# 2.25MHz 1A Synchronous Step-Down Converter

## General Description

The RT8057A is a high efficiency Pulse-Width-Modulated (PWM) step-down DC/DC converter, capable of delivering 1A output current over a wide input voltage range from 2.7V to 5.5V. The RT8057A is ideally suited for portable electronic devices that are powered from 1-cell Li-ion battery or from other power sources such as cellular phones, PDAs, hand-held devices, game console and related accessories.

The internal synchronous rectifier with low $R_{DS(ON)}$ dramatically reduces conduction loss at PWM mode. No external Schottky diode is required in practical applications. The RT8057A enters Low Dropout Mode when normal Pulse -Width Mode cannot provide regulated output voltage by continuously turning on the upper P-MOSFET. The RT8057A enters shut-down mode and consumes less than 1μA when the EN pin is pulled low. The switching ripple is easily smoothed-out by small package filtering elements due to a fixed operating frequency of 2.25MHz.

The RT8057A is available in a small WDFN-6SL 2x2 package.

## Ordering Information

RT8057A□□(2)

- Pin 1 Orientation
  (2) : Quadrant 2, Follow EIA-481-D
  (For WDFN-6SL 2x2 Only)
- Package Type
  QW : WDFN-6SL 2x2 (W-Type)
  J5 : TSOT-23-5
- Lead Plating System
  G : Green (Halogen Free and Pb Free)

Note :

Richtek products are :

- RoHS compliant and compatible with the current requirements of IPC/JEDEC J-STD-020.
- Suitable for use in SnPb or Pb-free soldering processes.

## Features

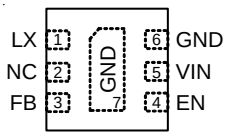
- **2.7V to 5.5V Wide Input Operation Range**
- **2.25MHz Fixed-Frequency PWM Operation**
- **Up to 1A Output Current**
- **Up to 90% Efficiency**
- **0.6V Reference Allows Low Output Voltage**
- **Internal Soft-Start**
- **No Schottky Diode Required**
- **Internal Compensation to Reduce External Components**
- **Low Dropout Operation : 100% Duty Cycle**
- **RoHS Compliant and Halogen Free**

## Applications

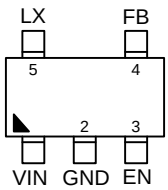
- Portable Instruments
- Game Console and Accessories
- Microprocessors and DSP Core Supplies
- Cellular Phones
- Wireless and DSL Modems
- PC Cards

## Pin Configurations

(TOP VIEW)




WDFN-6SL 2x2

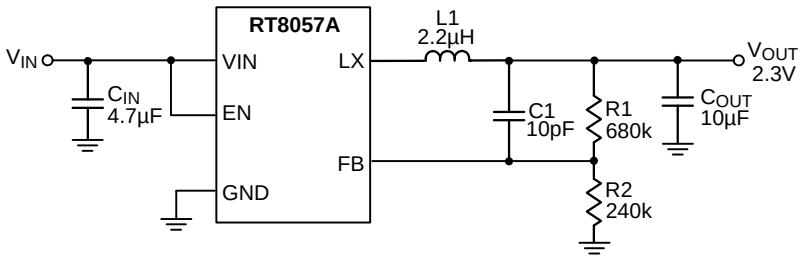



TSOT-23-5

## Marking Information

For marking information, contact our sales representative directly or through a Richtek distributor located in your area.

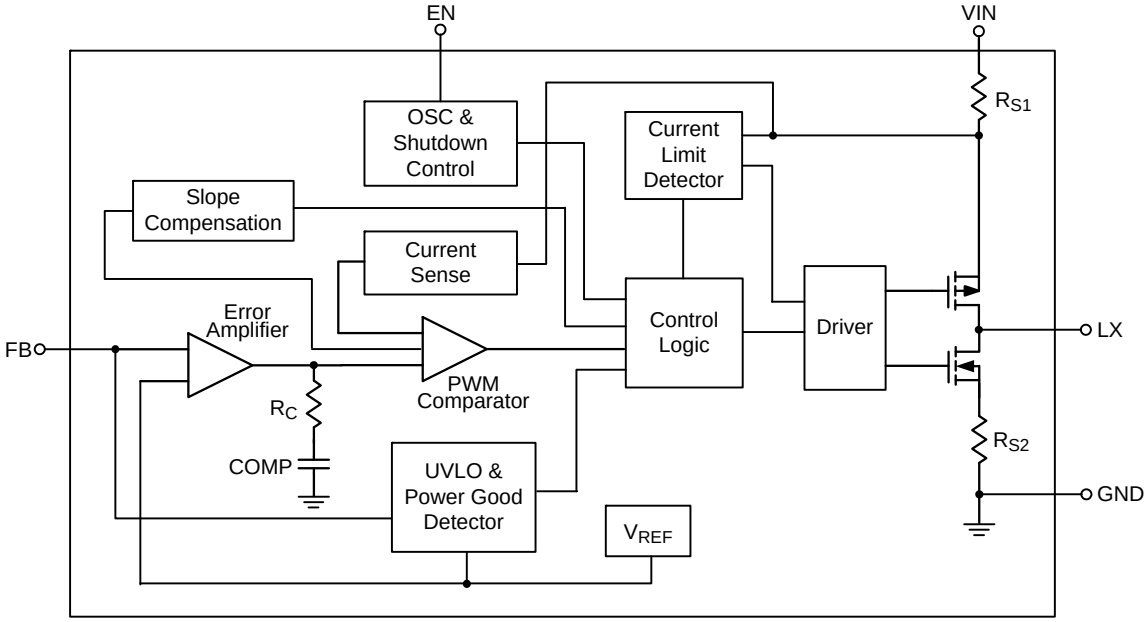
## Typical Application Circuit




## Function Pin Description

| Pin No. | | Pin Name | Pin Function |
|---|---|---|---|
| WDFN-6SL 2x2 | TSOT-23-5 | | |
| 1 | 5 | LX | Switch Node. Connect to the external inductor. |
| 2 | -- | NC | No Internal Connection. Connect to GND. |
| 3 | 4 | FB | Feedback Pin. Connect to the external resistor divider. The FB reference voltage is 0.6V typically. |
| 4 | 3 | EN | Chip Enable (Active High). |
| 5 | 1 | VIN | Power Input. Connect to the input capacitor. |
| 6, 7 (Exposed Pad) | 2 | GND | Power GND. The Exposed Pad must be soldered to a large PCB and connected to GND for maximum power dissipation. |

## Function Block Diagram



## Absolute Maximum Ratings (Note 1)

- Supply Input Voltage, $V_{IN}$ ---------- 6.5V
- Power Dissipation, $P_D$ @ $T_A$ = 25°C
  - WDFN-6SL 2x2 ---------- 0.833W
  - TSOT-23-5 ---------- 0.625W
- Package Thermal Resistance (Note 2)
  - WDFN-6SL 2x2, $\theta_{JA}$ ---------- 120°C/W
  - WDFN-6SL 2x2, $\theta_{JC}$ ---------- 8.2°C/W
  - TSOT-23-5, $\theta_{JA}$ ---------- 160°C/W
- Lead Temperature (Soldering, 10 sec.) ---------- 260°C
- Junction Temperature ---------- 150°C
- Storage Temperature Range ---------- –65°C to 150°C
- ESD Susceptibility (Note 3)
  - HBM ---------- 2kV
  - MM ---------- 200V

## Recommended Operating Conditions (Note 4)

- Supply Input Voltage, $V_{IN}$ ---------- 2.7V to 5.5V
- Junction Temperature Range ---------- –40°C to 125°C
- Ambient Temperature Range ---------- –40°C to 85°C

## Electrical Characteristics

($V_{IN}$ = 3.6V, $T_A$ = 25°C unless otherwise specified)

| Parameter | | Symbol | Test Conditions | Min | Typ | Max | Unit |
|---|---|---|---|---|---|---|---|
| Output Current | | $I_{OUT}$ | $V_{IN}$ = 2.7V to 5.5V | -- | -- | 1 | A |
| Quiescent Current | | $I_Q$ | $I_{OUT}$ = 0mA | -- | 81 | -- | μA |
| Reference Voltage | | $V_{REF}$ | | –2 | -- | 2 | % |
| | | | Note 5 | –2.5 | -- | 2.5 | |
| Under Voltage Lockout Threshold | | $V_{UVLO}$ | $V_{IN}$ Rising | 2 | 2.2 | 2.4 | V |
| | | | Hysteresis | -- | 0.2 | -- | V |
| Shutdown Current | | $I_{SHDN}$ | | -- | 0.1 | 1 | μA |
| Switching Frequency | | | | -- | 2.25 | -- | MHz |
| EN Threshold Voltage | Logic-High | $V_{IH}$ | | 1 | -- | $V_{IN}$ | V |
| | Logic-Low | $V_{IL}$ | | -- | -- | 0.4 | V |
| Thermal Shutdown Temperature | | $T_{SD}$ | | -- | 150 | -- | °C |
| Switch On Resistance | High Side | $R_{DS(ON)\_H}$ | $I_{SW}$ = 0.2A | -- | 250 | -- | mΩ |
| | Low Side | $R_{DS(ON)\_L}$ | $I_{SW}$ = 0.2A | -- | 200 | -- | mΩ |
| Peak Current Limit | | $I_{LIM}$ | | 1.1 | 1.5 | 2 | A |
| Output Voltage Line Regulation | | | $V_{IN}$ = 2.7V to 5.5V | -- | -- | 1 | %/V |
| Output Voltage Load Regulation | | | 0mA < $I_{OUT}$ < 0.6A | -- | -- | 1 | % |
| Start-Up Time | | $t_{ss}$ | Guaranteed by Design | 200 | 300 | 400 | μs |

**Note 1.** Stresses listed as the above "Absolute Maximum Ratings" may cause permanent damage to the device. These are for stress ratings. Functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may remain possibility to affect device reliability.

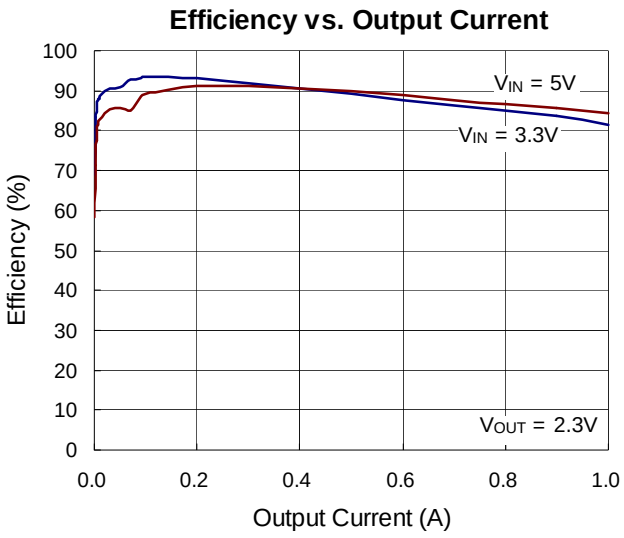
**Note 2.** $\theta_{JA}$ is measured in natural convection at $T_A$ = 25°C on a high effective thermal conductivity four-layer test board of JEDEC 51-7 thermal measurement standard. The measurement case position of $\theta_{JC}$ is on the exposed pad of the package.

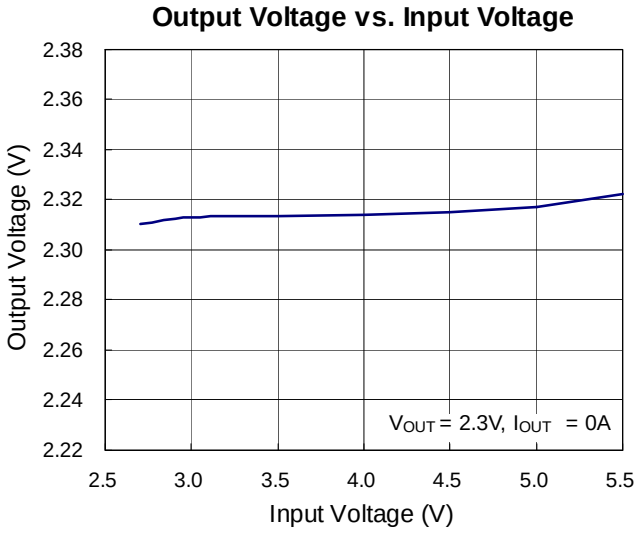
**Note 3.** Devices are ESD sensitive. Handling precaution is recommended.

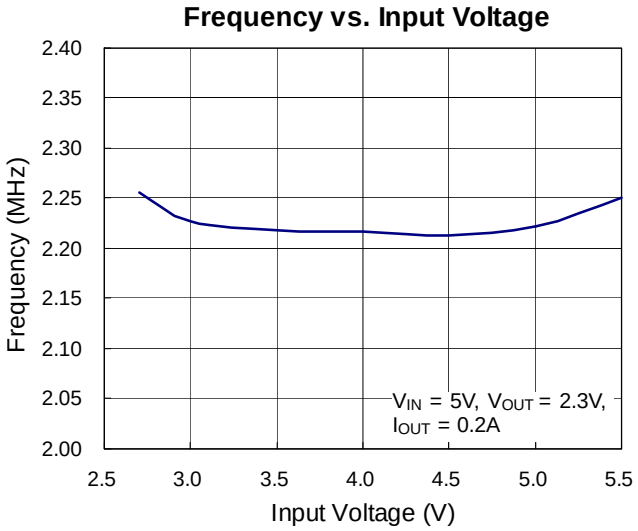
**Note 4.** The device is not guaranteed to function outside its operating conditions.

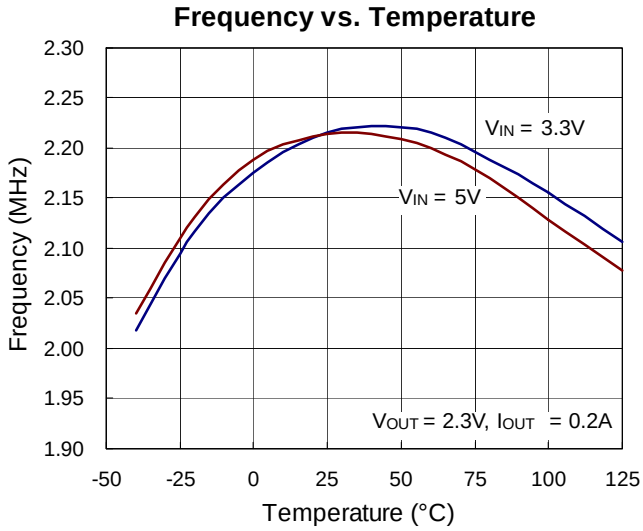
**Note 5.** The reference voltage accuracy is ±2.5% at recommended ambient temperature range, guaranteed by design.

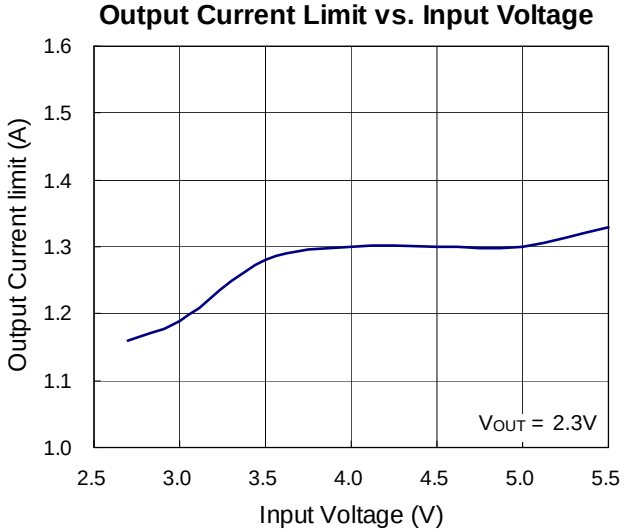
## Typical Operating Characteristics

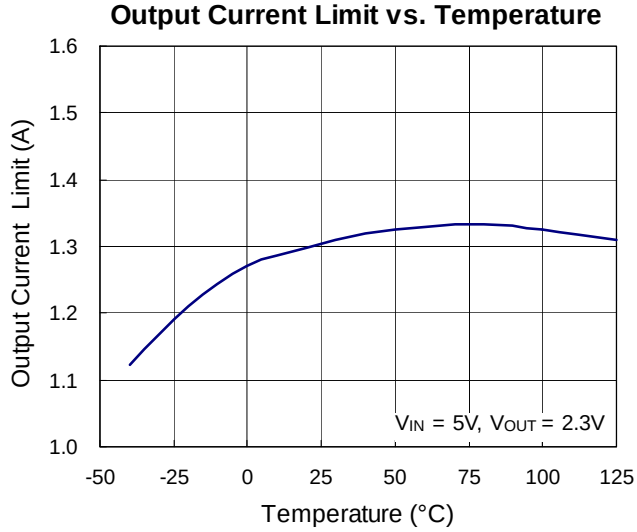


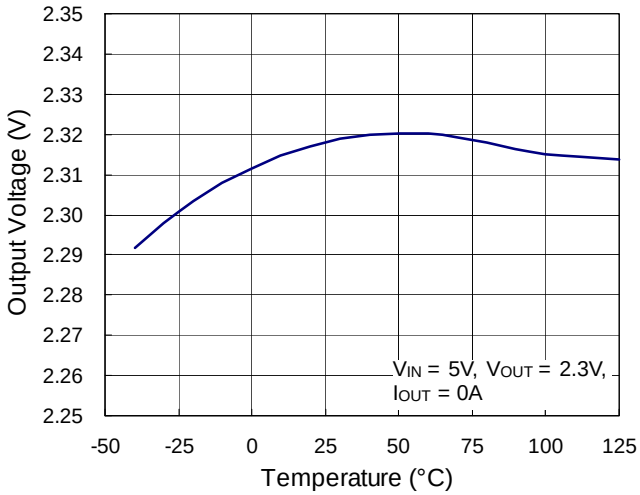



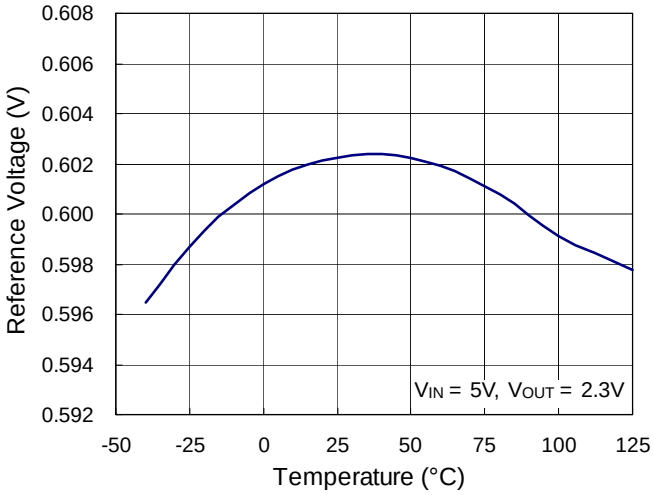



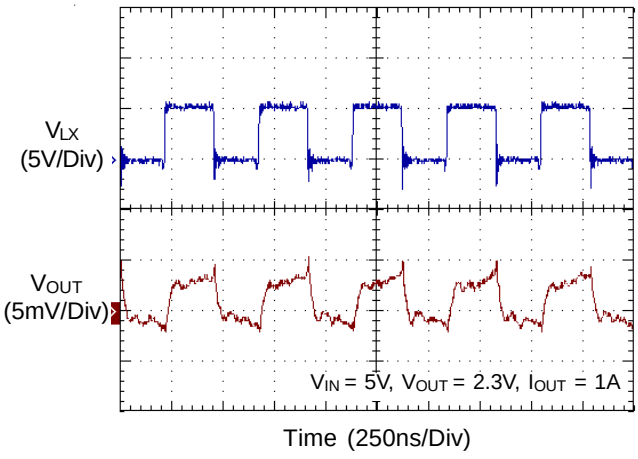
### Output Voltage vs. Temperature



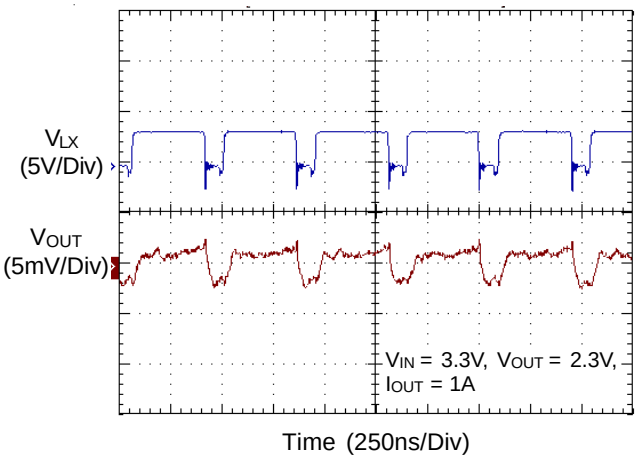

### Reference Voltage vs. Temperature



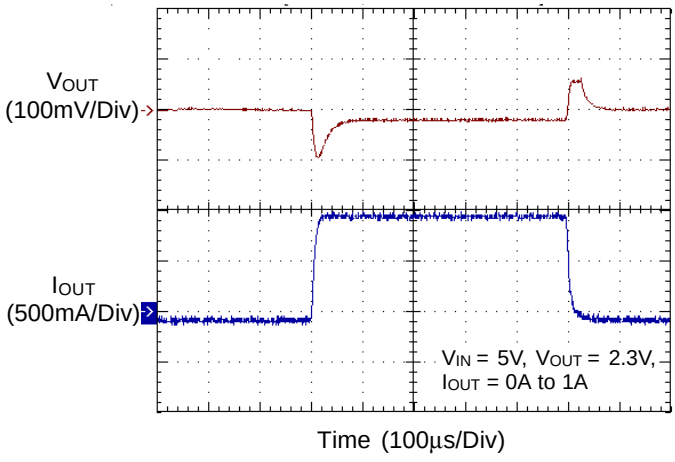

### Output Ripple




### Output Ripple




### Load Transient Response

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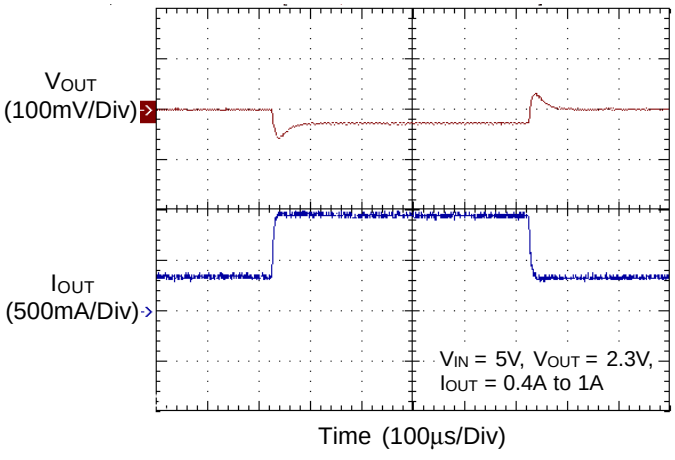
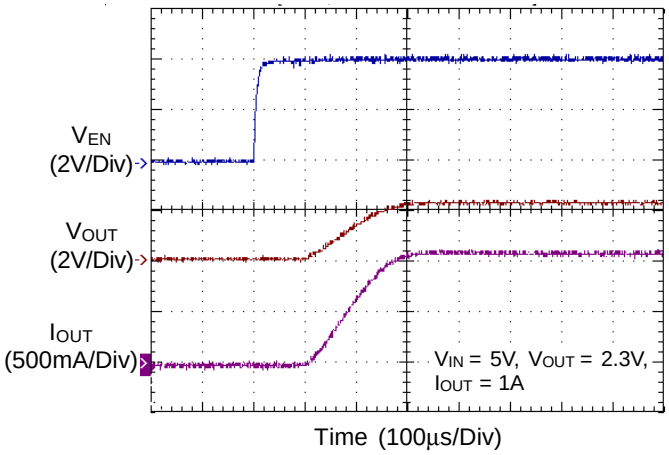
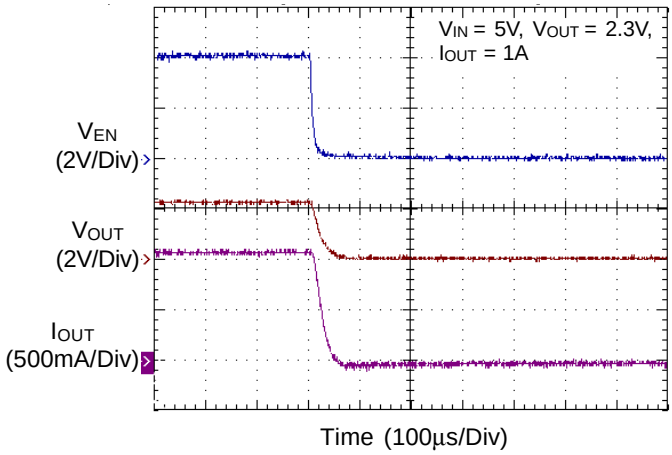


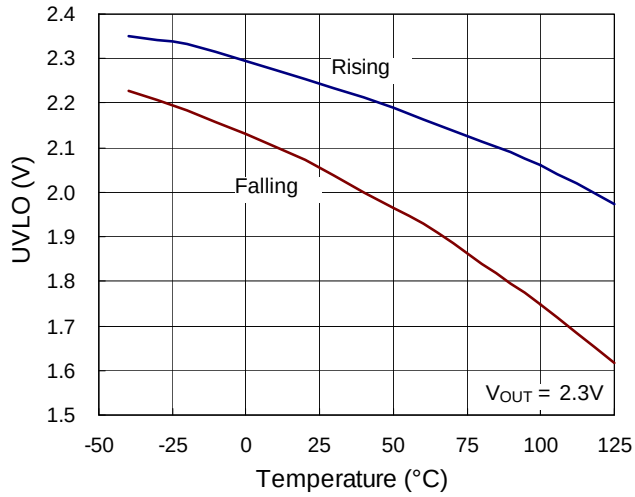
### Power On from EN



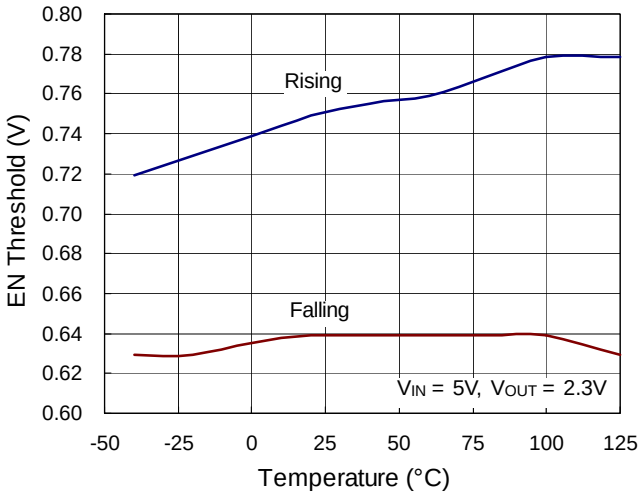

### Power Off from EN



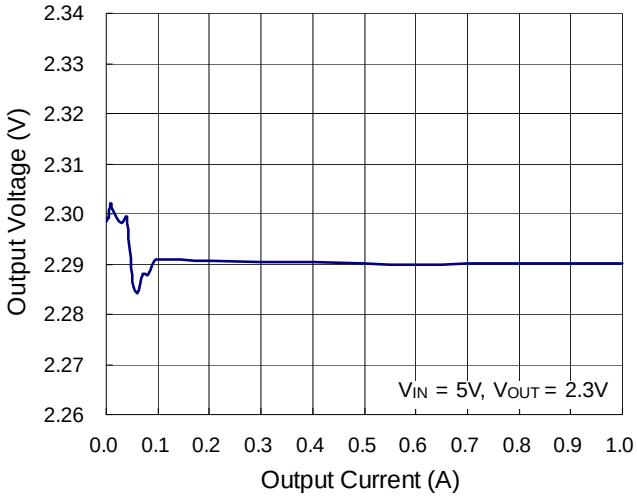

### UVLO vs. Temperature




### EN Threshold vs. Temperature




### Output Voltage vs. Output Current



## Application Information

The basic RT8057A application circuit is shown in Typical Application Circuit. External component selection is determined by the maximum load current and begins with the selection of the inductor value and operating frequency followed by $C_{IN}$ and $C_{OUT}$.

### Output Voltage Setting

The output voltage is set by an external resistive divider according to the following equation :

$$V_{OUT} = V_{REF} \times (1+\frac{R1}{R2})$$

where $V_{REF}$ equals to 0.6V typical. The resistive divider allows the FB pin to sense a fraction of the output voltage as shown in Figure 1.

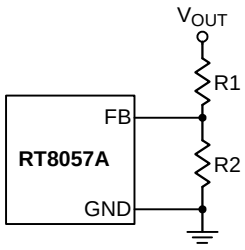



Figure 1. Setting the Output Voltage

### Soft-Start

The RT8057A contains an internal soft-start clamp that gradually raises the clamp on the FB pin.

### 100% Duty Cycle Operation

When the input supply voltage decreases toward the output voltage, the duty cycle increases toward the maximum on-time. Further reduction of the supply voltage forces the main switch to remain on for more than one cycle, eventually reaching 100% duty cycle.

The output voltage will then be determined by the input voltage minus the voltage drop across the internal P-MOSFET and the inductor.

### Low Supply Operation

The RT8057A is designed to operate down to an input supply voltage of 2.7V. One important consideration at low input supply voltages is that the $R_{DS(ON)}$ of the P-Channel and N-Channel power switches increases. The user should calculate the power dissipation when the RT8057A is used at 100% duty cycle with low input voltages to ensure that thermal limits are not exceeded.

### Under Voltage Protection (UVP)

The output voltage can be continuously monitored for under voltage protection. When the output voltage is less than 33% of its set voltage threshold after OCP occurs, the under voltage protection circuit will be triggered to auto re-softstart.

### Input Voltage Over Voltage protection ($V_{IN}$ OVP)

When the input voltage ($V_{IN}$) is higher than 6V, $V_{IN}$ OVP will be triggered and the IC stops switching. Once the input voltage drops below 6V, the IC will return to normal operation.

### Output Over Voltage Protection ($V_{OUT}$ OVP)

When the output voltage exceeds more than 5% of the nominal reference voltage, the feedback loop forces the internal switches off within 50μs. Therefore, the output over voltage protection is automatically triggered by the loop.

### Short Circuit Protection

When the output is shorted to ground, the inductor current decays very slowly during a single switching cycle. A current runaway detector is used to monitor inductor current. As current increases beyond the control of current loop, switching cycles will be skipped to prevent current runaway from occurring.

**Table 1. Inductors**

| Component Supplier | Series | Inductance (μH) | DCR (mΩ) | Current Rating (mA) | Dimensions (mm) |
|---|---|---|---|---|---|
| TAIYO YUDEN | NR4018 T2R2M | 2.2μH | 60 | 2700 | 4 X 4 X 1.8 |

### $C_{IN}$ and $C_{OUT}$ Selection

The input capacitance, $C_{IN}$, is needed to filter the trapezoidal current at the source of the top MOSFET. To prevent large ripple voltage, a low ESR input capacitor sized for the maximum RMS current should be used. RMS current is given by :

$$I_{RMS} = I_{OUT(MAX)} \frac{V_{OUT}}{V_{IN}} \sqrt{\frac{V_{IN}}{V_{OUT}} - 1}$$

This formula has a maximum at $V_{IN} = 2V_{OUT}$, where $I_{RMS} = I_{OUT}/2$. This simple worst case condition is commonly used for design because even significant deviations do not result in much difference. Choose a capacitor rated at a higher temperature than required.

Several capacitors may also be paralleled to meet size or height requirements in the design.

The selection of $C_{OUT}$ is determined by the effective series resistance (ESR) that is required to minimize voltage ripple and load step transients, as well as the amount of bulk capacitance that is necessary to ensure that the control loop is stable. Loop stability can be checked by viewing the load transient response. The output ripple, $\Delta V_{OUT}$, is determined by :

$$\Delta V_{OUT} \leq \Delta I_L \left[ ESR + \frac{1}{8fC_{OUT}} \right]$$

The output ripple is highest at maximum input voltage since $\Delta I_L$ increases with input voltage. Multiple capacitors placed in parallel may be needed to meet the ESR and RMS current handling requirements. Dry tantalum, special polymer, aluminum electrolytic and ceramic capacitors are all available in surface mount packages. Special polymer capacitors offer very low ESR, but have lower capacitance density than other types. Tantalum capacitors have the highest capacitance density, but it is important to only use types that have been surge tested for use in switching power supplies. Aluminum electrolytic capacitors have significantly higher ESR, but can be used in cost-sensitive applications provided that consideration is given to ripple current ratings and long term reliability. Ceramic capacitors have excellent low ESR characteristics, but can have a high voltage coefficient and audible piezoelectric effects.

The high Q of ceramic capacitors with trace inductance can also lead to significant ringing.

### Using Ceramic Input and Output Capacitors

Higher value, lower cost ceramic capacitors are now becoming available in smaller case sizes. Their high ripple current, high voltage rating and low ESR make them ideal for switching regulator applications. However, care must be taken when these capacitors are used at the input and output. When a ceramic capacitor is used at the input and the power is supplied by a wall adapter through long wires, a load step at the output can induce ringing at the input, $V_{IN}$. At best, this ringing can couple to the output and be mistaken as loop instability. At worst, a sudden inrush of current through the long wires can potentially cause a voltage spike at $V_{IN}$ large enough to damage the part.

**Table 2. Capacitors for $C_{IN}$ and $C_{OUT}$**

| Component Supplier | Part No. | Capacitance (μF) | Case Size |
|---|---|---|---|
| MuRata | GRM31CR71A475KA01 | 4.7μF | 1206 |
| MuRata | GRM31CR71A106KA01 | 10μF | 1206 |

### Thermal Considerations

For continuous operation, do not exceed absolute maximum junction temperature. The maximum power dissipation depends on the thermal resistance of the IC package, PCB layout, rate of surrounding airflow, and difference between junction and ambient temperature.

The maximum power dissipation can be calculated by the following formula :

$P_{D(MAX)} = (T_{J(MAX)} - T_A) / \theta_{JA}$

where $T_{J(MAX)}$ is the maximum junction temperature, $T_A$ is the ambient temperature, and $\theta_{JA}$ is the junction to ambient thermal resistance.

For recommended operating condition specifications of the RT8057A, the maximum junction temperature is 125°C and $T_A$ is the ambient temperature. The junction to ambient thermal resistance, $\theta_{JA}$, is layout dependent. For WDFN-6SL 2x2 package, the thermal resistance, $\theta_{JA}$, is 120°C/W on a standard JEDEC 51-7 four-layer thermal test board. For TSOT-23-5 package, the thermal resistance, $\theta_{JA}$, is 160°C/W on a standard JEDEC 51-7 four-layer thermal test board. The maximum power dissipation at $T_A$ = 25°C can be calculated by the following formula :

$P_{D(MAX)}$ = (125°C – 25°C) / (120°C/W) = 0.833W for WDFN-6SL 2x2 package

$P_{D(MAX)}$ = (125°C – 25°C) / (160°C/W) = 0.625W for TSOT-23-5 package

he maximum power dissipation depends on the operating ambient temperature for fixed $T_{J(MAX)}$ and thermal resistance, $\theta_{JA}$. For the RT8057A packages, the derating curve in Figure 1 allows the designer to see the effect of rising ambient temperature on the maximum power dissipation.

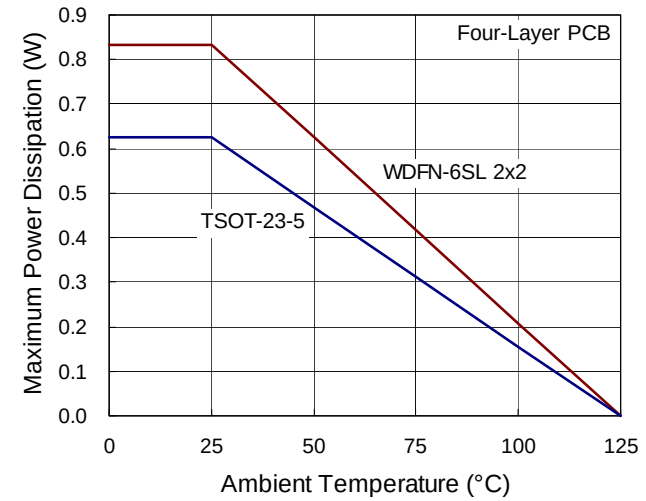
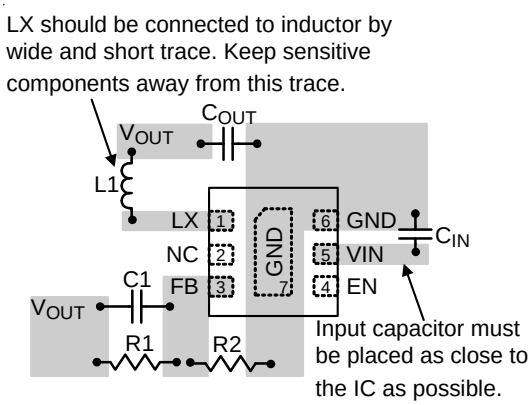



Figure 2. Derating Curve for the RT8057A Package

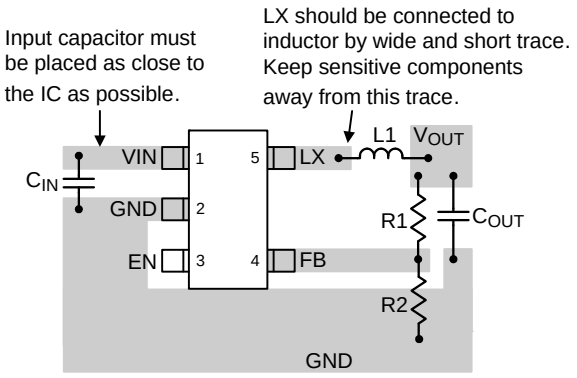
## Layout Considerations

Follow the PCB layout guidelines for optimal performance of the RT8057A.

- Connect the terminal of the input capacitor(s), $C_{IN}$, as close as possible to the VIN pin. This capacitor provides the AC current into the internal power MOSFETs.
- LX node experiences high frequency voltage swing and should be kept within a small area. Keep all sensitive small-signal nodes away from the LX node to prevent stray capacitive noise pick up.
- Flood all unused areas on all layers with copper. Flooding with copper will reduce the temperature rise of power components. Connect the copper areas to any DC net ($V_{IN}$, $V_{OUT}$, GND, or any other DC rail in the system).
- Connect the FB pin directly to the feedback resistors. The resistive voltage divider must be connected between $V_{OUT}$ and GND.



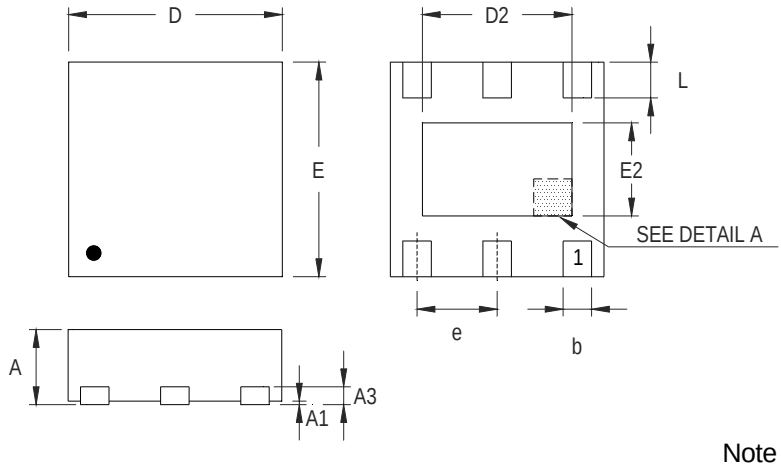

(a) For WDFN-6SL 2x2 Package

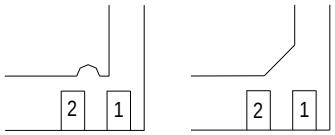



(a) For TSOT-23-5 Package

Figure 3. PCB Layout Guide

## Outline Dimension



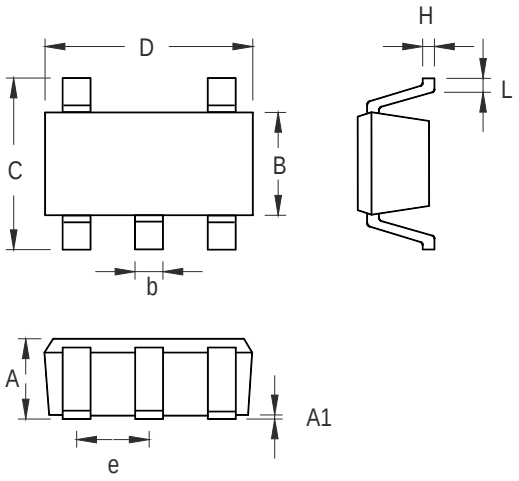

**DETAIL A**

Pin #1 ID and Tie Bar Mark Options

Note : The configuration of the Pin #1 identifier is optional, but must be located within the zone indicated.

| Symbol | Dimensions In Millimeters | | Dimensions In Inches | |
|---|---|---|---|---|
| | Min | Max | Min | Max |
| A | 0.700 | 0.800 | 0.028 | 0.031 |
| A1 | 0.000 | 0.050 | 0.000 | 0.002 |
| A3 | 0.175 | 0.250 | 0.007 | 0.010 |
| b | 0.200 | 0.350 | 0.008 | 0.014 |
| D | 1.900 | 2.100 | 0.075 | 0.083 |
| D2 | 1.550 | 1.650 | 0.061 | 0.065 |
| E | 1.900 | 2.100 | 0.075 | 0.083 |
| E2 | 0.950 | 1.050 | 0.037 | 0.041 |
| e | 0.650 | | 0.026 | |
| L | 0.200 | 0.300 | 0.008 | 0.012 |

**W-Type 6SL DFN 2x2 Package**




| Symbol | Dimensions In Millimeters | | Dimensions In Inches | |
|---|---|---|---|---|
| | Min | Max | Min | Max |
| A | 0.700 | 1.000 | 0.028 | 0.039 |
| A1 | 0.000 | 0.100 | 0.000 | 0.004 |
| B | 1.397 | 1.803 | 0.055 | 0.071 |
| b | 0.300 | 0.559 | 0.012 | 0.022 |
| C | 2.591 | 3.000 | 0.102 | 0.118 |
| D | 2.692 | 3.099 | 0.106 | 0.122 |
| e | 0.838 | 1.041 | 0.033 | 0.041 |
| H | 0.080 | 0.254 | 0.003 | 0.010 |
| L | 0.300 | 0.610 | 0.012 | 0.024 |

**TSOT-23-5 Surface Mount Package**

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