interrupt is asserted.

The transport channel interface also has a programmable timer that counts at a rate of 27 MHz and can be used to generate an interrupt upon rollover.

## 1.4 Memory Hierarchy

The MAP-CA DSP supports several on-chip memories and access to SDRAM and other memories via the PCI bus. The VLIW is equipped with a 32 KB instruction cache and 32 KB data cache used for caching instructions and data from SDRAM. In addition to supporting I-ALU ports, the data cache supports a port to the DTS (Data Transfer Switch), which makes data in the data cache available to the DataStreamer controller.

A 4 KB instruction memory and a 4 KB data memory are used by the VLx coprocessor. The Video Filter uses a 6 KB line buffer memory. These memories, totaling 14 KB, are also accessible by the VLIW core through un-cached load/store operations. In addition, these memories are also available to the DataStreamer DMA controller and for external use via PCI. The line buffer memory is used to store the content of Flash ROM at system boot up.

### 1.4.1 Caches

The MAP-CA DSP has a 32 KB instruction cache and a separate, multi-bank 32 KB data cache. Both caches are physically addressed, so that problems of aliasing and context switching do not arise. For fast address translation, the cache index is virtual but the tags are physical.

The instruction cache holds instructions in a compressed form. It is organized as a two-way set associative cache with a LRU replacement algorithm.

The data cache is a 32 KB, four-way set-associative (with true LRU replacement), write-back cache. The data cache supports four simultaneous 64-bit data accesses per cycle. The cache is non-blocking; up to 8 outstanding misses to different cache lines and up to 48 outstanding misses overall are allowed.

### 1.4.2 Address Translation

The MAP-CA DSP provides memory management support in the form of separate TLBs for the instruction stream, each I-ALU data access, and the DataStreamer DMA controller. The four TLBs

(ITLB, two DTLBs, and DSTLB) can be programmed independently.

The DTS-ID is part of the virtual address and can be used to direct accesses when the TLBs are disabled.

Each TLB has sixteen fully-associative entries. Each entry contains a Virtual Page Number (VPN), an 8-bit Address Space Identifier (ASID), access protection bits, and page size information. Each entry can map a page of any valid size, where the valid sizes are 16 KB, 64 KB, 256 KB, 1 MB, 4 MB, 16 MB, 64 MB, 256 MB, and 1 GB.

When a TLB miss occurs, an exception is generated. The exception handler can modify a TLB entry and retry the failed operation. Separate exception handlers can be installed for data, instruction, and DataStreamer controller TLB misses.

## 1.5 Databases and Controllers

The various buses and controllers on the MAP-CA digital signal processor are described in the following sections.

### 1.5.1 Memory Interface Controller

The Memory Controller Unit allows customers to easily build high-performance, external memory up to 128 MB using SDRAM/SGRAM without any external glue logic. Local memory supports externally initiated PCI accesses through the Address Translation Unit within the PCI module.

The Memory Controller Unit also includes hardware that queues, prioritizes, and transfers data from memory to memory or from memory to cache asynchronously to the initiating software.

The on-chip core PLL generates the clock for the memory controller and provides clock synchronization between the MAP-CA DSP and external SDRAM. This provides support for various combinations of CPU core and memory speeds.

### 1.5.2 Data Transfer Switch (DTS)

The DTS is a split-transaction bus. The DTS contains the data and address buses, a high speed bridging system, and a bus arbiter. The bridge, arbiter, and bus arrangement is a very high-speed communication solution that allows multiple media applications to be executed concurrently.

The arbiter can handle multiple requestors using priority based scheduling.

### 1.5.3 DataStreamer DMA Controller

The DataStreamer DMA controller is a high performance, programmable DMA engine that performs buffered data transfer between different MAP-CA DSP memory subsystems or between memories and I/O devices. The DataStreamer controller is programmed and controlled by software. The DataStreamer controller then performs the requested transfer without further intervention from the core.

The DataStreamer controller can perform the following classes of transfers:

- memory-to-memory: perform block transfers, preload data into the cache, fill a memory region with 0 or 1 bits
- memory-to-I/O and I/O-to-memory: perform I/O transfers

The DataStreamer DMA controller features include:

- an 8 KB internal memory that can be partitioned into as many as 64 variable-sized buffers. Each buffer is simultaneously the sink for an input I/O or memory channel and the source for an output I/O or memory channel.
- sixty-four independent programmable channels for transfers between various memories and the DataStreamer controller's internal buffer,
- Channel programs, called Descriptor Lists, allow transfers of arbitrary or infinite length to be specified. Regular and irregular patterns of contiguous or non-contiguous transfers are easy to specify.
- Memories that can be read or written include SDRAM, on-chip memories, and PCI bus accessible memories. Cache preloading can also be performed.
- Interrupts can be triggered by descriptors, allowing end-of-transfer or mid-stream interrupts to be generated.

### 1.5.4 PCI Bus

The PCI unit implements a 32-bit PCI 2.1 interface with speed up to 66 MHz. The PCI interface is a single function device with two BARs. Certain fields in the configuration registers may be initialized on power-up through ROM control. As a PCI target, the PCI interface allows access to the MAP-CA digital signal processor SDRAM (coherently or non-coherently with respect to the data

cache). The PCI interface also allows access to several programmer-visible control registers, PIO space and SDRAM. As a PCI master, the PCI interface allows the VLIW core, the DataStreamer DMA controller, and coprocessors to initiate PCI bus requests. The PCI unit can initiate memory, I/O and configuration commands on the PCI bus.

The MAP-CA digital signal processor can act as a host on the PCI bus. There are three pairs of request/grant lines for other devices on a PCI bus. This enables a multi-processor configuration to connect up to four MAP-CAs together on a PCI bus without a bridge.

The PCI interface implements two separate interrupt lines. If the MAP-CA DSP is not the host, any internal interrupt can be routed to any of these PC interrupts. If the MAP-CA DSP is the host, the PCI interrupts are sampled by the MAP-CA DSP and can be routed to the MAP-CA DSP VLIW core.

The MAP-CA DSP is a 3.3V-only I/O device. If the MAP-CA DSP is used in a system with a 5V PCI bus architecture, then a 5V-to-3.3V level translator is required.

### 1.5.5 I/O Bus

All on-chip peripheral devices are connected via the internal I/O Bus (IOB). This is a 32-bit internal bus running at one half of the VLIW core frequency. The IOB connects to the DTS through the DataStreamer controller. The IOB can handle real-time requests.

## 1.6 Coprocessors

Coproprocessors on the MAP-CA DSP help off-load “serial” tasks from the VLIW core or accelerate special purpose processing for video operations. The coprocessors operate in parallel with the VLIW core resulting in improved video processing.

### 1.6.1 VLx

The Variable Length Encoder/Decoder (Figure 1-2) or VLx is a 16-bit RISC coprocessor with thirty-two 16-bit registers. The VLx off-loads the bit sequential tasks of variable length encoding and variable length decoding (VLE/VLD) from the VLIW CPU core and accelerates applications such as JPEG, MPEG, H.263, JBIG, and DV. The VLx includes special purpose hardware for bitstream processing, hardware-accelerated MPEG-2 table lookup, and general purpose variable length decoding.

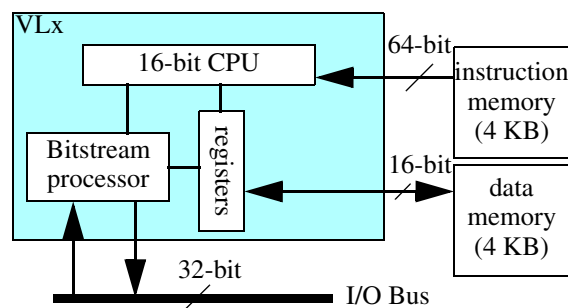


Figure 1-2: The VLx Coprocessor

### 1.6.2 Video Filter

A polyphase (8 phase) 2D Video Filter takes 4:2:0 or 4:2:2 YUV stream as input and scales either up or down as required. 4 (vertical)  $\times$  5 (horizontal) filters support up to 768 horizontal pixels, 3  $\times$  5 up to 1024 horizontal pixels, and 2  $\times$  5 up to 1536 horizontal pixels. The Video Filter pumps out scaled 4:4:4 YUV data to the DRC through the video bus. The Video Filter can also pump out 4:4:4 YUV data to the SDRAM for debug purposes. Its features are described below.

- Supports 8-bit coefficients.
- Supports both interspersed and co-sited pixel positioning.
- Supports vertical 4-tap polyphase (8-phase) filters for luminance and chrominance.
- Supports horizontal 5-tap polyphase (8-phase) filters for luminance and chrominance.
- Can scale up to a maximum resolution of 2047 $\times$ 2047 (depends upon memory bandwidth available for the video scaling operation).
- Can scale up from a minimum resolution of 17 $\times$ 4.
- The maximum scale down ratio is 1:7.

### 1.6.3 DES Module

The MAP-CA digital signal processor includes hardware support for encryption and decryption of data according to the *National Bureau of Standards Data Encryption Standard* and certain implementations thereof as defined in *FIPS Publications 46-2, 46-3, 74, and 81* and *ANSI Publication X9.52-1998*. For more information on this support, contact your Equator Technologies or Hitachi Sales Representative.

## 1.7 I/O Interfaces

### 1.7.1 Audio Interfaces

#### 1.7.1.1 IEC958 Audio Interface

This interface supports several audio standards:

- Sony/Philips Digital Interface (S/PDIF)
- Audio Engineering Society/European Broadcast Union (AES/EBU) interface
- TOSLINK interface (requires external IR devices)

The MAP-CA DSP IEC958 interface can insert even or odd parity on each sub-frame of the output bit stream.

#### 1.7.1.2 I<sup>2</sup>S Interface

The Inter-IC Sound (I<sup>2</sup>S) interface drives high quality audio D/A converters for home theater. The MAP-CA digital signal processor interface meets the requirements of the standard serial data protocol and provides connection for up to three stereo DACs and one ADC. The interface supports 48 kHz, 44.1 kHz, and 32 kHz audio sample rates. Simultaneous input and output must be at the same sample rate. The MAP-CA DSP IIS supports both master and slave mode interface. In slave mode there is the choice of using either external inputs or internally generated signals for the sample rate clock and serial bit clock.

### 1.7.2 Video Interfaces

The MAP-CA DSP provides two video input ports and one video output port. Each input port supports either transport channel interface input or ITU-R BT.601/656 input. The output port supports an ITU-R BT.601/656 compliant output.

In addition, the primary video input port and/or the output port can be used in a general purpose mode for transferring data (general purpose data port) for input or output respectively.

#### 1.7.2.1 Transport Channel Interfaces

The video input unit implements two DVB compliant transport channel interfaces which receive demodulated channel data in transport layer format. The transport channel interface (TCI) accepts MPEG-2 system transport packets in either byte parallel or, by default, bit serial form. Data rates up to 80Mbps (serial) or 30 MB/s (byte-wide parallel) are supported. By default, serial data is input on `tci_data[0]` and parallel data is input on

`tci_data[7:0]` with bit 7 the most significant. These orientations can be reversed by PIO programming.

The TCI synchronizes packet data received in broadcast applications such as satellite or cable. The TCI can detect inline sync bytes, which are the first byte of every transport header. Alternatively, the TCI can utilize the external `tci_sync` signal. Once byte-sync has been detected, the TCI moves byte-aligned data into the MAP-CA DSP's memory using the DataStreamer DMA controller.

The number of bytes in each packet is programmable. At the end of every packet, the TCI appends an eight-byte postscript that includes time stamps from the local clock counters. This information can be used in conjunction with the Program Clock References embedded in the transport stream to track the timing reference. This is accomplished by using a software loop filter to implement a 1-bit sigma-delta modulator to provide a controlling voltage for the external VCXO that drives `pclk`. The sigma-delta data stream is output at 1.5 MHz on the `tci_vdac` pin via the primary TCI.

In addition to the local clock counters, there is a programmable 27 MHz timer in each TCI module that generates an interrupt on overflow (rollover).

#### 1.7.2.2 ITU-656 Input Interface

This interface provides direct connection to an ITU-R BT.601/656 format NTSC/PAL video input decoder. The external decoder can be controlled using the I<sup>2</sup>C Serial Bus.

#### 1.7.2.3 ITU-656 Output Interface

A glueless interface to a NTSC/PAL video encoder is provided, enabling the MAP-CA digital signal processor to directly generate high-quality NTSC or PAL video-output signals. This interface supports 8-bit 525 and 625 line resolutions with either separate H/Vsync (ITU-R BT. 601) or inline sync (ITU-R BT.656). Advanced video post-filtering on the MAP-CA DSP processor via software can produce flicker-free output when converting interlace-to-progressive output. The external NTSC/PAL encoder can be controlled using the I<sup>2</sup>C Serial Bus.

#### 1.7.2.4 General Purpose Data Port (GPDP)

The general purpose data port provides an 8-bit parallel input/output port. Together with a clock and a

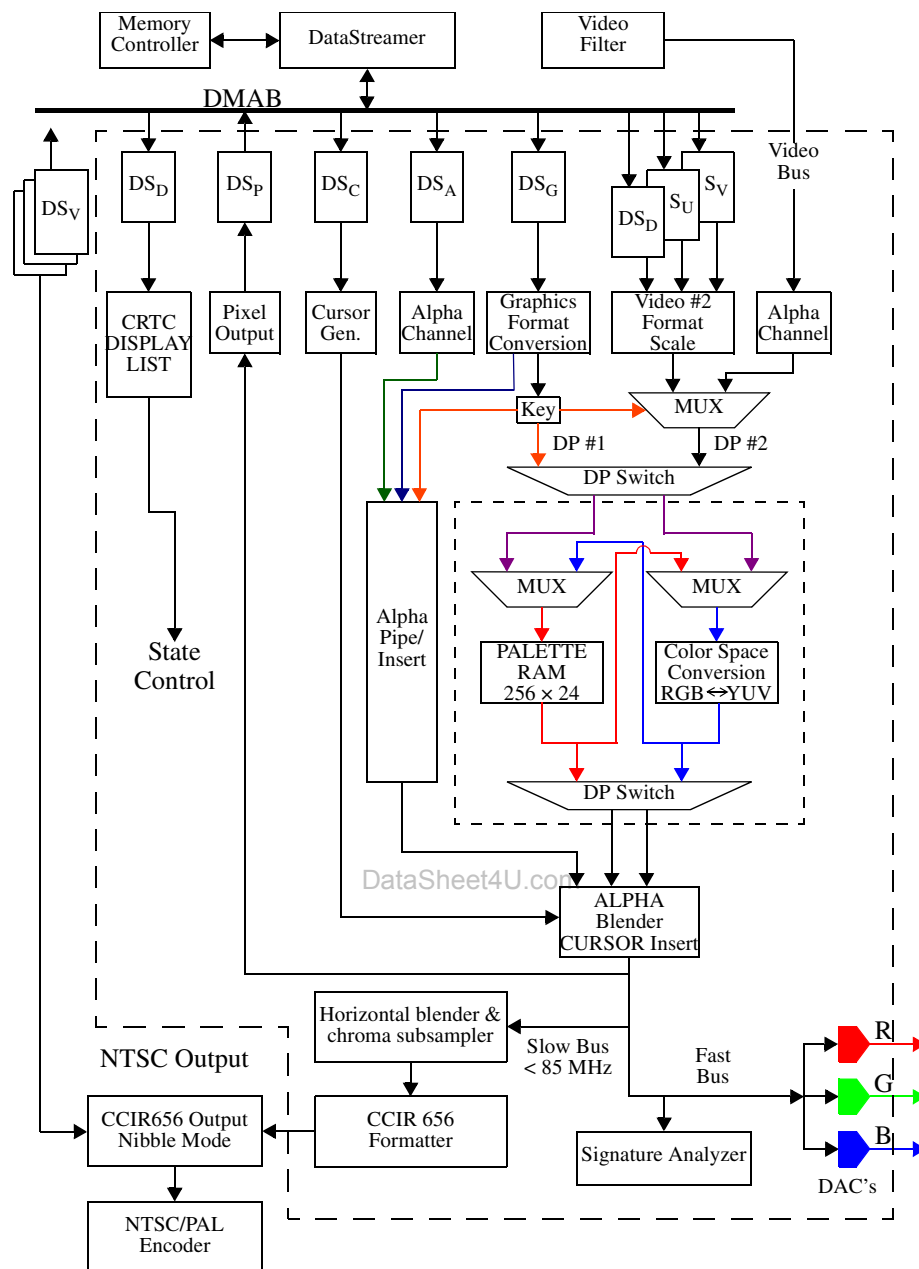


Figure 1-3: Display Refresh Controller

couple of handshake signals, this provides an alternative to PCI for multiple MAP chips to communicate. The data bandwidth supported is up to 60 MB/s depending upon other system activity.

### 1.7.3 Display Refresh Controller

Sophisticated video blending, 2D graphics with alpha blending, PIP, and hardware cursor overlays for EPGs (Electronic Program Guides) and navigation services have been designed into the Display Refresh Controller. Color space conversion,

Gamma correction, and choice of YCbCr or RGB output format is supported. See Figure 1-3 on page 9.

The DRC supports a maximum screen resolution of 1280 × 1024. This monitor resolution requires a minimum pixel clock frequency of 108 MHz to support a rate of 16 bits per pixel.

### 1.7.4 DACs

The MAP-CA DSP RGB DACs (digital-to-analog converters) are part of the Display Refresh Control-

ler block. The 8-bit DACs allow pixel clock rates up to 110 MHz. The MAP-CA DSP generates RS-343A compatible monitor signals into doubly-terminated 75  $\Omega$  load and is capable of driving standard SVGA monitors.

The full scale output level is determined by an external reference voltage  $V_{ref}$  at 1.235V and an external resistor  $R_{nominal} = 1117 \Omega$ . The full scale level can be adjusted by adjusting the resistor value.

The DACs output the three primary analog color signals – red video, green video and blue video – with the video sync information superimposed on the green video output. Also, separate  $h_{sync}$  and  $v_{sync}$  reference signals are provided.

### 1.7.5 I<sup>2</sup>C Interface Unit

The Inter-IC (I<sup>2</sup>C) bus was originally developed by Philips to facilitate communications and control among integrated circuits in consumer electronics. Using this two-wire serial interface, the MAP-CA DSP can function as a master or slave device to relay status and control information to external devices.

The I<sup>2</sup>C interface unit has an additional output signal,  $iic\_select$  that allows MAP-CA DSP software to control an external analog multiplexer/level converter that can switch between a regular I<sup>2</sup>C bus and any other external bus (such as DDC for a monitor interface). This signal can also be used as a general purpose output.

### 1.7.6 ROM Controller

The ROM Controller (ROMCON) unit performs four distinct functions.

- The Chip Configuration and ROM Boot Sequencer is a state machine for reading chip configuration and boot code at system startup.
- The Flash ROM Interface controls the actual reading and writing of an off-chip flash ROM device.
- The Interrupt Controller/Collector provides a means for enabling, setting, and clearing hardware and software interrupts to the VLIW core and PCI bus controller.
- The PLL I/O provides PIO access to the programmable registers related to the various on-chip PLLs. The three PLLs for the core/SDRAM, pixel, and audio clocks are programmed indirectly via PIO registers within the ROMCON unit.

The purpose of the Configuration/Boot Sequencer is to control the boot up process of the chip. During reset, the resistor straps connected to the  $ntsc\_out\_data[7:0]$  pins are examined to determine how MAP-CA digital signal processor will configure itself and boot. If the resistor straps indicate to boot from ROM, the Boot Sequencer directs the Flash ROM Interface Controller to transfer bytes from the external ROM device to the MAP-CA DSP configuration registers and to the PCI configuration registers. The 6 KB line buffer memory of the Video Filter is then used to store the bootstrap program for system boot up. ROMCON copies the next 6 KB from ROM into the Video Filter memory (VfMem) through an 8-bit configuration bus. After the boot code has been loaded, ROMCON uninstalls (restarts) the VLIW CPU, which in turn begins to execute the boot code out of VfMem. The ROMCON unit operates at 27 MHz during the configuration loading, since the core PLL cannot be programmed to be taken out of bypass mode until after the VLIW core has been uninstalled.

Alternatively, for booting via the PCI interface, ROMCON plays a mostly passive role. In this case, an external host loads the VfMem with boot code and initiates boot of the VLIW core via a PIO write to uninstall the VLIW CPU.

ROMCON also runs power-on diagnostics during the boot and may be paused at various points for status testing. ROMCON requires minimal chip resources so that standard power-on diagnostics can run without having to bring up all portions of the chip, allowing the chip to be tested in more manageable stages.

### 1.7.7 Reset Strap

During reset, the eight  $ntsc\_out\_data$  pins are used as inputs to read pre-boot configuration settings. These are settings that must be known before the actual boot process begins – namely, whether the system should boot from ROM and whether PCI should serve as host for its bus. There are also four straps available whose meaning can be defined in software.

Each pin strap is pulled high (to Vdd33) or low (to GND) through a 4.7 K $\Omega$  resistor. The pins are sampled into flip-flops until reset is de-asserted and then saved in the software-visible StrapBits field of PIO register ConfigBusControl. For more information see Section 4.8.6, “Reset Straps,” on page 27.



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# Chapter 2                      Software Development

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The iMMediaTools software toolkit includes

- iMMediaC optimizing, parallelizing C-language compiler
- FIRtree Media Intrinsic C-language extensions
- Assembler
- Linker
- Source-level debugger
- Assembly-level debugger
- Profiling, translating simulator
- Two virtual-machine simulators, one nearly cycle-accurate
- Assorted libraries

The iMMediaC compiler supports development in a host environment that differs from the target environment. The virtual-machine simulators allow testing and debugging on your host system. The supported host development environments are Microsoft Windows NT and Red Hat Linux.

## 2.1 The C Compiler

The MAP-CA digital signal processor development system includes the iMMediaC compiler with FIRtree Media Intrinsic extensions. The FIRtree extensions are proprietary SIMD-style high-speed media processing extensions.

The iMMediaC compiler uses aggressive optimization and global scheduling technology (including trace scheduling) to deliver full hardware performance without using laborious assembly language programming. The compiler allows programmers to focus efforts on algorithm optimization versus assembly scheduling, resource allocation, and debugging.

Unlike existing DSPs or dedicated-function devices which have heretofore been used to meet media processing requirements, the MAP-CA DSP is programmed in a high-level language (C). Benefits of programming in C include

- Reduced development costs
- Reduced time to market
- Lower system costs
- Reduced maintenance time
- Software-based upgrades

The compiler uses complex inline expansion, assertions, and loop unrolling with trace frequency estimation algorithms to maximize C code efficiency. The optimizations include

- Uncover instruction-level parallelism
- Manage registers, pipelines and functional units
- Generate instruction operation schedules that exploit parallelism
- Support extensive global optimization, analysis and scheduling
- Provide local scheduling and optimization
- Support media-oriented machine facilities
- Manages all timing dependencies to maximize scheduling efficiency

The iMMediaC compiler shell program lets you compile, examine, test, profile, assemble, and link source programs with a single command, by using various options.

### 2.1.1 The FIRtree Media Intrinsic Extensions

The FIRtree Media Intrinsic C-language extensions read 128-bit words of data memory, which contain multiple data items, and perform operations simultaneously on each of the items within the word. The FIRtree Media Intrinsic extensions perform operations on partitioned native data types within 32-bit or 64-bit operands, making use of the PLV and PLC registers on the IG-ALU to store intermediate values.

## 2.2 Libraries

The iMMediaTools Software Development Toolkit includes standard C runtime libraries and libraries specifically designed to support MAP-CA digital

signal processor resources, such as the DataStreamer DMA controller and VLx coprocessor for media applications.

## 2.3 Assembler

The MAP-CA DSP assembler lets the programmer take the assembly language source files generated by the compiler, and convert them into object code files, ready for the linker. Developers will not typically write assembly language modules themselves.

## 2.4 Linker

The linker combines object code files into an executable module, accepting both object files and libraries as input. During linking, the linker resolves all external references.

## 2.5 Debugger

The development environment includes a source-level debugger based on gdb (GNU debugger). The extended gdb (egdb) runs on both the Windows NT and Red Hat Linux platforms. The egdb debugger allows the user to:

- load a MAP-CA DSP application from the host PC file system onto the MAP-CA DSP and run it
- set, list, and clear software and hardware breakpoints
- single step through both C source code and assembly instructions
- source level debug of optimized C code
- examine and deposit values into local variables, global variables and PIO space
- examine and deposit values into all registers
- examine the stack, including stack backtracing

Numerous freeware or low-cost gdb GUI front ends exist on both the Windows NT and Linux platforms that will transparently layer on egdb's command line interface and provide a window-based debugging environment.

## 2.6 Simulators

The software developer's toolkit includes three software simulators: trsim, sim, and casim.

Trsim is a high speed, instruction level simulator of the MAP-CA digital signal processor core unit. This simulator works on an intermediate representation of a software program and can be used to initially develop applications and experiment with the performance of different algorithms and use of compiler options.

Sim is also a high-speed, instruction-level simulator. This simulator works off actual MAP-CA DSP binaries. Sim is a functional simulator of the MAP-CA DSP core unit with data cache, DataStreamer controller, and a subset of I/O devices. It provides runtime checks against resource constraints and all MAP-CA DSP features necessary to simulate a MAP-CA DSP running a real-time operating system and applications.

Casim is a nearly cycle-accurate software simulator for the MAP-CA DSP and models more accurately the core, DataStreamer DMA controller, VLx, instruction and data caches, memories, and buses. Like sim, this simulator operates off actual MAP-CA digital signal processor binaries. Casim provides more detailed runtime checks against resource constraints. Casim provides visibility into internal machine state and bandwidth and augments debugging and tuning of interactions between VLIW core, DataStreamer controller, and VLx programs.

Sim and casim work in conjunction with the source-level debugger.

## 2.7 Boot

Software must contain boot code for execution on MAP-CA DSP. This boot code configures TLBs and caches. Equator software tools can automatically add boot code when building an application.

# Chapter 3 BGA PIN\_OUT Assignments

Signal assignment on the BGA352 package is shown here. Figure 3-1 shows the bottom view, with the balls facing the viewer. Figure 3-2 shows the top view of the chip with pin assignments..

In Table 3-1, pins marked with an (b) have multiple functions assigned (see Chapter 4)

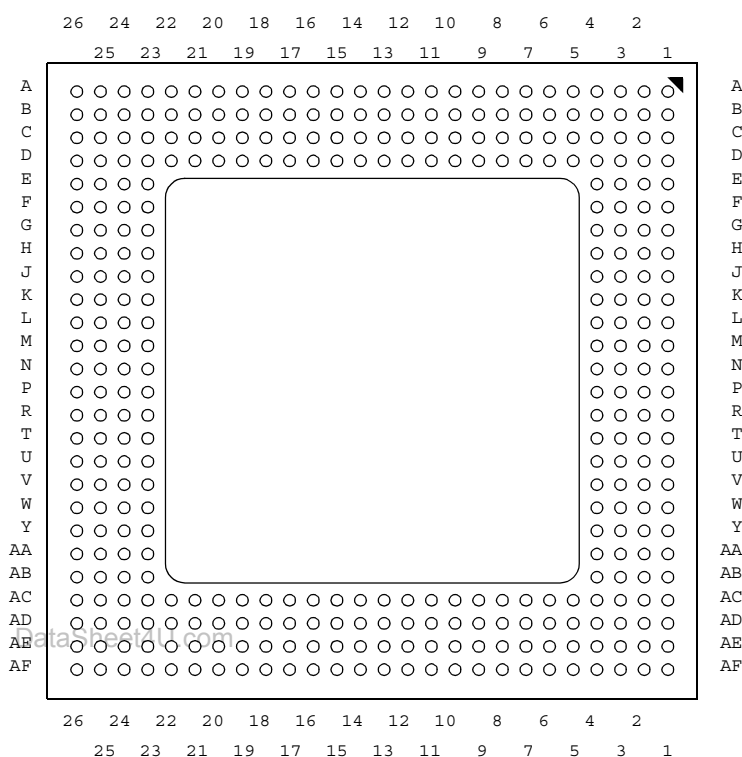


Figure 3-1: View of balls from the bottom

Table 3-1: Pin Assignments

Signal Name	Ball	Signal Name	Ball	Signal Name	Ball	Signal Name	Ball
audioclk_byp_in	A2 <sup>a</sup>	ntsc_out_data[6]	C6 <sup>b</sup>	video_ina[8]	B8 <sup>b</sup>	video_inb[4]	C12 <sup>b</sup>
aVdd18	B3	ntsc_out_data[7]	D7 <sup>b</sup>	video_ina[9]	C9 <sup>b</sup>	video_inb[5]	B12 <sup>b</sup>
aVss	C4	video_ina[0]	B6 <sup>b</sup>	tcia_inuse	A8	video_inb[6]	A13 <sup>b</sup>
aVdd18	D5	video_ina[1]	C7 <sup>b</sup>	tcia_sync	A9	video_inb[7]	C13 <sup>b</sup>
aVss	A3	video_ina[2]	D8 <sup>b</sup>	tcia_clk	C10	video_inb[8]	D13 <sup>b</sup>
pixelclk_byp_in	B4 <sup>b</sup>	video_ina[3]	A6 <sup>b</sup>	tcia_vdac	B10	video_inb[9]	B13 <sup>b</sup>
ntsc_out_data[3]	C5 <sup>b</sup>	video_ina[4]	B7 <sup>b</sup>	ntsc_ina_clk27	A10 <sup>b</sup>	tcib_inuse	A14
ntsc_out_data[4]	B5 <sup>b</sup>	video_ina[5]	A7 <sup>b</sup>	video_inb[0]	C11 <sup>b</sup>	tcib_sync	B14
ntsc_out_data[5]	A4 <sup>b</sup>	video_ina[6]	C8 <sup>b</sup>	video_inb[1]	B11 <sup>b</sup>	tcib_clk	C14
		video_ina[7]	D9 <sup>b</sup>	video_inb[2]	A11 <sup>b</sup>	ntsc_inb_clk27	D14
				video_inb[3]	D12 <sup>b</sup>	vsync	A16
						hsync	B15
						tms	B16
						trst	C15
						aVss	A17
						gdac_fscale	B17
						gdac_comp	C16
						aVdd33	D16
						gdac_green	C17
						gdac_blue	A18
						gdac_red	A19
						gdac_cvgg(aVss)	C18

Signal Name	Ball	Signal Name	Ball	Signal Name	Ball	Signal Name	Ball	Signal Name	Ball
aVddx (aVdd18)	D17	sddata[26]	J25	aVssq (aVss)	AD23	sddata[58]	AF12	pci_frame_	AC2
aVssx (aVss)	B19	sddata[27]	K24	aVddq (aVdd18)	AE24	sddata[59]	AE12	pci_cbe_[2]	AC1
tdi	A20	sddqm[3]	K26	aVdd18	AF25	sddqm[7]	AD12	pci_ad[16]	AB1
tck	B20	sddata[28]	L24	sdclk_byp_in	AC21 <sup>c</sup>	sddata[60]	AC12	pci_ad[17]	AA2
tdo	C19	sddata[29]	L25	pclk	AD22 <sup>d</sup>	sddata[61]	AF11	pci_ad[18]	AA3
no connection	A21	sddata[30]	M25	coreclk_byp_in	AE23 <sup>c</sup>	sddata[62]	AD11	pci_ad[19]	AA1
sddata[0]	B21	sddata[31]	M24	pci_clk	AF24	sddata[63]	AF10	pci_ad[20]	Y3
sddata[1]	C20	sdcs_[0]	M26	sddata[34]	AD21	pci_ad[0]	AE10	pci_ad[21]	Y2
sddata[2]	D19	sdcs_[1]	N24	sddata[35]	AF23	pci_ad[1]	AD10	pci_ad[22]	W1
sddata[3]	A23	sdcs_[2]	N25	sddqm[4]	AF22	pci_ad[2]	AE9	pci_ad[23]	W2
sddqm[0]	B22	sdcs_[3]	P26	sddata[36]	AE21	pci_ad[3]	AF8	pci_idsel	W3
sddata[4]	C21	sdrtnclk	P24	sddata[37]	AC19	pci_ad[4]	AD9	pci_cbe_[3]	V1
sddata[5]	A24	sdclk	R25	sddata[38]	AD20	pci_ad[5]	AE8	pci_ad[24]	V3
sddata[6]	B23	sdclk1	R24	sddata[39]	AF21	pci_ad[6]	AC9	pci_ad[25]	U1
sddata[7]	C22	sdclk2	T26	sddata[40]	AF20	pci_ad[7]	AD8	pci_ad[26]	U4
sddata[8]	D21	sdras_	T25	sddata[41]	AC18	pci_cbe_[0]	AF7	pci_ad[27]	U3
sddata[9]	A25	sdcas_	T24	sddata[42]	AD19	pci_ad[8]	AF6	pci_ad[28]	U2
sddata[10]	B24	sdwe_	U25	sddata[43]	AE19	pci_ad[9]	AD7	pci_ad[29]	T3
sddata[11]	C23	sdadr[0]	U24	sddqm[5]	AD18	pci_ad[10]	AC8	pci_ad[30]	T2
sddqm[1]	D22	sdadr[1]	V26	sddata[44]	AF19	pci_ad[11]	AE6	pci_ad[31]	R3
sddata[12]	D24	sdadr[2]	V24	sddata[45]	AE18	pci_ad[12]	AF5	pci_req_[0]	R2
sddata[13]	B26	sdadr[3]	W25	sddata[46]	AD17	pci_ad[13]	AF4	pci_req_[1]	R1
sddata[14]	C25	sdadr[4]	Y26	sddata[47]	AE17	pci_ad[14]	AD6	pci_req_[2]	P3
sddata[15]	E24	sdadr[5]	W24	sddata[48]	AF17	pci_ad[15]	AF3	pci_gnt_[0]	P1
sddata[16]	D25	sdadr[6]	Y25	sddata[49]	AD16	pci_cbe_[1]	AE4	pci_gnt_[1]	N1
sddata[17]	D26	sdadr[7]	Y24	sddata[50]	AF16	pci_par	AD5	pci_gnt_[2]	N2
sddata[18]	E25	sdadr[8]	AA25	sddata[51]	AC15	pci_serr_intb_	AC6	pclk_out	N3 <sup>e</sup>
sddata[19]	F24	sdadr[9]	AA24	sddqm[6]	AD15	pci_stop_	AF2	pci_rst_	M2
sddqm[2]	F26	sdadr[10]	AB26	sddata[52]	AE15	pci_devsel_	AE3	pci_inta_	M3
sddata[20]	G24	sdadr[11]	AD26	sddata[53]	AF15	pci_trdy_	AD4	pci_pme_	L1
sddata[21]	G25	sdadr[12]	AB24	sddata[54]	AD14	pci_irdy_	AC5	iic_sda	L2
sddata[22]	H25	sdadr[13]	AD25	sddata[55]	AE14	Vss	AD2	iic_sck	L3
sddata[23]	H24	sddata[32]	AC24	sddata[56]	AF13	Vdd18	AE1	iic_select	K1
sddata[24]	H26	sddata[33]	AB23	sddata[57]	AD13	Vdd18	AC3	iis_in_data	K3
sddata[25]	J24	aVss	AC22			Vss	AB3	iis_in_lr	J1

Signal Name	Ball	Signal Name	Ball	Signal Name	Ball	Signal Name	Ball	Signal Name	Ball
iis_in_bclk	K4	Vdd18	V23	Vdd33	R23	Vss	E26	Vss	AE5
iis_out_data[0]	J2 <sup>b</sup>	Vdd18	W23	Vdd33	U23	Vss	F25	Vss	AF1
iis_out_data[1]	J3 <sup>b</sup>	Vdd18	AA23	Vdd33	Y23	Vss	G26	Vss	AE2
iis_out_data[2]	H2 <sup>b</sup>	Vdd18	AC20	Vdd33	AD24	Vss	J26	Vss	AD1
iis_out_lr	G1	Vdd18	AC16	Vdd33	AC23	Vss	K25	Vss	AB2
iis_out_bclk	H3	Vdd18	AC11	Vdd33	AC17	Vss	L26	Vss	Y1
iec958_in	G2	Vdd18	AC7	Vdd33	AC14	Vss	N26	Vss	V2
iec958_out	G3	Vdd18	AB4	Vdd33	AC13	Vss	P25	Vss	T1
rom_cs_	F1	Vdd18	AA4	Vdd33	AC10	Vss	R26	Vss	P2
ntsc_out_hsync	F3 <sup>b</sup>	Vdd18	W4	Vdd33	AD3	Vss	U26	Vss	M1
ntsc_out_vsync	E1 <sup>b</sup>	Vdd18	T4	Vdd33	AC4	Vss	V25	Vss	K2
ntsc_out_data[0]	E2 <sup>b</sup>	Vdd18	P4	Vdd33	Y4	Vss	W26	Vss	H1
ntsc_out_data[1]	E3 <sup>b</sup>	Vdd18	N4	Vdd33	V4	Vss	AA26	Vss	F2
ntsc_out_data[2]	C1 <sup>b</sup>	Vdd18	M4	Vdd33	R4	Vss	AB25	Vss	D1
audioclk_out	C2	Vdd18	J4	Vdd33	L4	Vss	AC26	Vss	D2
pixelclk_out	D3 <sup>f</sup>	Vdd18	G4	Vdd33	H4	Vss	AC25	a. If unused, NC <sub>i</sub> (Vss) and tie to ground b. Multi-function pins c. connect to 27 MHz or lower clock d. connect to 27 MHz clock e. If unused, NC <sub>o</sub> and floating f. Leave floating	
Vdd18	D6	Vdd33	F4	Vss	A1	Vss	AE26		
Vdd18	D11	Vdd33	C3	Vss	B1	Vss	AE25		
Vdd18	D15	Vdd33	D4	Vss	B2	Vss	AF26		
Vdd18	D20	Vdd33	E4	Vss	A5	Vss	AE22		
Vdd18	D23	Vdd33	D10	Vss	B9	Vss	AE20		
Vdd18	E23	Vdd33	D18	Vss	A12	Vss	AF18		
Vdd18	G23	Vdd33	C24	Vss	A15	Vss	AE16		
Vdd18	H23	Vdd33	D23	Vss	B18	Vss	AF14		
Vdd18	K23	Vdd33	F23	Vss	A22	Vss	AE13		
Vdd18	M23	Vdd33	J23	Vss	A26	Vss	AE11		
Vdd18	P23	Vdd33	L23	Vss	B25	Vss	AF9		
Vdd18	T23	Vdd33	N23	Vss	C26	Vss	AE7		

	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20	21	22	23	24	25	26	
A	Vss	audioclk_byp_in	aVss	ntsc_out_data[5]	Vss	video_ina[3]	video_ina[5]	tcia_inuse	tcia_sync	ntsc_inaclk27	video_inb[2]	Vss	video_inb[6]	tcib_inuse	Vss	vsync	aVss	gdac_blue	gdac_red	tdi	no connection	Vss	sddata[31]	sddata[5]	sddata[9]	Vss	A
B	Vss	Vss	aVdd18	pixelclk_byp_in	ntsc_out_data[4]	video_ina[0]	video_ina[4]	video_ina[8]	Vss	tcia_vdac	video_inb[1]	video_inb[5]	video_inb[9]	tcib_sync	hsync	tms	gdac_fscale	Vss	aVss(aVss)	tck	sddata[0]	sddata[0]	sddata[6]	sddata[10]	Vss	sddata[13]	B
C	ntsc_out_data[2]	audioclk_byp_in	Vdd33	aVss	ntsc_out_data[3]	ntsc_out_data[6]	video_ina[1]	video_ina[6]	video_ina[9]	tcia_clk	video_inb[0]	video_inb[4]	video_inb[7]	tcib_clk	trst	gdac_comp	gdac_green	gdac_cv	tdo	sddata[1]	sddata[4]	sddata[7]	sddata[11]	Vdd33	sddata[14]	Vss	C
D	Vss	Vss	pixelclk_out	Vdd33	aVdd18	Vdd18	ntsc_out_data[7]	video_ina[2]	video_ina[7]	Vdd33	Vdd18	video_inb[3]	video_inb[8]	ntsc_inbclk27	Vdd18	aVdd33	aVdd18	Vdd33	sddata[2]	Vdd18	sddata[8]	sddata[11]	Vdd33	sddata[12]	sddata[16]	sddata[17]	D
E	ntsc_out_data[0]	ntsc_out_data[1]	Vdd33																				Vdd18	sddata[15]	sddata[18]	Vss	E
F	rom_cs_	Vss	ntsc_out_hsync	Vdd18																			Vdd33	sddata[19]	Vss	sddqm[2]	F
G	iis_out_lr	iec958_in	iec958_out	Vdd18																			Vdd18	sddata[20]	sddata[21]	Vss	G
H	Vss	iis_out_data[2]	iis_out_bclk	Vdd33																			Vdd18	sddata[23]	sddata[22]	sddata[24]	H
J	iis_in_lr	iis_out_data[0]	iis_out_data[1]	Vdd18																			Vdd33	sddata[25]	sddata[26]	Vss	J
K	iic_select	Vss	iis_in_data	iis_in_bklk																			Vdd18	sddata[27]	Vss	sddqm[3]	K
L	pci_pme_	iic_sda	iic_sck	Vdd33																			Vdd33	sddata[28]	sddata[29]	Vss	L
M	Vss	pci_rst_	pci_inta_	Vdd18																			Vdd18	sddata[31]	sddata[30]	sdcs_0	M
N	pci_gnt_1	pci_gnt_2	pclk_out	Vdd18																			Vdd33	sdcs_1	sdcs_2	Vss	N
P	pci_gnt_0	Vss	pci_req_2	Vdd18																			Vdd18	sdrtnclk	Vss	sdcs_3	P
R	pci_req_1	pci_req_0	pci_ad_3	Vdd33																			Vdd33	sdclk1	sdclk	Vss	R
T	Vss	pci_ad_30	pci_ad_29	Vdd18																			Vdd18	sdcas_	sdras_	sdclk2	T
U	pci_ad_25	pci_ad_28	pci_ad_27	pci_ad_26																			Vdd33	sdadr[0]	sdwe_	Vss	U
V	pci_cbe_3	Vss	pci_ad_24	Vdd33																			Vdd18	sdadr[2]	Vss	sdadr[1]	V
W	pci_ad_22	pci_ad_23	pci_idsel	Vdd18																			Vdd18	sdadr[5]	sdadr[3]	Vss	W
Y	Vss	pci_ad_21	pci_ad_20	Vdd33																			Vdd33	sdadr[7]	sdadr[6]	sdadr[4]	Y
AA	pci_ad_19	pci_ad_17	pci_ad_18	Vdd18																			Vdd18	sdadr[9]	sdadr[8]	Vss	AA
AB	pci_ad_16	Vss	Vdcb(Vss)	Vdd18																			sddata[33]	sdadr[12]	Vss	sdadr[10]	AB
AC	pci_cbe_2	pci_frame_	Vdcb(Vdd18)	Vdd33	pci_indy_	pci_serr_inb_	Vdd18	pci_ad_10	pci_ad_6	Vdd33	Vdd18	sddata[60]	Vdd33	Vdd33	sddata[51]	Vdd18	Vdd33	sddata[41]	sddata[37]	Vdd18	sdclk_byp_in	aVss	Vdd33	sddata[32]	Vss	Vss	AC
AD	Vss	Vsbb(Vss)	Vdd33	pci_trdy_	pci_par	pci_ad_9	pci_ad_7	pci_ad_4	pci_ad_11	sddata[62]	sddqm[7]	sddata[57]	sddata[54]	sddqm[6]	sddata[49]	sddata[46]	sddqm[5]	sddata[42]	sddata[38]	pcik	aVss(aVss)	Vdd33	sdadr[13]	sdadr[11]	AD		
AE	Vsbc(Vdd18)	Vss	pci_devse_	pci_ad_11	Vss	pci_ad_11	Vss	pci_ad_5	pci_ad_2	pci_ad_0	Vss	sddata[59]	Vss	sddata[55]	sddata[52]	Vss	sddata[47]	sddata[45]	sddata[43]	Vss	sddata[36]	Vss	coreclk_byp_in	aVdd(aVdd18)	Vss	Vss	AE
AF	Vss	pci_stop_	pci_ad_15	pci_ad_13	pci_ad_12	pci_ad_8	pci_cbe_0	pci_ad_3	Vss	sddata[63]	sddata[61]	sddata[58]	sddata[56]	Vss	sddata[53]	sddata[50]	sddata[48]	Vss	sddata[44]	sddata[40]	sddata[39]	sddqm[4]	sddata[35]	pci_clk	aVdd18	Vss	AF

- Vdd33
- SDRAM
- I<sup>2</sup>C
- Multi-function pins
- aVdd18
- aVss
- Vss
- PLLs
- CRT (DRC)
- Vdd18
- PCI
- JTAG
- I<sup>2</sup>S

Figure 3-2: MAP-CA DSP pins viewed from top

# Chapter 4 Signal Descriptions

## 4.1 Interface Summary

( ) --- valid during boot-up  
< > --- valid during reset  
# --- active low signal

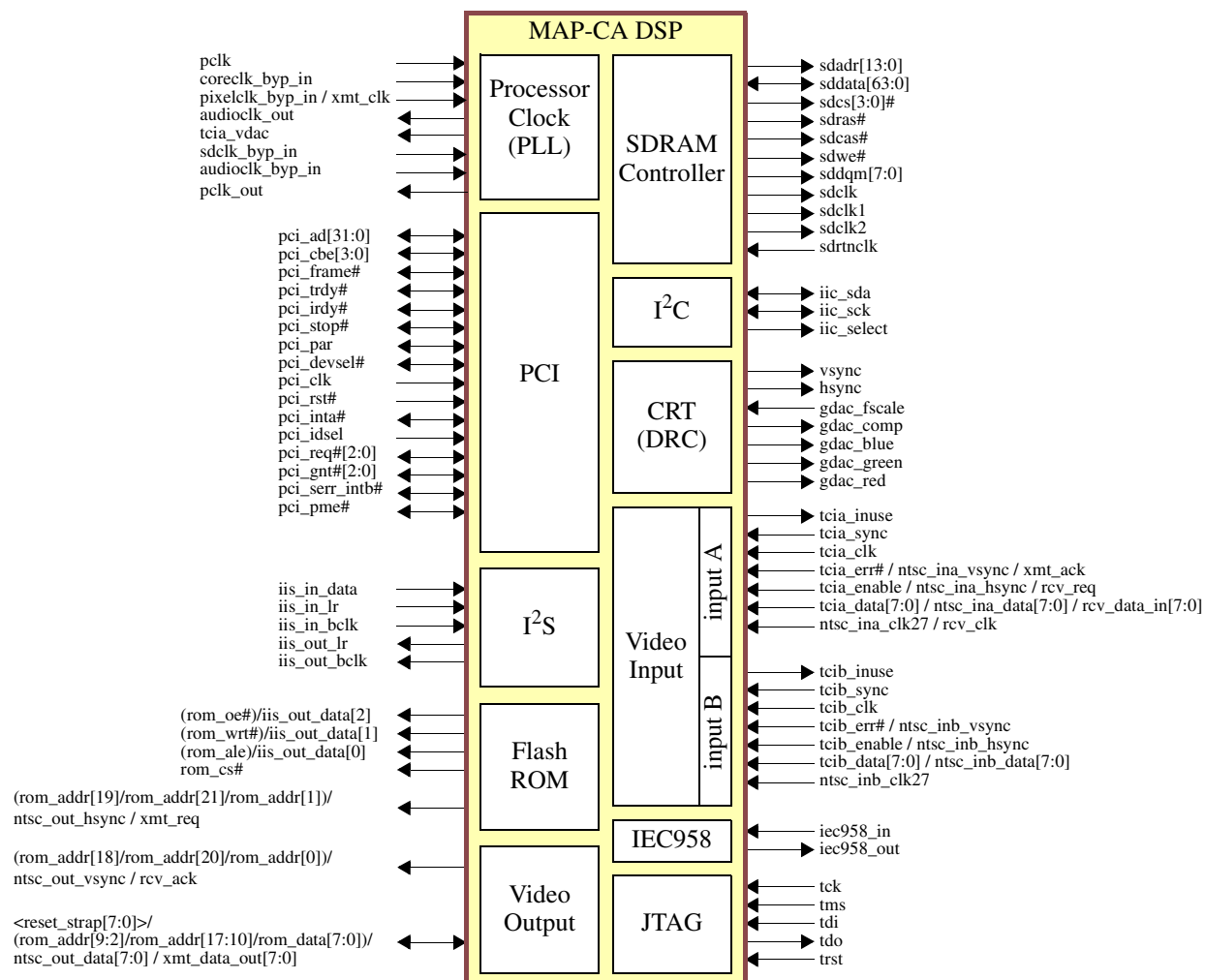


Figure 4-1: MAP-CA DSP Interface

## 4.2 Legend

All tables in this section use the following abbreviations:

- A ..... analog signal
- B ..... bi-directional
- I ..... input
- O ..... output
- od ..... open drain or active pull down
- trailing “\_” ..... active low
- () ..... parentheses indicate number of shared signal pins

## 4.3 Processor Clock

There is a single 27 MHz reference clock signal for generating internal clocks.

Signal	# of Pins	I/O	Description
pclk	1	I	27 MHz VCXO input that serves as the reference signal to the three on-chip PLLs. It is also a reference clock for the video output and the I <sup>2</sup> C interface
coreclk_byp_in	1	I	Bypass input for the core clock
sdclk_byp_in	1	I	Bypass input for the SDRAM clock.
pixelclk_byp_in	1	I	Transmit clock for GPDP (xmt_clk) or PLL-bypass input for pixel clock
audioclk_byp_in	1	I	Bypass input for the audio clock.
pclk_out (coreclk_out)	1	O	Output for the core clock.
pixelclk_out	1	O	Output for the pixel clock.
audioclk_out	1	O	Audio clock output used for I <sup>2</sup> S interface master MCLK
tcia_vdac	1	O	Sigma-delta output from software loop filter for controlling external VCXO for pclk
TOTAL	9		

Table 4-1: Processor Clock Signal Description



## 4.4 SDRAM

The MAP-CA DSP supports either SDRAM or SGRAM memory system using the signals shown in Table 4-2. The MAP-CA digital signal processor supports a memory system of 64-bit or 32-bit data width. The MAP-CA DSP supports DRAM widths of 8-bit, 16-bit or 32-bits.

Signal	# of Pins	I/O	Description
sdadr[13:0]	14	O	Address lines indicate row addresses when sdras_ is active and indicate column addresses when sdcas_ is active
sddata[63:0]	64	B	Data Input/Output lines transfer data between the memory and the MAP-CA DSP. These are also input mask bits for Write-per-Bit. When block write is activated, these lines provide column address mask
sdc*_[3:0]	4	O	Chip Select signal lines indicate that the command on the output lines is for each memory chip. If this signal is high, the output command(s) will be ignored by each corresponding memory chip
sdras_	1	O	sdras_ is part of the output command to the SDRAM/SGRAM
sdcas_	1	O	sdcas_ is part of the output command to the SDRAM/SGRAM
sdwe_	1	O	Write enable (sdwe_) is part of the output command
sddqm[7:0]	8	O	During read, sddqm=1 turns off the output buffers of SDRAM/SGRAM. During write, sddqm=1 prevents a write to the current memory location
sdclk,sdclk1,sdclk2	3	O	sdclk, sdclk1 and sdclk2 are driven by the MAP-CA DSP SDRAM clock. All SDRAM/SGRAM input signals are sampled on the positive edge of sdclk
sdrtnclk	1	I	sdrtnclk is driven by sdclk. This signal is used for latching the data from SDRAM/SGRAM
TOTAL	97		

Table 4-2: Memory Interface Signals

## 4.5 PCI Bus

The MAP-CA DSP provides the PCI bus as the primary system interface. Table 4-3 lists the PCI signals.

Signal	# of Pins	I/O	Description
pci_ad[31:0]	32	B	PCI multiplexed address and data lines. The address is driven when pci_frame_ is first asserted. Data is transferred on this bus in subsequent clocks.
pci_cbe_[3:0]	4	B	For PCI cycles, the bus command and byte enables are used to transfer the PCI command during the address phase and are used to transfer byte lane enables during subsequent data phases.
pci_frame_	1	B	Transaction framing for PCI transfers. The initial assertion indicates the address phase and the start of a PCI transaction. Continued assertion determines the burst size of the transaction.
pci_trdy_	1	B	The target ready signal is asserted when the PCI target is ready for a data transfer.
pci_irdy_	1	B	The initiator ready signal is asserted when the PCI master is ready for a data transfer.
pci_stop_	1	B	pci_stop_ is asserted by the target to request the master to stop the current transaction.
pci_par	1	B	A single parity bit is calculated over pci_ad[31:0], and pci_c_be[3:0] and transferred over this signal.
pci_devsel_	1	B	A target asserts the pci_devsel_ signal line to indicate it has decoded the address on the pci_ad[31:0] bus and will participate in (claim) the current transaction. The PCI bus master must monitor the pci_devsel_ signal line to determine if a target bus has claimed the transaction or if it will execute a Master Abort termination. pci_devsel_ will be tri-stated from the leading edge of pci_rst_. pci_devsel_ remains tri-stated until driven by the target.
pci_clk	1	I	pci_clk provides timing for all PCI transactions on the PCI bus. All other PCI signals are sampled on the rising edge of pci_clk, and all timing parameters are defined with respect to this edge. NOTE: The MAP-CA DSP core clock PLL does not use this clock as a core PLL reference clock.
pci_rst_	1	I	This signal indicates a reset of all PCI resources. In addition, the internal MAP-CA DSP CPU core, etc. are reset by the assertion of this signal. PCI pad cell drivers are disabled by the assertion of this signal, as specified in the PCI 2.1 document.

Table 4-3: PCI Interface Signals

Signal	# of Pins	I/O	Description
pci_inta_	1	B (od)	When the MAP-CA digital signal processor is not designated as the PCI bus “HOST”, then this signal is the open drain output generating an asynchronous level sensitive interrupt on the PCI bus. The MAP-CA DSP pad cell does <u>not</u> contain a pull up for this signal. When the MAP-CA DSP is designated as the PCI bus “HOST” then this signal is an interrupt request input from PCI devices. The MAP-CA DSP sees and utilizes this interrupt.
pci_idsel	1	I	The initialization device select is used as a slot addressed chip select input during configuration read and write transactions. This signal is inactive in a self-hosted configuration.
pci_req_[2:0]	3	B	The assertion of pci_req_ in a non-self-hosted configuration indicates that the MAP-CA DSP desires the use of the PCI bus.
pci_gnt_[2:0]	3	B	The assertion of pci_gnt_ in a non-self-hosted environment indicates that the MAP-CA DSP has been granted the use of the PCI bus.
pci_serr_intb_	1	B (od)	This open drain output may generate either an asynchronous level sensitive interrupt or the PCI bus or optionally signal SYSTEM ERROR. See the MAP-CA digital signal processor configuration control register for the signal’s current use. When the MAP-CA DSP is not designated as the PCI bus “HOST”, then this signal is the open drain output generating an asynchronous level sensitive interrupt on the PCI bus. The MAP-CA DSP pad cell does <u>not</u> contain a pull up for this signal. When the MAP-CA DSP is designated as the PCI bus “HOST” , then this signal is an interrupt request input from PCI devices. The MAP-CA DSP sees and utilizes this interrupt.
pci_pme_	1	B (od)	When MAP-CA DSP is not in PCI host mode, this signal is an open drain output used to request a change in power management state. When MAP-CA DSP is in PCI host mode, this is an input signal to which PCI devices indicate changes in power management state.
TOTAL	54		

Table 4-3: PCI Interface Signals

## 4.6 IEC958

The MAP-CA DSP provides an IEC958 interface as shown in Table 4-4.

Signal	# of Pins	I/O	Description
iec958_in	1	I	Serial input line for IEC958 digital audio.
iec958_out	1	O	Serial output line for IEC958 digital audio.
TOTAL	2		

Table 4-4: IEC958 Interface Signals

## 4.7 I<sup>2</sup>S

The MAP-CA DSP provides interfaces for I<sup>2</sup>S digital audio input and output as shown in Table 4-5.

Signal	# of Pins	I/O	Description
iis_in_data	1	I	Serial data input
iis_in_lr	1	I	Select left/right channel in the serial input
iis_in_bclk	1	I	I <sup>2</sup> S input bit clock
iis_out_data[2:0]	3	O	Serial output data (3 stereo lines)
iis_out_lr	1	O	Select left/right channel in serial output lines
iis_out_bclk	1	O	I <sup>2</sup> S output bit-rate clock
TOTAL <sup>a</sup>	8		

Table 4-5: I<sup>2</sup>S Interface Signals

- a. audioclk\_out, defined in Section 4.3 on page 18, serves as the output MCLK in master mode.

## 4.8 Multi-Function Signal Pins

Table 4-6 lists the pin name and pin number of the shared pins on the MAP-CA digital signal processor, along with the signal name associated with each mode-specific function.

Ball #	Pin Name	ITU-656 Mode	TCI Mode	GPDP Mode	ROM Boot Mode	Reset Mode
B4	pixelclk_byp_in	-	-	xmt_clk	-	-
C5	ntsc_out_data[3]	-	-	xmt_data_out[3]	rom_addr[5,13]/rom_data[3]	reset_strap[3]
B5	ntsc_out_data[4]	-	-	xmt_data_out[4]	rom_addr[6,14]/rom_data[4]	reset_strap[4]
A4	ntsc_out_data[5]	-	-	xmt_data_out[5]	rom_addr[7,15]/rom_data[5]	reset_strap[5]
C6	ntsc_out_data[6]	-	-	xmt_data_out[6]	rom_addr[8,16]/rom_data[6]	reset_strap[6]
D7	ntsc_out_data[7]	-	-	xmt_data_out[7]	rom_addr[9,17]/rom_data[7]	reset_strap[7]
B6	video_ina[0]	ntsc_ina_data[0]	tcia_data[0]	rcv_data_in[0]	-	-
C7	video_ina[1]	ntsc_ina_data[1]	tcia_data[1]	rcv_data_in[1]	-	-
D8	video_ina[2]	ntsc_ina_data[2]	tcia_data[2]	rcv_data_in[2]	-	-
A6	video_ina[3]	ntsc_ina_data[3]	tcia_data[3]	rcv_data_in[3]	-	-
B7	video_ina[4]	ntsc_ina_data[4]	tcia_data[4]	rcv_data_in[4]	-	-
A7	video_ina[5]	ntsc_ina_data[5]	tcia_data[5]	rcv_data_in[5]	-	-
C8	video_ina[6]	ntsc_ina_data[6]	tcia_data[6]	rcv_data_in[6]	-	-
D9	video_ina[7]	ntsc_ina_data[7]	tcia_data[7]	rcv_data_in[7]	-	-
B8	video_ina[8]	ntsc_ina_hsync	tcia_enable	rcv_req	-	-
C9	video_ina[9]	ntsc_ina_vsync	tcia_err#	xmt_ack	-	-
A10	ntsc_ina_clk27	-	-	rcv_clk	-	-
C11	video_inb[0]	ntsc_inb_data[0]	tcib_data[0]	-	-	-
B11	video_inb[1]	ntsc_inb_data[1]	tcib_data[1]	-	-	-
A11	video_inb[2]	ntsc_inb_data[2]	tcib_data[2]	-	-	-
D12	video_inb[3]	ntsc_inb_data[3]	tcib_data[3]	-	-	-
C12	video_inb[4]	ntsc_inb_data[4]	tcib_data[4]	-	-	-
B12	video_inb[5]	ntsc_inb_data[5]	tcib_data[5]	-	-	-
A13	video_inb[6]	ntsc_inb_data[6]	tcib_data[6]	-	-	-
C13	video_inb[7]	ntsc_inb_data[7]	tcib_data[7]	-	-	-
D13	video_inb[8]	ntsc_inb_hsync	tcib_enable	-	-	-
B13	video_inb[9]	ntsc_inb_vsync	tcib_err#	-	-	-
J2	iis_out_data[0]	-	-	-	rom_ale	-
J3	iis_out_data[1]	-	-	-	rom_wrt#	-
H2	iis_out_data[2]	-	-	-	rom_ole#	-
F3	ntsc_out_hsync	-	-	xmt_req	rom_addr[21,19,1]	-
E1	ntsc_out_vsync	-	-	rcv_ack	rom_addr[20,18,0]	-
22	ntsc_out_data[0]	-	-	xmt_data_out[0]	rom_addr[2,10]/rom_data[0]	reset_strap[0]
E3	ntsc_out_data[1]	-	-	xmt_data_out[1]	rom_addr[3,11]/rom_data[1]	reset_strap[1]
C1	ntsc_out_data[2]	-	-	xmt_data_out[2]	rom_addr[4,12]/rom_data[2]	reset_strap[2]

Table 4-6: Multiple Signal Pins

### 4.8.1 Transport Channel Interfaces (TCI)

The MAP-CA digital signal processor provides two parallel/serial TCI interfaces. See Table 4-7 and

Signal	# of Pins	I/O	Description
tcia_data[7:0]	(8)	I	TCI data input: All 8 bits of input are used in parallel mode. Only tci_data[0] is used in serial mode
tcia_enable	(1)	I	Transport channel enable: 1 = accept a TCI input sample 0 = ignore TCI input sample
tcia_inuse	1	O	Video input port external mux select. This pin can also be used as a general purpose output if dynamic muxing of the video input is not required.
tcia_sync	1	I	Demodulator/FEC has marked the synchronization point in the MPEG2 transport stream packet. The packet size is programmable.
tcia_err_	(1)	I	This active low signal indicates that the Demodulator/FEC has detected an uncorrectable error in the current packet.
tcia_clk	1	I	Transport channel clock
TOTAL	3(10)		

Table 4-7: Primary TCI Interface Signals

Table 4-8.

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Signal	# of Pins	I/O	Description
tcib_data[7:0]	(8)	I	TCI data input: All 8-bits of input are used in parallel mode. Only tci_data[0] is used in serial mode.
tcib_enable	(1)	I	Transport channel enable: 1 = accept a TCI input sample 0 = ignore TCI input sample
tcib_inuse	1	O	Video input port external mux select. This pin can also be used as a general purpose output if dynamic muxing of the video input is not required.
tcib_sync	1	I	Demodulator/FEC has marked the synchronization point in the MPEG2 transport stream packet. The packet size is programmable.
tcib_err_	(1)	I	This active low signal indicates that the Demodulator/FEC has detected an uncorrectable error in the current packet.
tcib_clk	1	I	Transport channel clock
TOTAL	3(10)		

Table 4-8: Secondary TCI Interface Signals

## 4.8.2 ITU-656 Inputs

Analog video can be digitized according to ITU-R BT.601/656 via a NTSC, PAL or SVIDEO decoder and then input to the MAP-CA DSP. The interface signals for the primary and secondary video inputs are shown in Table 4-9 and Table 4-10.

Signal	# of Pins	I/O	Description
ntsc_ina_clk27	1	I	27MHz clock from video decoder
ntsc_ina_hsync	(1)	I	Horizontal sync
ntsc_ina_vsync	(1)	I	Vertical sync
ntsc_ina_data[7:0]	(8)	I	ITU-R BT.601/656 formatted video input stream
TOTAL	1(10)		

Table 4-9: Primary ITU-R BT.601/656 Input Interface Signals

Signal	# of Pins	I/O	Description
ntsc_inb_clk27	1	I	27MHz clock from video decoder
ntsc_inb_hsync	(1)	I	Horizontal sync
ntsc_inb_vsync	(1)	I	Vertical sync
ntsc_inb_data[7:0]	(8)	I	ITU-R BT.601/656 formatted video input stream
TOTAL	1(10)		

Table 4-10: Secondary ITU-R BT.601/656 Input Interface Signals

## 4.8.3 ITU-656 Output

The MAP-CA digital signal processor has a digital ITU-R BT.601/656 output interface as shown in Table 4-11.

Signal	# of Pins	I/O	Description
pclk	(1)	I	27MHz pixel clock from VCXO clock input. See Section 4.3.
ntsc_out_hsync	1	O	Horizontal sync
ntsc_out_vsync	1	O	Vertical sync
ntsc_out_data[7:0]	8	B	ITU-R BT.601/656 formatted NTSC/PAL output data configured as input only when used for ROM interface or reset strap
TOTAL	10(1)		

Table 4-11: ITU-R BT.601/656 Output Interface Signals

#### 4.8.4 General Purpose Data Port (GPDP)

The primary video input port and the video output port can be coupled together to function as a general purpose 8-bit duplex data port.

Signal	# of Pins	I/O	Description
xmt_clk	(1)	I	transmitter output clock, shared with pixelclk_byp_in
xmt_req	(1)	O	GPDP on MAP-CA digital signal processor is ready to transmit data. Shared with ntsc_out_hsync
xmt_ack	(1)	I	output source is ready to accept data, shared with video_ina[9]
xmt_data_out[7:0]	(8)	O	Parallel data output on ntsc_out_data[7:0]
rcv_clk	(1)	I	Receiver input clock shared with ntsc_ina_clk27
rcv_req	(1)	I	Input source is ready to send data. Shared with video_ina[8]
rcv_ack	(1)	O	GPDP on MAP-CA DSP is ready to accept data, shared with ntsc_out_vsync
rcv_data_in[7:0]	(8)	I	Parallel data input on video_ina[7:0]
TOTAL	(22)		

Table 4-12: GPDP Interface Signals

#### 4.8.5 Flash ROM

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A Flash ROM (EEPROM) interface is provided on the MAP-CA digital signal processor to assist in boot-up. The Flash ROM is active during the boot up process and must be disabled through its chip-select signal. See Section 3.3.2, *Equator Hardware Reference Manual, Volume 4, MAP-CA I/O Interfaces* for information about programming the timing parameters of the ROM interface.



Signal	# of Pins	I/O	Description
rom_cs_	1	O	This pin is the chip enable signal.
rom_oe_	(1)	O	This active low signal is asserted on ROM read cycles. This signal is multiplexed with iis_out_data[2].
rom_wrt_	(1)	O	This active low signal is asserted on ROM write cycles. This signal is multiplexed with iis_out_data[1].
rom_ale	(1)	O	Address latch enable: rom_addr[9:2] and rom_addr[19:18] are latched on the falling edge, rom_addr[17:10] and rom_addr[21:20] are latched on the rising edge; this signal is multiplexed with iis_out_data[0].
rom_addr[19:18]/ rom_addr[21:20]/ rom_addr[1:0]	(2)	O	rom_addr[19]/rom_addr[21]/rom_addr[1] is muxed with ntsc_out_hsync. rom_addr[18]/rom_addr[20]/rom_addr[0] is muxed with ntsc_out_vsync.
rom_addr[9:2]/ rom_addr[17:10]/ rom_data[7:0]	(8)	B	ROM data and address bus. These signals are multiplexed with ntsc_out_data[7:0].
TOTAL	1(13)		

Table 4-13: ROM Interface Signals

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#### 4.8.6 Reset Straps

The board resistor straps are sampled on the de-assertion (rising edge) of pci\_rst\_.

Signal	# of Pins	I/O	Description
reset_strap[7:4] (sw_strap[3:0])	(4)	I	Resistor straps for use by software. These signals are multiplexed with rom_data[7:4] and ntsc_out_data[7:4].
reset_strap[1] (pci_host)	(1)	I	VDD = 1 = MAP-CA DSP is hosting the primary PCI bus GND = 0 = MAP-CA DSP is not hosting The signal is multiplexed with rom_data[1] and ntsc_out_data[1].
reset_strap[0] (rom_boot)	(1)	I	VDD = 1 = Flash ROM is used for boot GND = 0 = ROMless boot The signal is multiplexed with rom_data[0] and ntsc_out_data[0].
TOTAL	(6)		

Table 4-14: Reset Straps

## 4.9 Analog CRT

An RGB monitor can be directly driven by the MAP-CA digital signal processor.

Signal	# of Pins	I/O	Description
vsync	1	O	Vertical synchronization signal for CRT
hsync	1	O	Horizontal synchronization signal for CRT
gdac_fscale	1	A	Full scale current adjusting resistor
gdac_comp	1	A	Vref bypass and compensation capacitor
gdac_blue	1	A	Analog blue output
gdac_green	1	A	Analog green output
gdac_red	1	A	Analog red output
TOTAL	7		

Table 4-15: CRT Interface Signals

## 4.10 I<sup>2</sup>C

The MAP-CA DSP provides an I<sup>2</sup>C interface to communicate with external peripherals such as NTSC decoders, NTSC encoders, and demodulators. The pin description is shown in Table 4-16.

Signal	# of Pins	I/O	Description
iic_sda	1	B (od)	I <sup>2</sup> C data line
iic_sck	1	B (od)	I <sup>2</sup> C clock line
iic_select	1	O	I <sup>2</sup> C-bus external mux select. This pin can also be used as general purpose output
TOTAL	3		

Table 4-16: I<sup>2</sup>C Interface Signals

## 4.11 Boundary Scan (JTAG)

MAP-CA DSP supports boundary scan interface based on *IEEE 1149.1 Standard Test Access Port and Boundary-Scan Architecture* for testing the MAP-CA DSP and other devices on the board. A BSDL (Boundary Scan Description Language) file is available from Equator.

Signal	# of Pins	I/O	Description
$t_{ck}$	1	I	Reference clock. This specifies the timing for all the transactions of JTAG interface.
$t_{ms}$	1	I	Test interface mode select signal.
$t_{di}$	1	I	Test interface data input
$t_{do}$	1	O	Test interface data output
$t_{rst}$	1	I	Active-low reset for the circuitry. It must be low at power up.
TOTAL	5		

Table 4-17: JTAG Interface Signals

### 4.11.1 Pull Up Resistors

The *IEEE 1149.1* requires that TDI, TMS, and TRST have internal pull-up resistors. In the MAP-CA digital signal processor, TRST has an internal pull-down resistor and TRST, as well as other TAP pins, can be left unconnected if the system does not use Boundary Scan.

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### 4.11.2 TAP State Machine

Figure 4-2 shows the internal states of the TAP (Test Access Port) state machine. These conform to the state transitions specified by *IEEE 1149.1*. The state changes according to the value on TMS at the rising edge of TCK. The TDI value is sampled at the rising edge of TCK, and shifted at the falling edge. The TDO value changes at the falling edge of TCK. When not in the Shift-DR or Shift-IR state, TDO has a high-impedance. If TRST is zero, TDO goes to the Test Logic Reset state asynchronously.

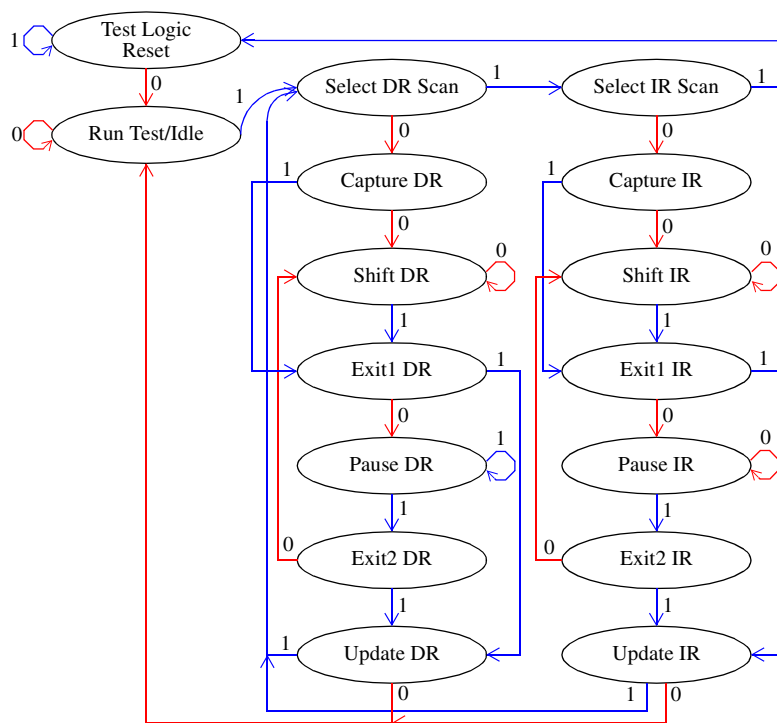


Figure 4-2: TAP State Transition Diagram

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### 4.11.3 Instruction Registers

The MAP-CA digital signal processor supports public instructions defined in *IEEE 1149.1*. Encoding of the Boundary Scan instructions are listed in Table 4-18. The instruction register is five bits long and instruction bit combinations not listed here are reserved.

Instruction	Code	Description
BYPASS	0b11111	connects TDI and TDO with one bit shift register
EXTEST	0b00000	The instruction connects the boundary scan registers and the MAP-CA digital signal processor system pins.
SAMPLE/ PRELOAD	0b11100	sample the MAP-CA DSP output or load the MAP-CA DSP input, without affecting the MAP-CA DSP system pins
IDCODE	0b11001	connects the MAP-CA DSP device with the Device Identification Register between TDI and TDO
INTEST	0b11011	The instruction connects boundary scan registers and MAP-CA DSP internals.
CLAMP	0b11101	sets all system output pins to the value previously loaded by boundary scan instruction.
HIGHZ	0b11110	places all system output pins in a high-impedance state.

Table 4-18: JTAG Instructions

### 4.11.4 Test Data Registers

Figure 4-3 illustrates the structure of the boundary scan logic..

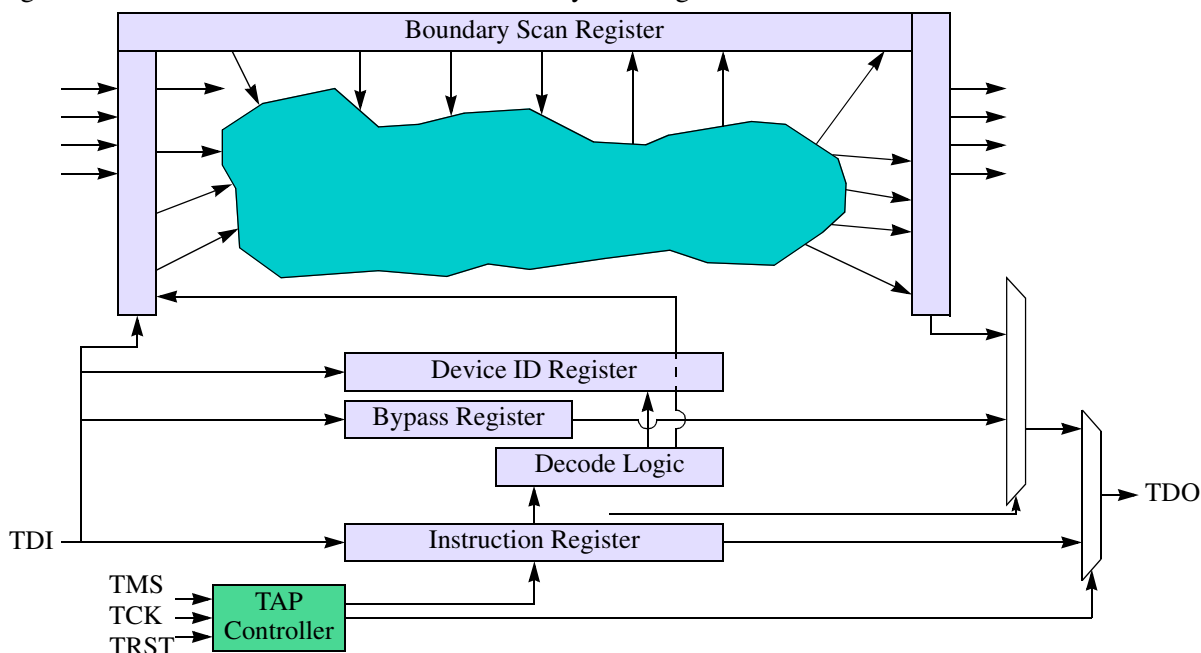


Figure 4-3: Boundary Scan Block Diagram

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### 4.11.5 Boundary Scan Register

The boundary scan register is a shift register on the PAD for controlling the MAP-CA digital signal processor's system pins. Using EXTEST and SAMPLE/PRELOAD commands, it can perform a JTAG (IEEE 1149.1) compliant boundary scan test. The relation between the MAP-CA DSP pins and the boundary scan register can be found in the BSDL.

### 4.11.6 Device Identification Register

Figure 4-4 shows the format of the Device Identification Register. For the MAP-CA DSP, the Manufacturer's ID is a twelve bit value equal to b000000001111. The upper twelve bits of the Part Number are b010001010011. The lower four bits of the Part Number and the Version Number varies according to the major or minor revision of the MAP-CA DSP.

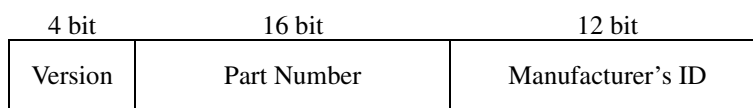


Figure 4-4: Device Identification Register

## 4.12 Power/Ground Pins

The following power and ground pins are provided on MAP-CA digital signal processor.

Name	# of Pins	Description
Vdd18	30	Digital Core Vdd
Vdd33	27	I/O Power Supply 3.3V
Vss	57	Digital Vss
aVdd18	3	Clean analog Vdd
aVss	4	Clean analog Vss
aVdd33	1	Clean analog Vdd
aVddq	1	Clean analog Vdd for Core PLL
aVssq	1	Clean analog Vss for Core PLL
aVddx	1	Clean analog Vdd for Video DAC
aVssx	1	Clean analog Vss for Video DAC
gdac_cvgg	1	Clean analog Vss for Video DAC
TOTAL	127	

Table 4-19: Power/Ground Pins

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## 4.13 Signal List Summary

Interface	Frequency	Pin Count	Remarks
MAP-CA DSP clocks	Various	9	reference clocks including bypass clocks
PCI	66/33 MHz	54	host/system interface
SDRAM	133 MHz	97	memory interfaces (SDRAM, EEPROM)
Flash ROM	5 MHz	1(13)	
Reset straps	-	(6)	configuration inputs
IEC958	32/44.1/48 kHz sample rate	2	digital audio I/O
I <sup>2</sup> S	32/44.1/48 kHz sample rate 1.536/2.116 MHz bit rate	8	digital audio I/O
ITU-R BT.601/656 In	27 MHz	28 (22) <sup>a</sup>	two video input streams: 2 TCI or 2 ITU-656, or 1 TCI and 1 ITU-656, or GPDP and 1 video input (TCI or ITU-656)
Parallel/Serial TCI	30/80 MHz		
GPDP	60 MHz		
ITU-R BT.601/656 Out	27 MHz	10(1)	digital video out or GPDP
CRT out	25-110 MHz	7	analog video out
I <sup>2</sup> C	100/400 kHz	3	peripheral control
Test	25 MHz	5	boundary scan (JTAG)
Signal pins total	-	220	
Power/Ground	-	127	
Miscellaneous	-	2	reserved (tie to ground)
	-	3	no connection (leave floating)
<b>TOTAL</b>		<b>352</b>	

Table 4-20: MAP-CA DSP Pin List

- a. See Section 4.8.4 for more information about the GPDP shared pins.

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# Chapter 5 External Connection Examples

## 5.1 SDRAM

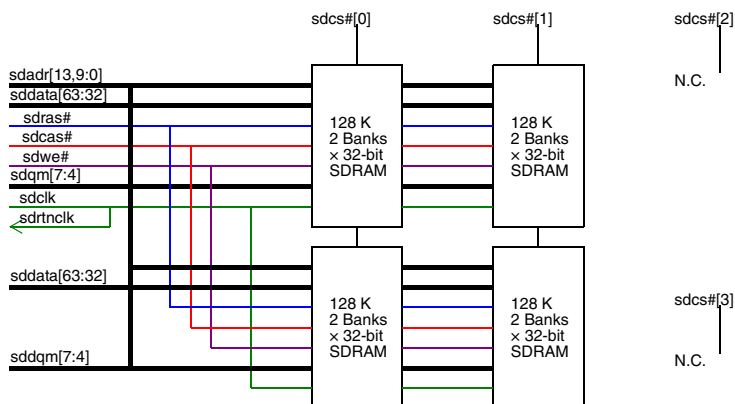


Figure 5-1: 64-bit, 4MB configuration using  $\times 32$ , 8Mb parts

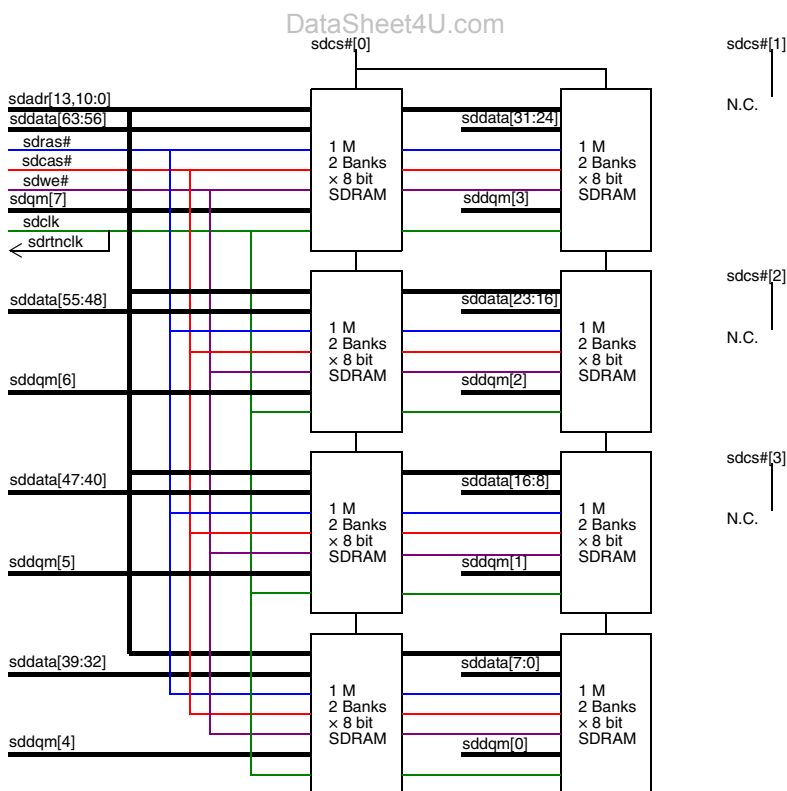


Figure 5-2: 64-bit, 16 MB configuration using  $\times 8$ , 16 Mb parts

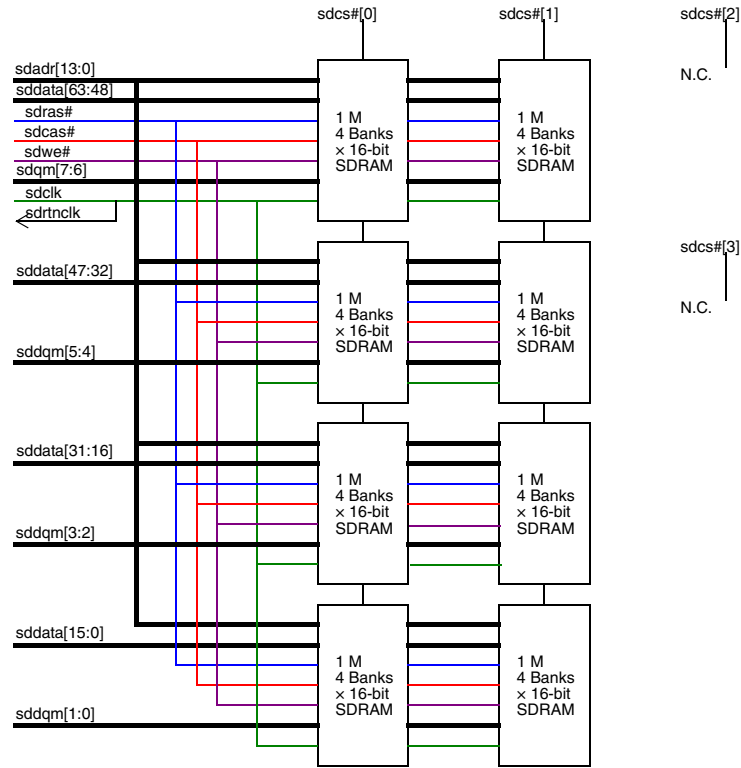


Figure 5-4: 64-bit, 64 MB configuration using x16, 64 Mb parts

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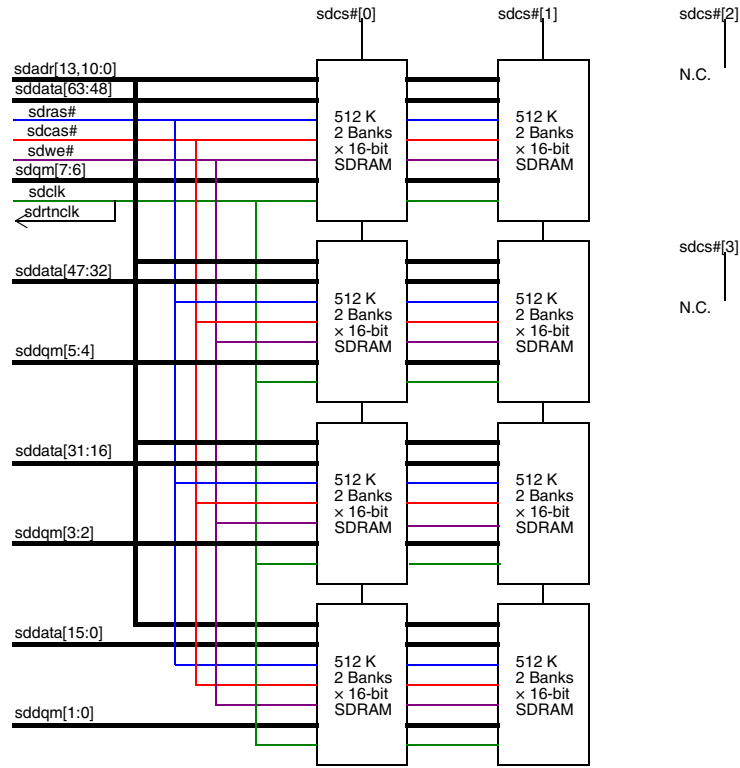


Figure 5-3: 64-bit, 16 MB configuration using  $\times 16$  16Mb parts

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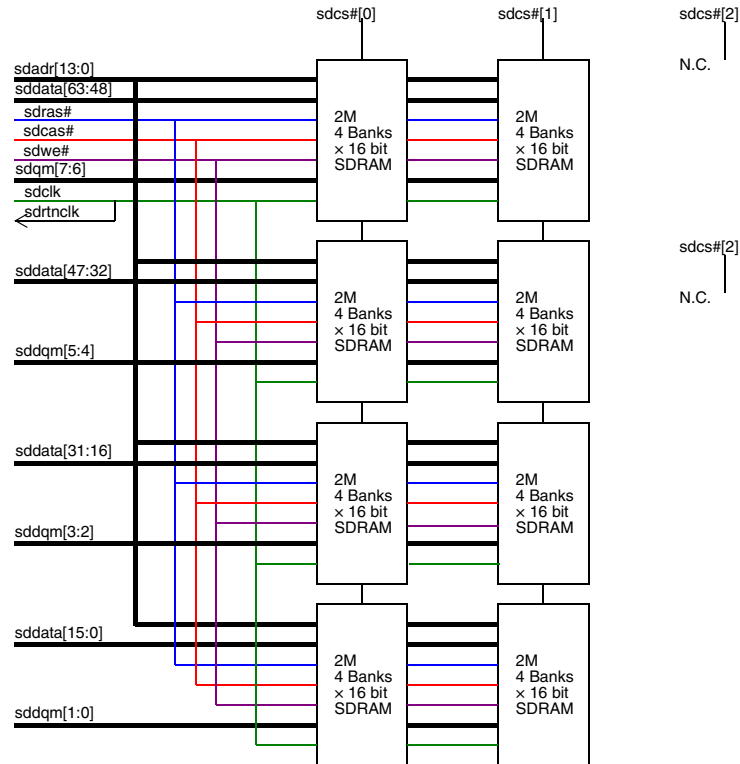


Figure 5-6: 64-bit, 128MB configuration using  $\times 16$ , 128Mb parts

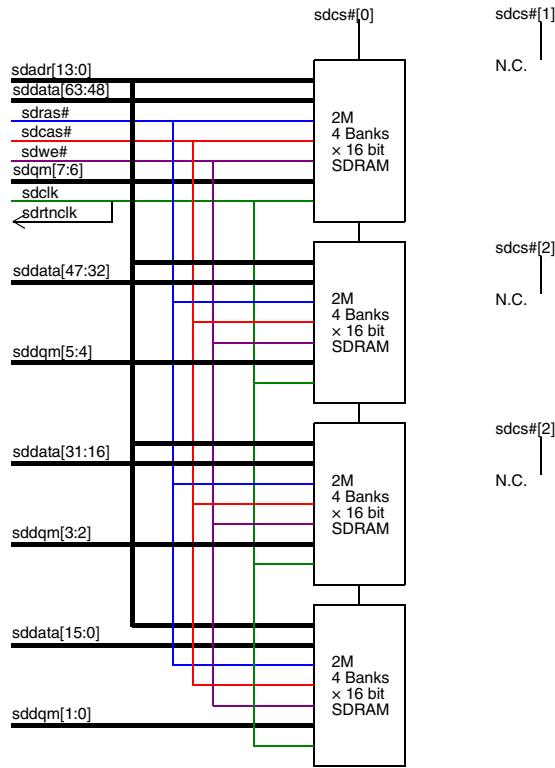


Figure 5-5: 64-bit, 64MB configuration using x16, 128Mb parts

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## 5.2 IEC958

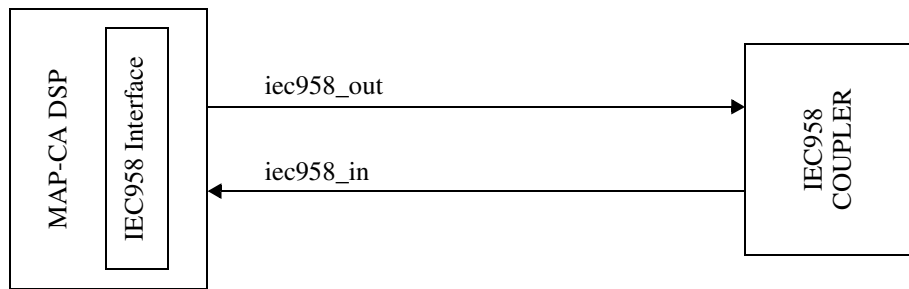


Figure 5-7: IEC958 Interface

### 5.3 I<sup>2</sup>S

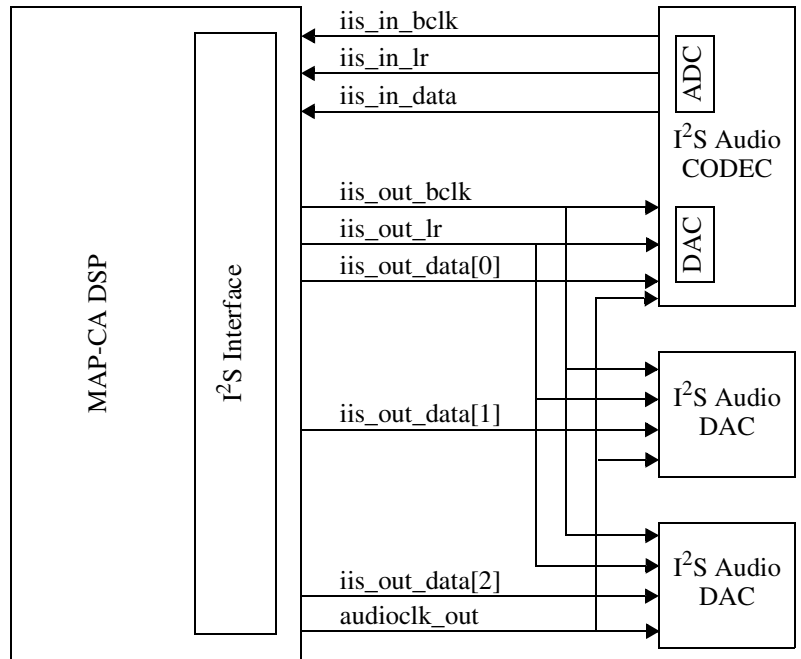


Figure 5-8: I<sup>2</sup>S Interface

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**NOTE:** Reference designs may use `iis_out_bclk` and `iis_out_lr` as input clocks. See the reference board documentation.

# 5.4 Transport Channel Interface (TCI)

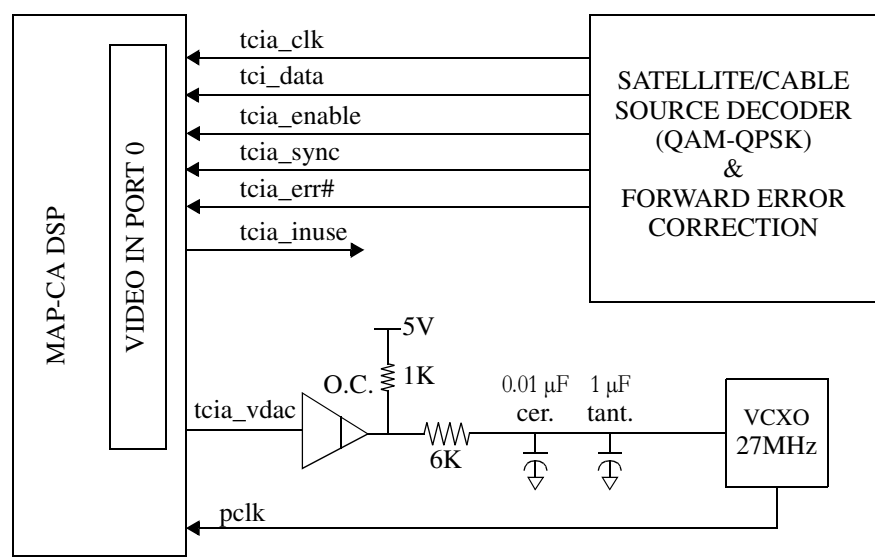


Figure 5-9: Transport Channel Interface

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### 5.5 NTSC Decoder

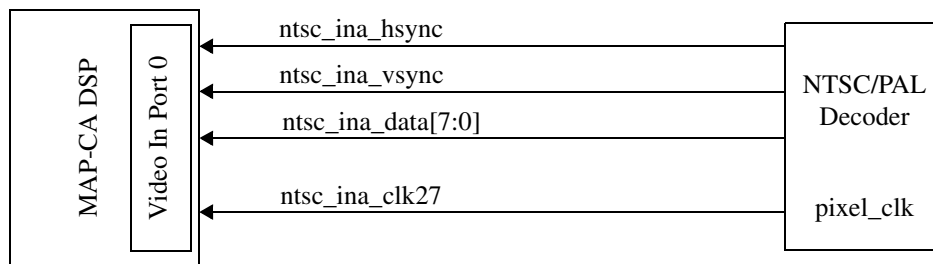


Figure 5-10: ITU-R BT.656 NTSC/PAL Decoder Interface

### 5.6 NTSC Encoder

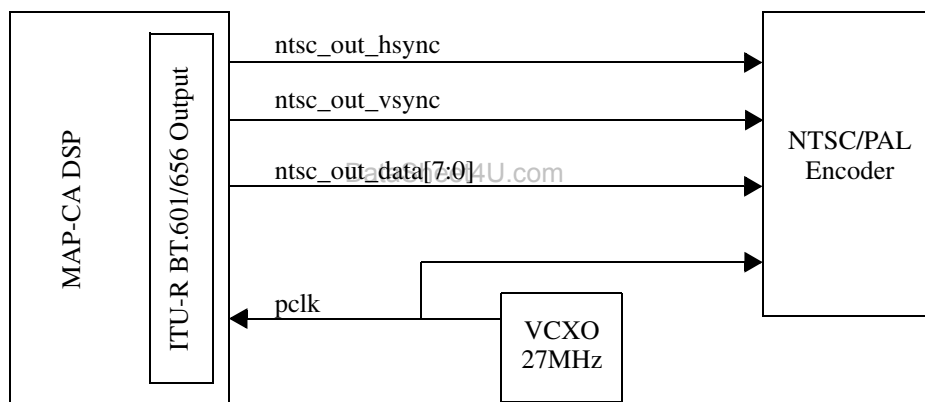


Figure 5-11: NTSC/PAL Encoder Interface

# 5.7 CRT

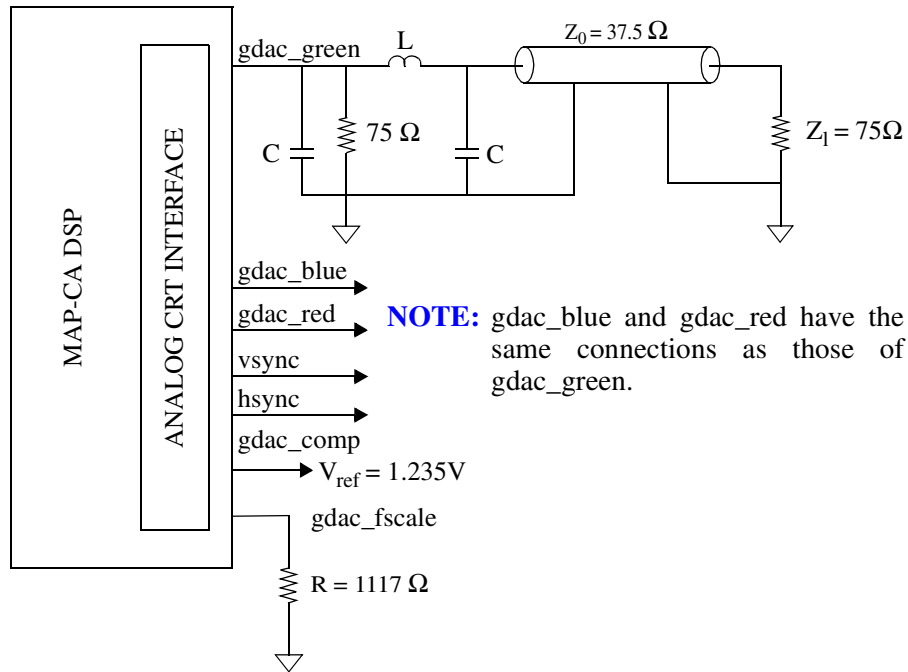


Figure 5-12: CRT

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# 5.8 I<sup>2</sup>C

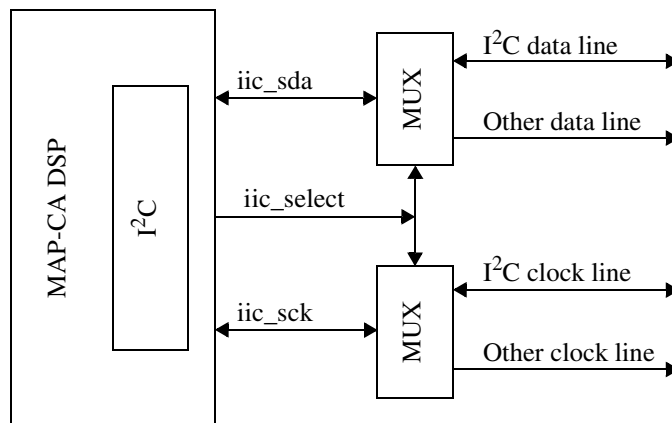


Figure 5-13: I<sup>2</sup>C Interface



# 5.9 ROM

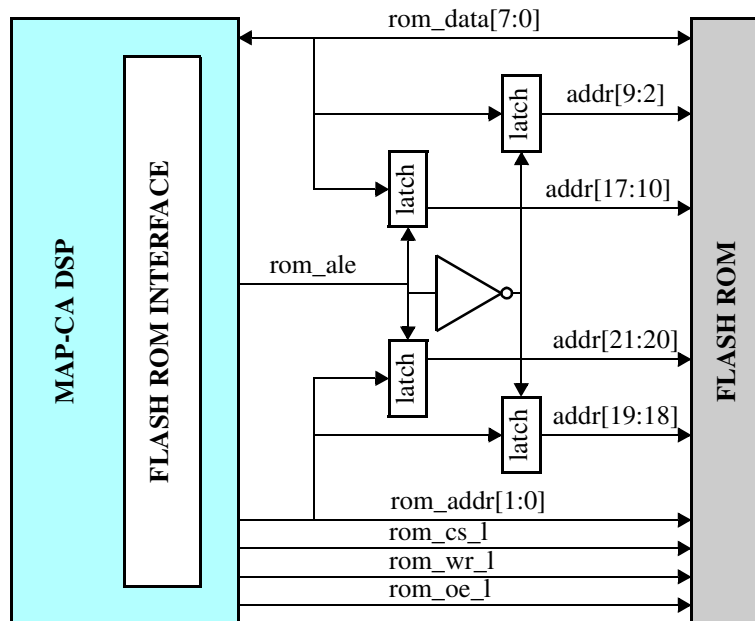


Figure 5-14: ROM Connections

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# Chapter 6 Electrical Specifications

## 6.1 Absolute Maximum Ratings

Stresses which occur above or below those listed in Table 6-1 may cause permanent damage to the MAP-CA digital signal processor. This is a stress rating only, and functional operation of the device at these or any other conditions above those listed in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Parameter	Min	Max	Unit
Vdd18 (measured to Vss)	-0.5	2.5	V
Vdd33 (measured to Vss)	-0.5	4.6	V
aVdd18 (measured to aVss)	-0.5	2.5	V
aVdd33 (measured to aVss)	-0.5	4.6	V
aVddq (measured to aVssq)	-0.5	2.5	V
aVddx (measured to aVss)	-0.5	2.5	V
Voltage on any signal pin <sup>a</sup>	-0.5	Vdd33 + 0.5	V
Storage Temperature	-55	125	°C
Junction Temperature	0	85	°C

Table 6-1: Absolute Maximum Ratings

- a. This device employs CMOS devices on all signal pins. It should be handled as an ESD sensitive device. Voltage on any signal pin that exceeds the power supply voltage by more than +0.5 V can induce destructive latchup.

## 6.2 Power Supply Specifications

Power Supply	Nominal Voltage	Voltage Variation
Vdd18	1.8 V	±5%
Vdd33	3.3 V	±5%
aVdd18	1.8 V	±5%
aVdd33	3.3 V	±5%
aVddx	1.8 V	±5%

Table 6-2: Voltage Variation

The typical estimated total power consumption is 6 W.



**NOTE:** During power-up, Vdd18 must come up before or simultaneously with Vdd33. Vdd33 should not exceed Vdd18 by more than 0.4 V while Vdd18 is less than 1V. There is no restriction on the power-down sequence.

Power Supply	Nominal Voltage	Estimated Max Steady State Current
Vdd18	1.8 V	2.8 A
Vdd33	3.3 V	0.25 A
aVdd18	1.8 V	10 mA
aVdd33	3.3 V	80 mA
aVddx	1.8 V	5 mA

Table 6-3: Steady State Current

## 6.3 DC Characteristics

Parameter	Description	Condition	Min	Max	Unit
$V_{IL}$	input low voltage	-	-0.5	0.7	V
$V_{IH}$	input high voltage	-	0.5 Vdd33	Vdd33 + 0.5	V
$V_{OL}$	output low voltage	$I_{out} = 1500 \mu A$	-	0.1 Vdd33	V
$V_{OH}$	output high voltage	$I_{out} = 500 \mu A$	0.9 Vdd33	-	V
$I_{LI}$	input leakage current	$0 < V_{in} < V_{dd33}$	-10	10	$\mu A$
$I_{OZ}$	tri-state output leakage	$0 < V_{in} < V_{dd33}$	-10	10	$\mu A$
$C_{IN}$	input pin capacitance	-	-	10	pF
$C_{IDSEL}$	idsel pin capacitance	-	-	8	pF
$C_{CLK}$	pci_clk pin capacitance	-	5	12	pF

Table 6-4: PCI Signals

Parameter	Description	Condition	Min	Max	Unit
V <sub>IL</sub>	input low voltage	-	-0.5	0.6	V
V <sub>IH</sub>	input high voltage	-	2.0	V <sub>dd33</sub> + 0.5	V
V <sub>OL</sub>	output low voltage	I <sub>out</sub> = 2 mA	-	0.4	V
V <sub>OH</sub>	output high voltage	I <sub>out</sub> = 2 mA	2.4	-	V
I <sub>LI</sub>	input leakage current	0 < V <sub>in</sub> < V <sub>dd33</sub>	-10	10	μA
I <sub>LIPU</sub>	leakage current of pull up pin	0 < V <sub>in</sub> < V <sub>dd33</sub>	-300	10	μA
I <sub>LIPD</sub>	leakage current of pull down pin	0 < V <sub>in</sub> < V <sub>dd33</sub>	-10	300	μA
I <sub>OZ</sub>	tri-state output leakage	0 < V <sub>in</sub> < V <sub>dd33</sub>	-10	10	μA
C <sub>IN</sub>	input pin capacitance	-	-	10	pF
C <sub>IO</sub>	input/output pin capacitance	-	-	12	pF

Table 6-5: Non-PCI Signals

Parameter	Min	Max	Unit
Operating Temperature	0	75	°C

Table 6-6: Temperature Rating



**NOTE:** The relationship between `gdac_fscale_f` and the full scale output current on IOG is:

$$\text{IOG (mA)} = 24.12 * \text{gdac\_comp(V)} / \text{gdac\_fscale\_f(ohm)}$$

where `gdac_comp` is 1.235 V (typical).

Parameter	Symbol	Min	Typical	Max	Unit
Resolutions (FS)	-	8	8	8	Bits
Integral nonlinearity	INL	-	-	± 2	LSB
Differential nonlinearity	DNL	-	-	± 1	LSB

Table 6-7: Video DAC Outputs

Parameter	Symbol	Min	Typical	Max	Unit
Monotonicity	-	-	guaranteed	-	-
White level relative to blank	-	17.69	19.05	20.40	mA
White level relative to black	-	16.74	17.62	18.50	mA
Black level relative to blank	-	0.95	1.44	1.90	mA
Blank level on IOG	-	6.29	7.62	8.96	mA
Blank level on IOR, IOB	-	0	5	50	$\mu$ A
Sync level on IOG	-	0	5	50	$\mu$ A
LSB size	-	-	69.1	-	$\mu$ A
DAC-to-DAC matching	-	-	2	5	%
Output resistance	RAOUT	-	37.5	-	$\Omega$

Table 6-7: Video DAC Outputs

## 6.4 AC Characteristics

The AC timing specifications listed in this section assume a 20 pF load on all output signals of the MAP-CA DSP.

### 6.4.1 PLL Reference Clock Input

Description	Value
Rise/fall time	2 ns maximum (0.4V to 2.4V)
Duty cycle	50% $\pm$ 10%
Maximum cycle to cycle jitter	$\pm$ 200 ps

Table 6-8: PLL Reference Clock Input Conditions

## 6.4.2 SDRAM Interface Timing

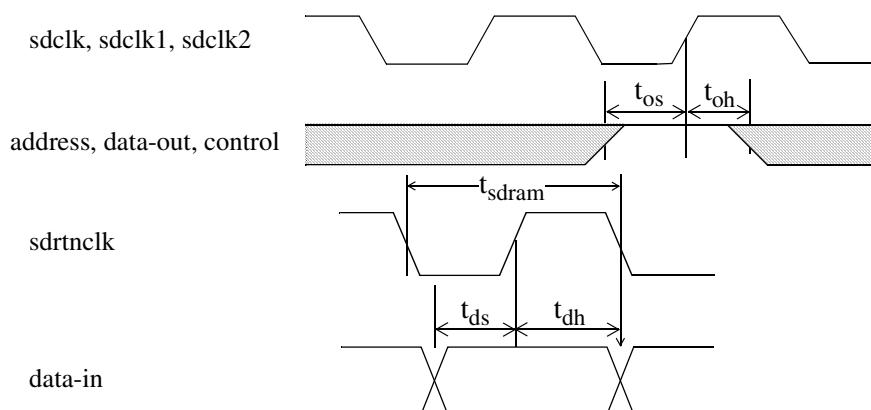


Figure 6-1: SDRAM Timing Measurement Conditions

Correct setup and hold times can be guaranteed through internal delay adjustment, controlled by bits [30:24] of the synchronizer/clock control register in MAP-CA DSP memory block. The SdMrckDly and SdMckDly fields control internal delay circuits that affect the timing of signals going to and from the SDRAM components. They should be adjusted so that setup and hold requirements of both the SDRAM and the MAP-CA DSP are met.

Symbol	Description	Min	Max	Unit
$f_{\text{sdrām}}$ <sup>a</sup>	sdclk frequency	-	133	MHz
$t_{\text{os}}$	Output setup time for address, data, control (sdcs_, sdras_, sdcas_, sdwe_, sddqm)	1.9	-	ns
$t_{\text{oh}}$ <sup>b</sup>	Output hold time of address, data, control (sdcs_, sdras_, sdcas_, sdwe_, sddqm)	1.5	-	ns
$t_{\text{ds}}$ <sup>c, d</sup>	Input data setup time	1	-	ns
$t_{\text{dh}}$ <sup>c, d</sup>	Input data hold time	2	-	ns

Table 6-9: SDRAM Interface Timing Parameters<sup>e</sup>

- $f_{\text{sdrām}} = 1 / t_{\text{sdrām}}$
- The center of the rising edges of sdclk1 and sdclk2 is used as the reference point.
- sdrtnclk is used as the reference clock.
- A matching mechanism is provided to compensate for the propagation delay through circuit board traces to and from the external SDRAM devices. To optimize read timing margin, sdrtnclk should be connected to sdclk with a dedicated trace, with an optional lumped RC load attached to the middle, to account for the number of SDRAM devices attached to the clock line.
- Measured with SdMrckDly = 0 and SdMckDly = 3.

### 6.4.3 PCI Bus Timing

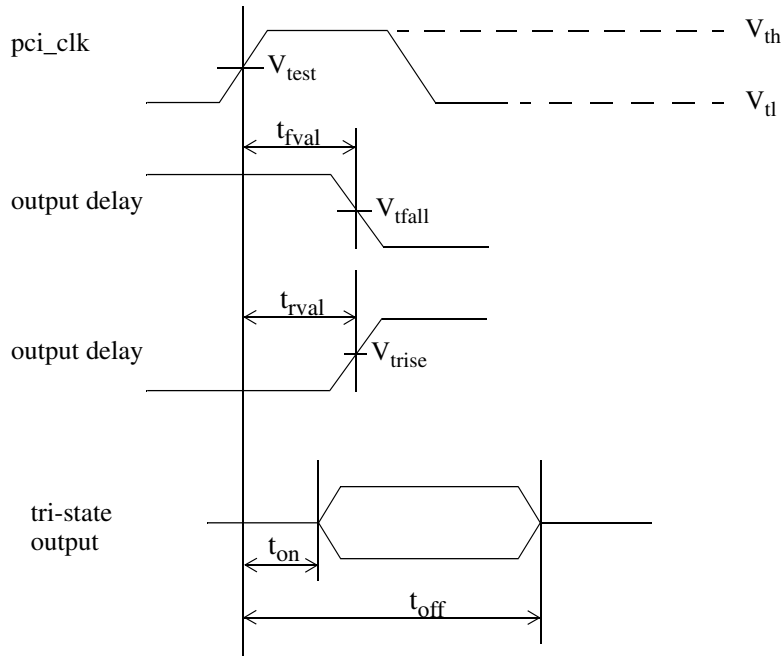


Figure 6-2: PCI Output Timing Measurement Conditions

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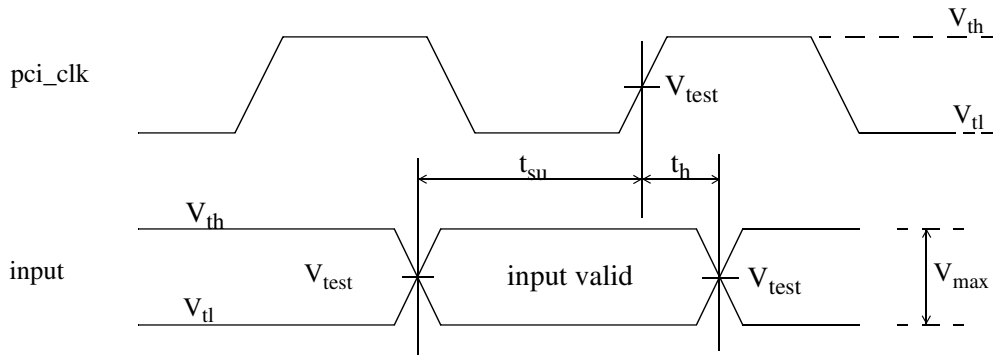


Figure 6-3: PCI Input Timing Measurement Conditions



Symbol	Description	Min	Max	Unit
$t_{\text{pci\_clk}}$	clock cycle time	15	30	ns
$t_{\text{high}}$	clock high time	6	-	ns
$t_{\text{low}}$	clock low time	6	-	ns
$\text{clk}_{\text{slew}}$	clock slew rate	1.5	4	V/ns
$t_{\text{su(bus)}}$	input set up time to clk, bussed signals	3	-	ns
$t_{\text{su(ptp)}}$	input set up time to clk, point-to-point signals	5	-	ns
$t_{\text{val(bus)}}$	clk to signal valid delay, bussed signals	2	6	ns
$t_{\text{val(ptp)}}$	clk to signal valid delay, point to point signal	2	6	ns
$t_{\text{on}}$	float to active delay	2	-	ns
$t_{\text{off}}$	active to float delay	-	14	ns
$t_{\text{h}}$	input hold time from clock	500	-	ps
$t_{\text{rst}}^{\text{a}}$	reset active time after power stable	1	-	ms
$t_{\text{rst-clk}}$	reset active time after clk stable	100	-	$\mu\text{s}$
$t_{\text{rst-off}}$	reset active to output float delay	-	40	ns

Table 6-10: PCI Interface Timing Parameters

- a.  $\text{pci\_rst}_i$  is asserted and de-asserted asynchronously with respect to  $\text{pci\_clk}$ . All output drivers are floated when  $\text{pci\_rst}_i$  is active.

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Symbol	Value	Unit
$V_{\text{max}}$	0.4 $V_{\text{dd33}}$	V
$V_{\text{test}}$	0.4 $V_{\text{dd33}}$	V
$V_{\text{tfall}}$	0.615 $V_{\text{dd33}}$	V
$V_{\text{th}}$	0.6 $V_{\text{dd33}}$	V
$V_{\text{tl}}$	0.2 $V_{\text{dd33}}$	V
$V_{\text{trise}}$	0.285 $V_{\text{dd33}}$	V
input signal slew rate	1.5	V/ns

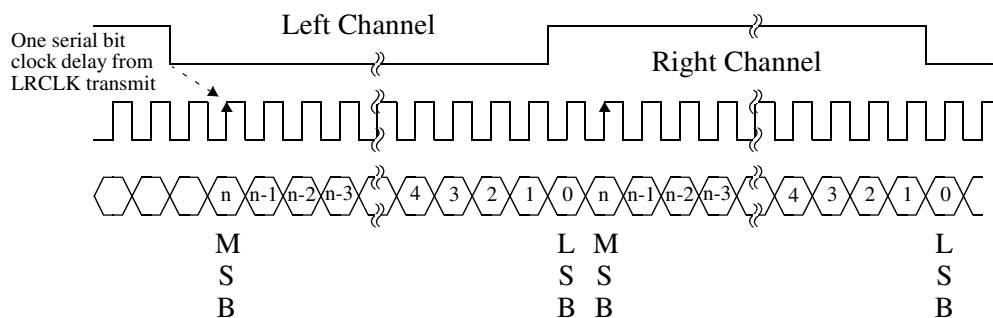
Table 6-11: PCI Measurement Conditions

## 6.4.4 IEC958 Interface Timing

Symbol	Description	Min	Typical	Max	Unit
$F_s$	Audio sample rate		32, 44.1, 48		kHz

Table 6-12: IEC958 Interface Timing Parameters

## 6.4.5 I<sup>2</sup>S Interface Timing

Figure 6-4: I<sup>2</sup>S Data Format

The audio PLL generates audioclk\_out for use by external codecs as master MCLK. MCLK can be programmed for either 12.288 MHz or 16.933 MHz to achieve the desired audio sample rate (LRCLK), according to Table 6-13. The MSB:LSB ordering can be reversed by software programming.

LRCLK <sup>a</sup> (kHz)	MCLK <sup>b</sup> (MHz)	SCLK <sup>c</sup> (MHz)	MCLK/ LRCLK	MCLK/ SCLK	BITS/ FRAME <sup>d</sup>	BITS/ CHANNEL
48	12.288	1.536	256	8	32	16
44.1	16.933 <sup>e</sup>	2.116	384	8	48	24
32	12.288	1.536	384	8	48	24

Table 6-13: I<sup>2</sup>S Clock Ratios

- LRCLK = iis\_in\_lr or iis\_out\_lr.
- MCLK = audioclk\_out; MCLK defaults to 27 MHz on power-up (audio PLL bypassed).
- SCLK = iis\_in\_bclk or iis\_out\_bclk.
- Bits/Frame = 2\*Bits/Channel = SCLK/LRCLK. When MCLK/LRCLK = 384, 24 bits per channel are always transmitted, but the number of valid bits is selectable from 16, 18, 20 or 24.
- MCLK defaults to 16.933 MHz when audio PLL is enabled.

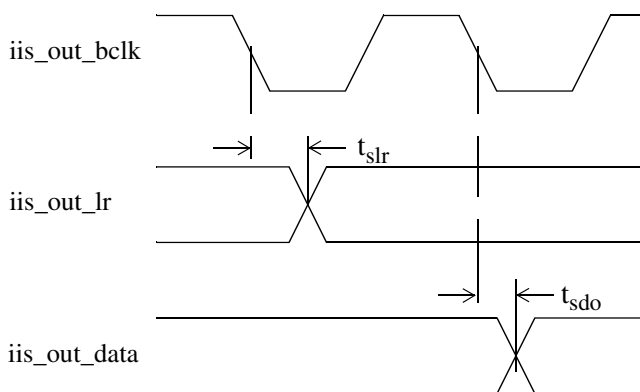


Figure 6-5: I<sup>2</sup>S Output Timing Measurement Conditions

Symbol	Description	Min	Max	Unit
$F_s$	Output sample rate	32, 44.1, 48		kHz
$t_{slr}$	SCLK falling to LRCLK delay	-10	10	ns
$t_{sdo}$	SCLK falling to SDATA valid	-10	10	ns

Table 6-14: I<sup>2</sup>S Output Timing Parameters

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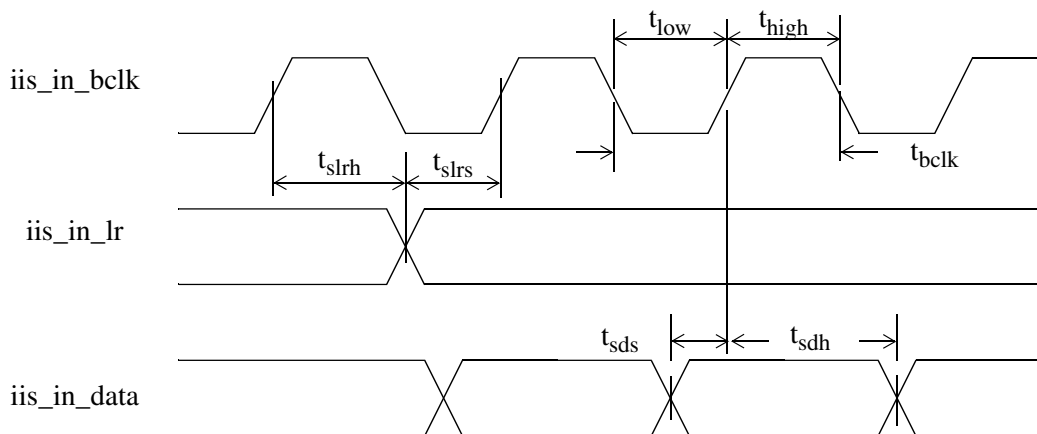
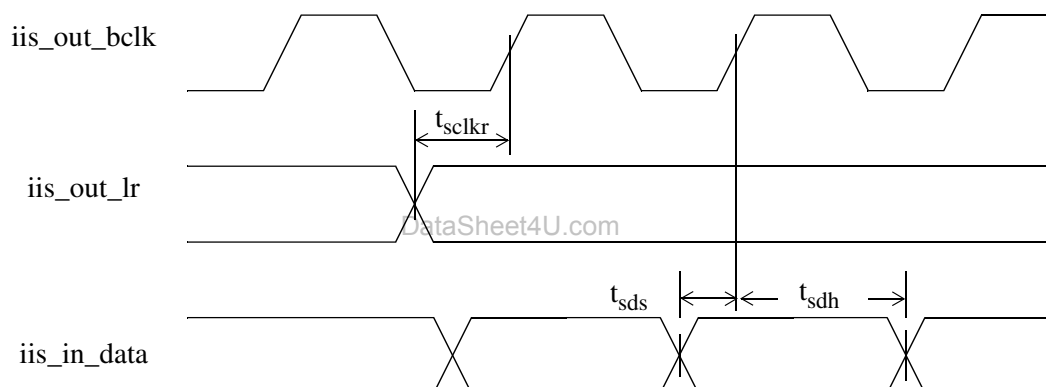


Figure 6-6: I<sup>2</sup>S Input Timing Measurement - Slave Mode

Symbol	Description	Min	Max	Unit
$F_s$	Input sample rate	32, 44.1, 48		kHz
$F_{mclk}$	MCLK frequency	see Table 6-13		From Table
$T_{bclk}$	SCLK period	$8/F_{mclk}$	-	From Table
$T_{low}$	SCLK pulse width low	$2/F_{mclk}$	-	From Table
$T_{high}$	SCLK pulse width high	$2/F_{mclk}$	-	From Table
$t_{slrh}$	SCLK rising to LRCLK hold	20	-	ns
$t_{slrs}$	SCLK rising to LRCLK setup	20	-	ns
$t_{sds}$	SDATA valid to SCLK rising setup	20	-	ns
$t_{sdh}$	SCLK rising to SDATA hold time	20	-	ns

Table 6-15: I<sup>2</sup>S Input Timing Parameters - Slave ModeFigure 6-7: I<sup>2</sup>S Input Timing Measurement - Master Mode

Symbol	Description	Min	Typical	Max	Unit
$F_s$	Input sample rate	-	32, 44.1, 48	-	kHz
$F_{mclk}$	MCLK frequency	see Table 6-13			From Table
$T_{bclk}$	SCLK period	$8/F_{mclk}$	-	-	From Table
$t_{sclkr}$	SCLK rising to LRCLK edge	-	$T_{bclk}/2$	-	$\mu$ s
$t_{sds}$	SDATA valid to SCLK rising setup	$(1/F_{mclk})+10$	-	-	ns
$t_{sdh}$	SCLK rising to SDATA hold	$(1/F_{mclk})+15$	-	-	ns

Table 6-16: I<sup>2</sup>S Input Timing Parameters - Master Mode

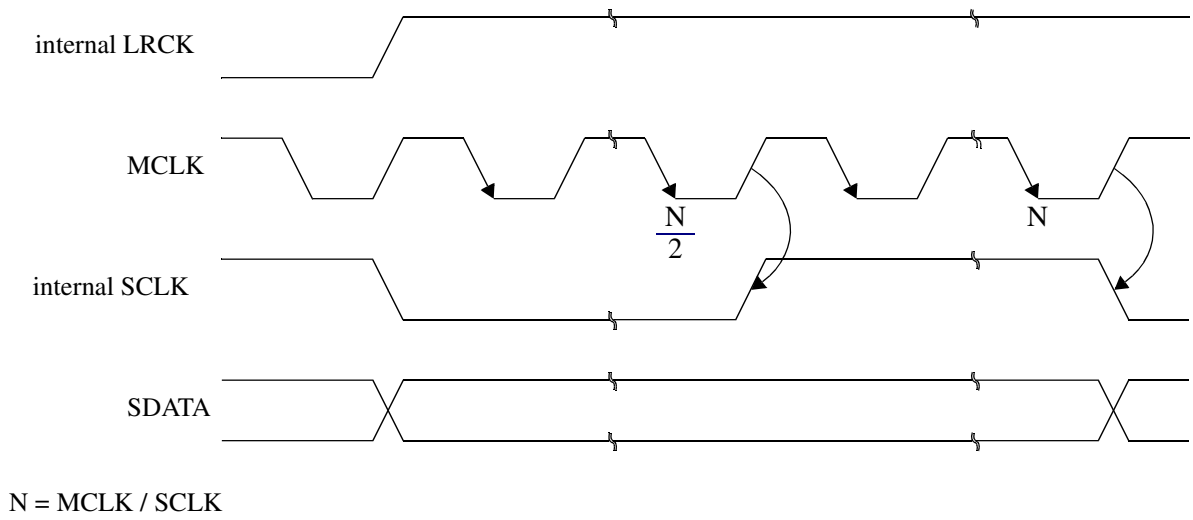


Figure 6-8: Internal Serial Clock Generation for I<sup>2</sup>S Master Mode

### 6.4.6 Transport Channel Interface Timing

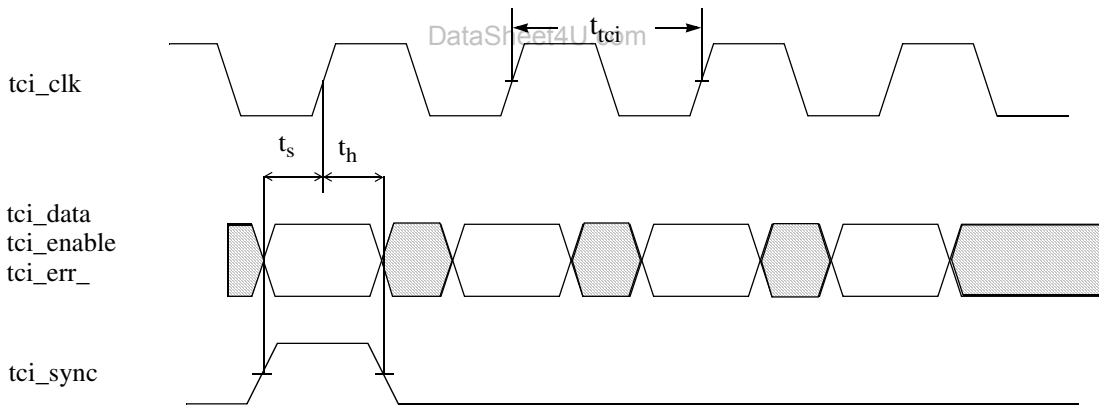


Figure 6-9: TCI Timing Measurement Conditions



**NOTE:** The timing diagram and Table 6-17 are relevant to the primary TCI input port. Timing relationships, input setup, and hold times are identical for the secondary TCI input port. However, the *tci\_vdac* output is not present on the secondary output port. Inputs *video\_ina*[9:0] includes *tci\_data*[7:0] (bits 7:0), *tci\_enable* (bit 8), and *tci\_err\_* (bit 9)

Symbol	Description	Min	Max	Unit
$f_{tci}$	TCI clock frequency <sup>a, b, c, d</sup>	-	27 (parallel) 80 (serial)	MHz
$t_h$	Hold time	1	-	ns
$t_s$	Setup time	4	-	ns

Table 6-17: TCI Timing Parameters

- In parallel mode, it is required that  $t_{core} > f_{tci\_clk}$ .
- In serial mode, it is required that  $t_{core} > \frac{1}{2} f_{tci\_clk}$ .
- For correct audio clock counter operation,  $f_{core}/f_{audio} > 4$ , where  $f_{audio}$  is the frequency of audioclk\_out produced by the on-chip PLL.
- For correct video clock counter operation,  $f_{core}/f_{video} > 4$ , where  $f_{video}$  is the frequency of pclk.

### 6.4.7 ITU-R BT.601/656 Interface Timing

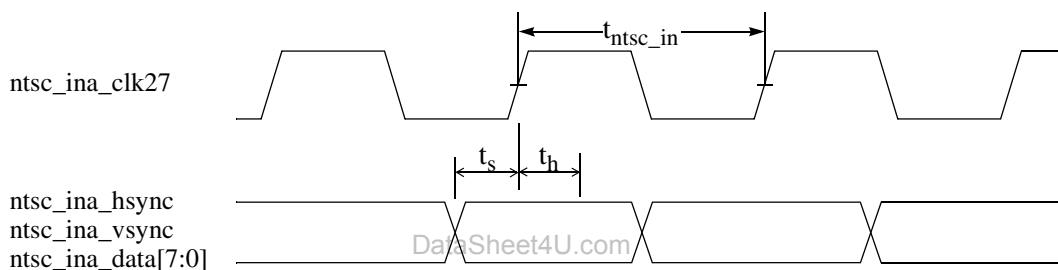


Figure 6-10: ITU-R BT.601/656 Input Timing Measurement Conditions

Symbol	Description	Min	Max	Unit
$f_{ntsc\_in}$	NTSC input clock frequency	-	27	MHz
$t_h$	Input hold time for video_ina[9:0], from cross over of rising clock	1	-	ns
$t_s$	Input setup time video_ina[9:0], to the cross over of rising clock	5	-	ns

Table 6-18: ITU-R BT.601/656 Input Interface Timing Parameters

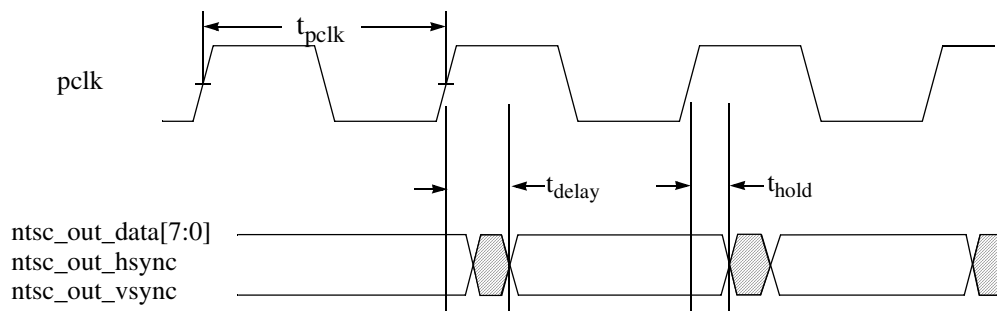


Figure 6-11: ITU-R BT.601/656 Output Timing Measurement Conditions



**NOTE:** This timing diagram and table include information for the primary video input port. The primary video input bus `video_ina[9:0]` includes the byte-wide data bus (bits 7:0), horizontal sync (bit 8), and vertical sync (bit 9). The secondary video input signals have identical timing relationships

Symbol	Description	Min	Max	Unit
$t_{pclk}$	pclk cycle time	18.5	37	ns
$t_{delay}$	Maximum delay time	-	11	ns
$t_{hold}$	Output hold time	4	-	ns

Table 6-19: ITU-R BT.601/656 Output Interface Timing Parameters

## 6.4.8 General Purpose Data Port

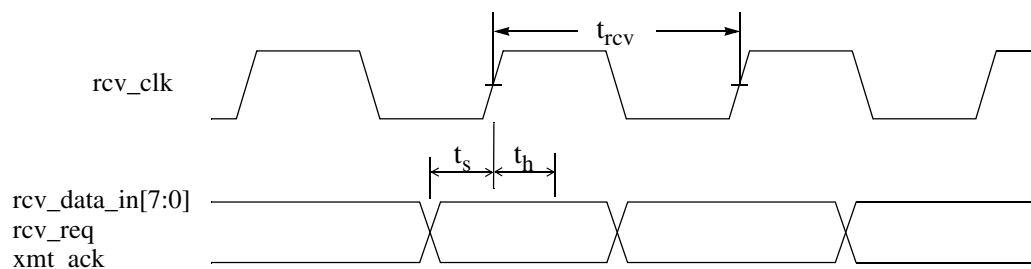


Figure 6-12: GDP Input Timing Measurement Conditions

Symbol	Description	Min	Max	Unit
$t_{rcv}$	Receive clock cycle time	-	60	MHz
$t_h$	Hold time	0	-	ns
$t_s$	Setup time	4	-	ns

Table 6-20: GPDP Input Timing Parameters

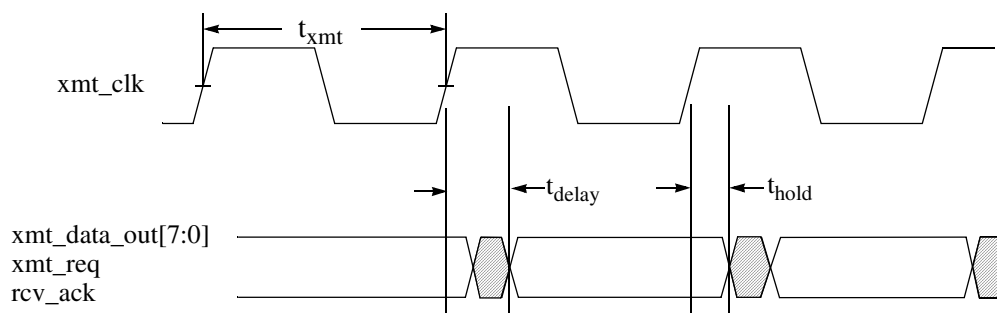
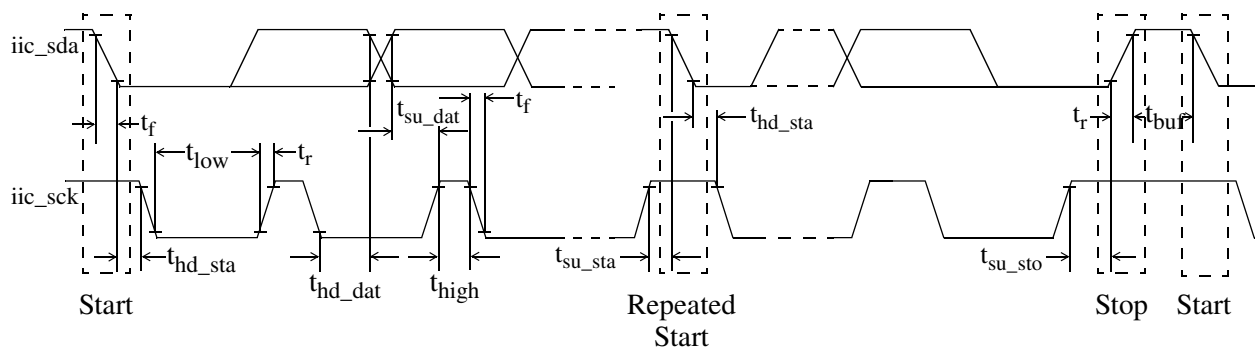


Figure 6-13: GPDP Output Timing Measurement Conditions

Symbol	Description	Min	Max	Unit
$t_{xmt}$	Transmit clock cycle time	-	60	MHz
$t_{hold}$	Output hold time	3.0	-	ns
$t_{delay}$	Maximum delay time	-	8	ns

Table 6-21: GPDP Output Timing Parameters

## 6.4.9 I<sup>2</sup>C Interface Timing

Figure 6-14: I<sup>2</sup>C Timing Measurement Conditions



Symbol <sup>a</sup>	Description	Standard Mode		Fast Mode		Unit
		Min	Max	Min	Max	
$f_{sck}$	iic_sck clock frequency.	0	100	0	400	kHz
$t_{hd\_sta}$	Hold time (repeated) START condition. After this period, the first clock pulse is generated.	2.3	-	0.57	-	$\mu$ s
$t_{low}$	LOW period of the iic_sck clock.	4.7	-	1.27	-	$\mu$ s
$t_{high}$	HIGH period of the iic_sck clock.	4.0	-	0.6	-	$\mu$ s
$t_{su\_sta}$	Setup time for a repeated START condition.	4.7	-	0.6	-	$\mu$ s
$t_{hd\_dat}$	Data hold time.	0 <sup>b</sup>	3.45	0 <sup>b</sup>	0.9	$\mu$ s
$t_{su\_dat}$	Data setup time.	250	-	250	-	ns
$t_r$	Rise time of both iic_sda and iic_sck signals.	-	1000	$20 + 0.1C_b^c$	300	ns
$t_f$	Fall time of both iic_sda and iic_sck signals.	-	300	$20 + 0.1C_b$	300	ns
$t_{su\_sto}$	Setup time for STOP condition.	4.0	-	0.6	-	$\mu$ s
$t_{buf}$	Bus free time between a STOP and START condition.	4.7	-	1.3	-	$\mu$ s
$C_b$	Capacitive load for each bus line.	-	400	-	400	pF

Table 6-22: I<sup>2</sup>C Interface Timing Parameters

- All values referred to  $V_{IHmin}$  and  $V_{ILmax}$  levels. See Table 6-5.
- A device must internally provide a hold time of at least 300 ns for the SDA signal (referred to the  $V_{IHmin}$  of the SCL signal) to bridge the undefined region of the falling edge of SCL.
- $C_b$  = total capacitance of one bus line in pF.

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# Appendix A      Glossary

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These acronyms and names used in this *Data Sheet*. These are their expansion or explanation.

BSDL .....	Boundary Scan Description Language
DataStreamer controller ....	High speed DMA engine that operates independent from VLIW core, also referred to as the DS controller
DRC .....	Display Refresh Controller
DTS .....	Data Transfer Switch - high speed MAP-CA DSP series internal data bus
FEC .....	Forward error correction
I-ALU .....	Integer ALU (Arithmetic Logic Unit) - performs loads, stores, branches, integer arithmetic, and logical operations
IG-ALU .....	Integer, Graphics unit - performs integer arithmetic and (partitioned) multimedia operations
MAP-CA DSP .....	Media Accelerated Processor for Consumer Appliances
MMU .....	Memory Management Unit
PLC .....	128-bit Partitioned Local Constant register
PLV .....	128-bit Partitioned Local Variable register
SIMD .....	Single Instruction Multiple Data
TLB .....	Translation Lookaside Buffer
VLx .....	A coprocessor on MAP-CA DSP that can accelerate variable length encoding and variable length decoding.

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# Appendix B Package Specifications

This appendix to the MAP-CA DSP Datasheet describes the specifications of the 352 pin BGA package

## B.1 Mechanical Specifications

### B.1.1 Outline and Footprint

The following sections use millimeter (mm) as the unit of measure.

#### B.1.1.1 Top and Bottom Views

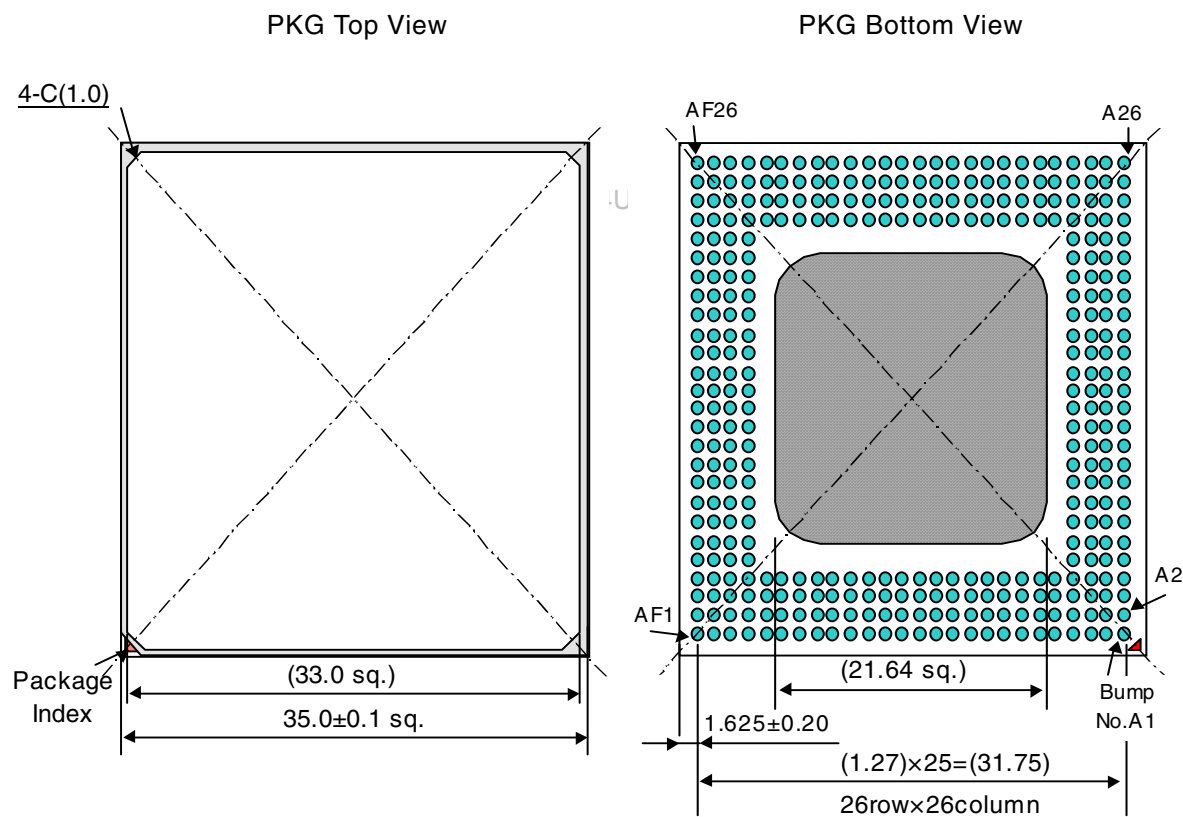


Figure B-1: 352 Pin BGA Outline and Footprint - Top and Bottom Views

### B.1.1.2 Side View

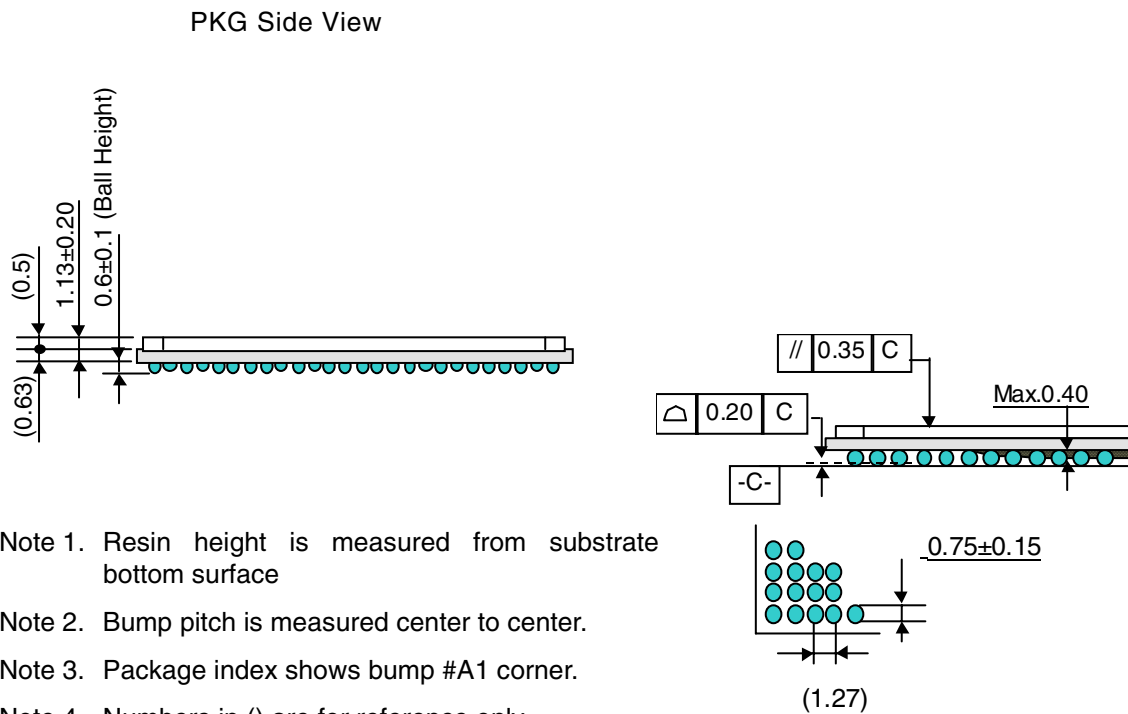


Figure B-2: 352 Pin BGA Outline and Footprint - Side Views

## B.2 Package Materials

### B.2.1 Materials Specification

Segment	Material
Package substrate	BT Resin
Encapsulation	Epoxy with filler
Heat Spreader	Cu + Ni plating
Solder Ball	37 Pb - 63 Sn

Table B-1: Material Used.

## B.2.2 Index Location

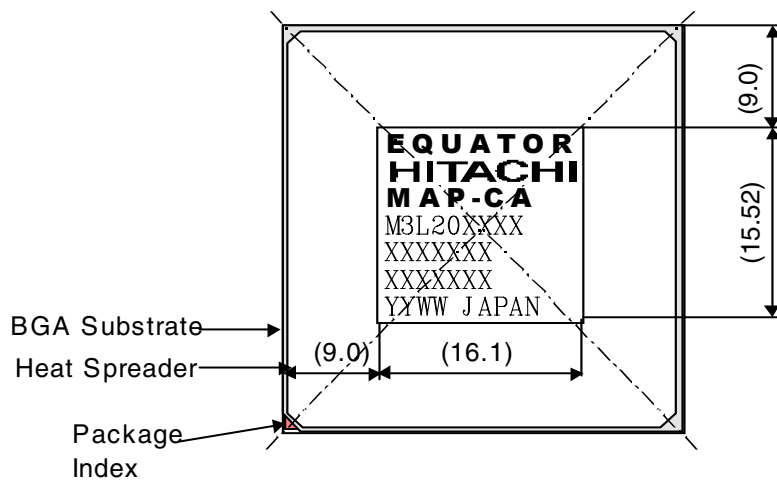


Figure B-3: Index Location

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